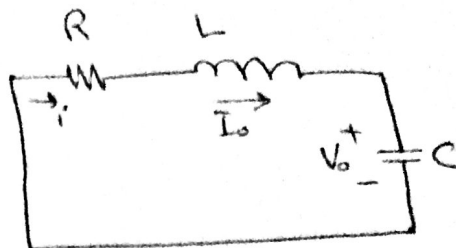


⇒ Natural Response of Series RLC Circuit

* Consider series RLC circuit as shown →
The circuit is being excited by the energy initially stored in the capacitor and inductor.



V_0 : initial voltage of capacitor
 I_0 : " inductor current.

By applying KVL:

$$iR + L \frac{di}{dt} + \frac{1}{C} \int_0^t i dt + V_0 = 0$$

⏟
voltage through capacitor
+ initial voltage in it.

To eliminate the Integral, we differentiate w.r.t 't'.
we get:

$$R \frac{di}{dt} + L \frac{d^2i}{dt^2} + \frac{i}{C} = 0$$

$$\therefore \frac{1}{C} \int_0^t i dt \times \frac{d}{dt} = \frac{i}{C}$$

$$\frac{d^2i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = 0$$

Here we see current w.r.t time 't'.

From previous practice on 1st order circuits we experience that solution is in exponential form.

So we assume: $i(t) = A e^{st}$ — (A)

where 'A' and 's' are constants to be determined

Now we write the characteristic eq. as:

$$s^2 + \frac{R}{L} s + \frac{1}{LC} = 0 \text{ — (B)}$$

We can solve it's quadratic eq as:

$$s_{1,2} = -\frac{R}{2L} \pm \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}} \quad \text{--- (C)}$$

$$s_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

$$s_2 = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}}$$

we get two roots of above eq.

Q.E:
 eg. $5x^2 + 3x + 3 = 0$
 $\rightarrow ax^2 + bx + c = 0$
 $x = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$

A more compact way of expressing those roots is:

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2}$$

$$s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2}$$

--- (D)

where $\alpha = \frac{R}{2L}$ and $\omega_0 = \frac{1}{\sqrt{LC}}$

- α is neper freq. expressed in n/s or damping factor.
- ω_0 " resonant " or undamped natural freq. in rad/s
- Neper freq: is measure of how circuit will die away.
- Resonant " : causing electric energy to oscillate between electric field of capacitor and magnetic field of inductor.

We can also write characteristic eq in (B) as:

$$s^2 + 2\alpha s + \omega_0^2 = 0$$

From eq. (C) we can say that there are two possible solutions for 'i' as:

$$i_1 = A_1 e^{s_1 t}$$

$$i_2 = A_2 e^{s_2 t}$$

So, the overall natural response of series RLC circuit is

$$i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$$

From eq (D) we infer that there are three types of solutions. (2)

- ① If $\alpha > \omega_0$: Overdamped case (when the roots of characteristic eq. are unequal and real).
- ② $\alpha = \omega_0$: Critically damped case (when the roots are equal & real)
- ③ $\alpha < \omega_0$: Underdamped case (when the roots are complex).

→ The Overdamped Case ($\alpha > \omega_0$):- We know from eq (C) and (E) that $C > \frac{4L}{R^2}$. When this happens both roots s_1 and s_2 are negative R^2 and real. So the response is.

$$i(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$$

→ The Critically Damped Case ($\alpha = \omega_0$):- When $\alpha = \omega_0$,

$$C = \frac{4L}{R^2} \quad \text{and:}$$

$$s_1 = s_2 = -\alpha = -\frac{R}{2L}$$

$$i(t) = (A_1 + A_2 t) e^{-\alpha t}$$

→ The Underdamped Case ($\alpha < \omega_0$):- when $\alpha < \omega_0$, $C < \frac{4L}{R^2}$

$$\text{and } s_1 = -\alpha + \sqrt{-(\omega_0^2 - \alpha^2)} = -\alpha + j\omega_d$$

$$s_2 = -\alpha - \sqrt{-(\omega_0^2 - \alpha^2)} = -\alpha - j\omega_d$$

we find

$$i(t) = e^{-\alpha t} (A_1 \cos \omega_d t + A_2 \sin \omega_d t)$$

Example 8.3 - In fig. 8.8, $R=40\Omega$, $L=4H$, $C=\frac{1}{4}F$.

Calculate the characteristic roots of the circuit. Is the natural response overdamped, underdamped or critically damped?

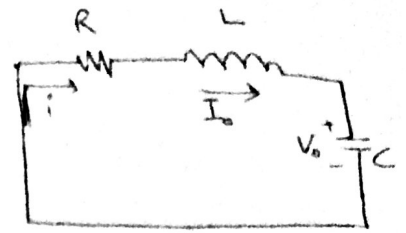


Fig. 8.8

Solu: we first calculate:

$$\alpha = \frac{R}{2L} = \frac{40}{2 \times 4} = 5$$

$$\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{4 \times \frac{1}{4}}} = 1$$

The roots are:

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2} = -5 + \sqrt{(5)^2 - (1)^2}$$

$$s_1 = -5 + \sqrt{24}$$

$$s_1 = -0.101$$

$$s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2} = -5 - \sqrt{(5)^2 - (1)^2}$$

$$s_2 = -9.899$$

From above calculations we see that $\alpha > \omega_0$ i.e. $5 > 1$. So it means response is "Overdamped".

Also from roots s_1 and s_2 are real and negative, we can decide that the circuit response is "Overdamped".

PP 8.3 - If $R=10\Omega$, $L=5H$, $C=2mF$. Find α , ω_0 , s_1 and s_2 . What type of natural response will the circuit have?

Solu: $\alpha = \frac{R}{2L} = \frac{10}{2(5)} = 1$, $\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{5(2mF)}} = 10$

The roots are:

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2} = -1 + \sqrt{(1)^2 - (10)^2}$$

$$s_1 = -1 + j9.95$$

$$s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2}$$

$$s_2 = -1 - \sqrt{(1)^2 - (10)^2}$$

$$s_2 = -1 - j9.95$$

From above calculations we see that $\alpha < \omega_0$ i.e. $1 < 10$. So it means response is "Underdamped".

Example 8.4: Find $i(t)$ in the circuit fig 8.10, Assume that circuit has reached steady state at $t=0^-$.

Solution: For $t < 0$, switch is closed. The capacitor behave as an open circuit, and inductor " " short circuit. So the equivalent circuit shown in fig 8.11(a) for $t < 0$:

Thus at $t = 0^-$

$$i(0) = \frac{V}{R_{eq}} = \frac{10V}{4+6} = 1A$$

$$V(0) = \frac{R_2}{R_1+R_2} V = \frac{6}{4+6} \times 10 = 6V$$

So now we know that $i(0)$ is the initial current through the inductor and $V(0)$ is the " voltage " capacitor.

Now when $t > 0$, the switch is open and so the volt. source is disconnected. Also the resistors 6Ω and 3Ω in fig. 8.10 are in series, so their equivalent resistance is $6\Omega + 3\Omega = 9\Omega$.

Now we calculate the roots s_1 and s_2 . we first find α and ω_0 as:

$$\alpha = \frac{R}{2L} = \frac{9}{2(0.5)} = 9, \quad \omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{(0.5)(0.02)}} = 10$$

$$s_1 = -\alpha + \sqrt{\alpha^2 - \omega_0^2} = -9 + \sqrt{81 - 100}$$

$$s_1 = -9 + j4.359$$

$$s_2 = -\alpha - \sqrt{\alpha^2 - \omega_0^2} = -9 - \sqrt{81 - 100}$$

$$s_2 = -9 - j4.359$$

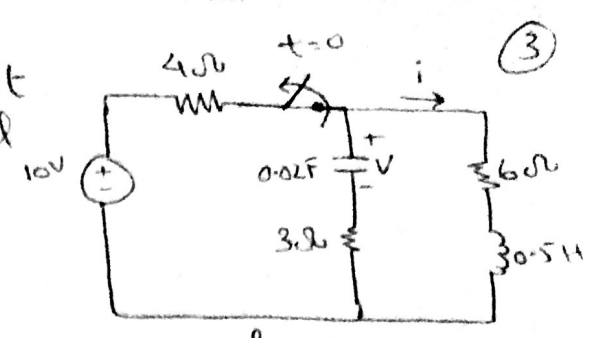


fig. 8.10

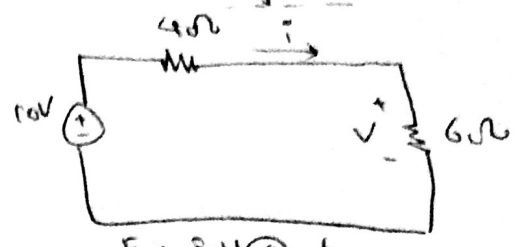


Fig 8.11(a) $t < 0$

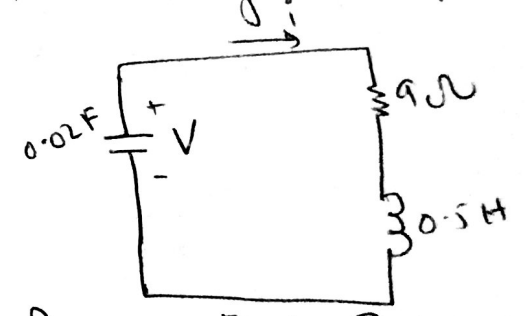


Fig 8.11(b) $t > 0$

From calculations we see that, $\alpha < \omega_0$ i.e. $9 < 10$, it means response is "underdamped". We use formulae for underdamped circuit response:

$$i(t) = A_1 e^{-\alpha t} \cos \omega_d t + B_1 e^{-\alpha t} \sin \omega_d t \quad \text{--- (A)}$$

$$\text{Here } \omega_d = \sqrt{\omega_0^2 - \alpha^2}$$

$$\omega_d = \sqrt{100 - 81}$$

$$\omega_d = 4.359$$

Substitute values in eq (A): we get

$$i(t) = e^{-9t} (A_1 \cos 4.359t + A_2 \sin 4.359t) \quad \text{--- (B)}$$

We obtain A_1 and A_2 using initial conditions, At $t=0$

$$i(0) = 1 = A_1$$

The circuit eq of fig. 8.11b is:

$$Ri(t) + L \frac{di(t)}{dt} + V_0 = 0$$

$$\text{or } \frac{di(t)}{dt} = -\frac{1}{L} (Ri(t) + V_0)$$

$$= -2(9(1) + (-6))$$

$$\left. \frac{di}{dt} \right|_{t=0} = -6 \text{ A/s}$$

$V(0) = V_0 = -6V$
because of sign convention.

Now taking derivative of $i(t)$ in eq (B):

$$\frac{di(t)}{dt} = e^{-9t} \frac{d}{dt} (A_1 \cos 4.359t + A_2 \sin 4.359t) + (A_1 \cos 4.359t + A_2 \sin 4.359t) \frac{d}{dt} e^{-9t}$$

$$= e^{-9t} [A_1 (-\sin 4.359t)(4.359) + A_2 \cos 4.359t(4.359)] + (\quad) e^{-9t} \frac{d}{dt} (-9e)$$

$$\frac{di}{dt} = (e^{-9t})(4.359)(-A_1 \sin 4.359t + A_2 \cos 4.359t) - 9e^{-9t}(A_1 \cos 4.359t + A_2 \sin 4.359t)$$

At $t=0$: we get above eq as:

$$-6 = 4.359(-0 + A_2) - 9(A_1 + 0)$$

$$-6 = -9 + 4.359A_2 \Rightarrow A_2 = 0.6882$$

∴ The complete solution is:

$$\boxed{i(t) = e^{-9t} (\cos 4.359t + 0.6882 \sin 4.359t) \text{ A}}$$