

IQRA NATIONAL UNIVERSITY



Digital Signal Processing **Sessional Assignment 2020**

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Question No (1)

Determine the response $y[n]$, $n \geq 0$ of the system described by the 2nd order difference equation:

$$\underline{y[n] - 3y[n-1] - 4y[n-2] = x[n] + 2x[n-1]}$$

to the input $x[n] = 4^n u[n]$

Solution:- $y[n] - 3y[n-1] - 4y[n-2] = x[n] + 2x[n-1]$ (n-1) \rightarrow ①

The homogenous equation of the system

$$y[n] - 3y[n-1] - 4y[n-2] = 0$$

The characteristic equation of the system

$$\lambda - 3\lambda^{-1} - 4\lambda^{-2} = 0$$

$$\lambda^2 - 3\lambda - 4 = 0$$

$$\lambda^2 - 4\lambda + \lambda - 4 = 0$$

$$\lambda(\lambda - 4) + 1(\lambda - 4) = 0$$

$$(\lambda - 4)(\lambda + 1) = 0$$

$$\lambda = 4, \lambda = -1$$

So

$$y_0[n] = C_1 [-1]^n u[n] + C_2 [4]^n u[n]$$

\Rightarrow Since 4 is a characteristic root and the excitation is

$$x[n] = 4^n u[n]$$

Assume a particular solution.

$$y_p[n] = k n 4^n u[n]$$

Then

$$k n 4^n u[n] - 3k[n-1] 4^{n-1} u[n-1] - 4k[n-2] 4^{n-2}$$

$$u[n-2] = 4^n u[n] + 2(4)^{n-1} u[n-1]$$

For $n=2$

$$k(32-12) = 4^2 + 8 = 24$$

$$k = \frac{6}{5}$$

The total solution is

$$\begin{aligned} y[n] &= y_p[n] + y_h[n] \\ &= \left[\frac{6}{5} n 4^n + (c_1 4^n + c_2 (-1)^n) \right] u[n] \end{aligned}$$

The solve for c_1 c_2 we assume

$$\text{that } y(-1) = y(-2) = 0$$

$$y(0) = 1 \quad \text{and} \quad y(1) = 3y(0) + 4 + 2 = 9$$

\Rightarrow Hence

$$c_1 + c_2 = 1$$

$$\frac{24}{5} + 4c_1 - c_2 = 9$$

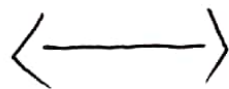
$$4c_1 - c_2 = \frac{21}{5}$$

$$c_1 = \frac{26}{25}$$

$$c_2 = \frac{-1}{25}$$

The total solution is

$$y(n) = \left[\frac{6}{5} n 4^n + \frac{26}{25} 4^n - \frac{1}{25} (-1)^n \right] u(n).$$



⇒ Part (B)

Determine the impulse response and unit step response of the system described by different equation.

$$\underline{y[n] = 0.6y[n-1] - 0.8y[n-2] + x[n]}$$

Solution:- $y[n] = 0.6y[n-1] - 0.8y[n-2] + x[n]$

$$y[n] - 0.6y[n-1] + 0.8y[n-2] - x[n]$$

⇒ homogenous eq

$$x[n] = 0$$

$$y[n] = 0.6y[n-1] + 0.8y[n-2] = 0$$

$$y^n[n] = \lambda^n$$

So $\lambda^n - 0.6\lambda^{n-1} + 0.8\lambda^{n-2} = 0$

$$\lambda^{n-2} (\lambda^2 - 0.6\lambda + 0.8) = 0$$

$$\lambda^2 - 0.6\lambda + 0.8 = 0$$

$$(\lambda - 0.2)(\lambda - 0.8) = 0$$

$$\lambda_1 = 0.2, \lambda_2 = 0.8$$

Thus the general form of the solution to the homogenous equation.

$$y^n[n] = c_1 (\lambda_1)^n + c_2 (\lambda_2)^n$$

$$y(n) = c_1 (0.2)^n + c_2 (0.4)^n \rightarrow \textcircled{1}$$

$$y^n(n) = c_1 \frac{1}{5}^n + c_2 \left(\frac{2}{5}\right)^n$$

$$y(0) = 1, \quad y(1) - 0.6y(0) = 0$$

$$y(1) = 0.6$$

$$\rightarrow c_1 = -1 \quad c_2 = 3$$

there for

$$h[n] = \left[-\left(\frac{1}{5}\right)^n + 2\left(\frac{2}{5}\right)^n \right] u(n)$$

The step response

$$y(n) = \sum_{k=0}^n h(n-k), \quad n > 0$$

$$= \sum_{k=0}^n \left[2\left(\frac{2}{5}\right)^{n-k} - \left(\frac{1}{5}\right)^{n-k} \right]$$

$$= \left(\frac{1}{0.12} \left(\frac{2^{n+1}}{5} - 1 \right) - \frac{1}{0.16} \left(\frac{1^{n+1}}{5} - 1 \right) \right) u(n)$$



Question NO (2)

=> Part (A)

Determine the casual signal $x(n]$ having
Z - transform.

$$x(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

Solution :- $x(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$

By Partial Fraction method

$$\frac{1}{(1-2z^{-1})(1-z^{-1})^2} = \frac{A}{(1-2z^{-1})} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$\frac{1}{(1-2z^{-1})(1-z^{-1})^2} = \frac{A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1})}{(1-2z^{-1})(1-z^{-1})^2}$$

$$1 = A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1})$$

eq (1)

put $z = 1$

$$1 = A(1-0)^2 + B(1-2)(1-1) + C(1)(1-2)$$

$$1 = 0 + 0 - C$$

$$1 = -C$$

$$C = -1$$

\Rightarrow put $z=2$ in eq (1)

$$1 = A \left(1 - \frac{1}{2}\right)^2 + B \left(1 - \frac{2}{2}\right) \left(1 - \frac{1}{2}\right) + C \left(\frac{1}{2}\right) \left(1 - \frac{2}{2}\right)$$

$$1 = A \left(\frac{1}{2}\right)^2 + B (1-1) \left(\frac{1}{2}\right) + C \left(\frac{1}{2}\right) (1-1)$$

$$1 = \frac{A}{4} + B(0) \left(\frac{1}{2}\right) + C \left(\frac{1}{2}\right) (0)$$

$$1 = \frac{A}{4} + 0 + 0$$

$$\boxed{A = 4}$$

put $z=3$ in eq (1)

$$1 = A \left(1 - \frac{1}{3}\right)^2 + B \left(1 - \frac{2}{3}\right) \left(1 - \frac{1}{3}\right) + C \left(\frac{1}{3}\right) \left(1 - \frac{2}{3}\right)$$

$$1 = A \left(\frac{4}{9}\right) + B \left(\frac{1}{3}\right) \left(\frac{2}{3}\right) + C \left(\frac{1}{3}\right) \left(\frac{1}{3}\right)$$

$$1 = \frac{4A}{9} + \frac{2B}{9} + \frac{1C}{9}$$

$$1 = \frac{4}{9}(4) + \frac{2B}{9} - \frac{1}{9}$$

$$1 + \frac{1}{9} - \frac{16}{9} = \frac{2B}{9}$$

$$-\frac{6}{9} \times \frac{9}{2} = B$$

$$\boxed{-3 = B}$$

Hence $x(n) = [4(2)^{n-3-n}] u(n)$.

$\langle \leftarrow \right\rangle$

Question NO (2)

=> Part (B)

Determine the Partial Fraction expansion of the following Proper Function.

$$x(z) = \frac{1}{1 - 1.5z^{-1} + 0.5z^{-2}}$$

Solution :- Eliminate the -ve power by multiplying both numerator and denominator by z^2

$$x(z) = \frac{z^2}{z^2 - 1.5z + 0.5}$$

$$\frac{x(z)}{z} = \frac{z}{(z-1)(z-0.5)} = \frac{A}{z-1} + \frac{B}{z-0.5}$$

$$z = (z-0.5)A + (z-1)B \rightarrow \textcircled{1} \quad \therefore \text{LCM}$$

Now set $z = p_1 = 1$ in eq $\textcircled{1}$ we eliminate the term involving A.

$$1 = (1 - 0.5)A$$

$$\boxed{A = 2}$$

Return to eq $\textcircled{1}$ $z = p_2 = 0.5$ then elementary the term involving A, So we have

$$0.5 = (0.5 - 1)B$$

$$\boxed{B = -1}$$

Question No (3)

=> part (A)

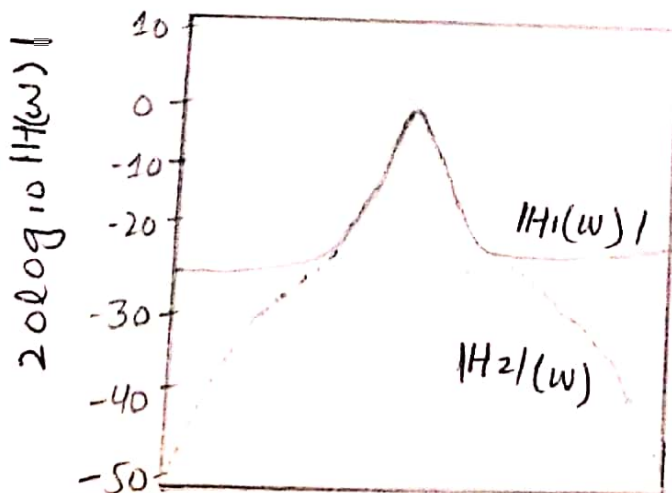
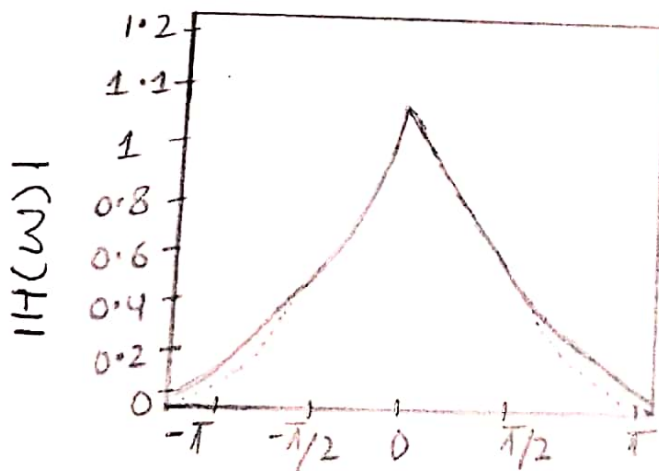
A two pole low pass filter has the system

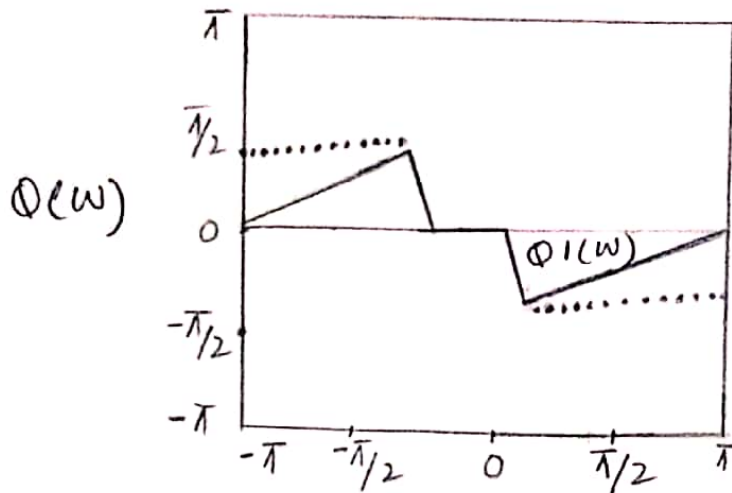
response
$$H(z) = \frac{b_0}{(1 - pz^{-1})^2}$$

=> Determine the values of b_0 and p such that the frequency response $H(\omega)$ satisfies the condition $H(0) = 1$ and

$$|H(\frac{\pi}{4})|^2 = \frac{1}{2}$$

Solution :- Linear Time Invariant System as Frequency Selective filters.





Magnitude of and phase response of (1) a single pole filter and (2) a one pole one zero filter.

$$H_1(z) = (1-a)(1-az^{-1})$$

$$H_2(z) = \frac{(1-a)}{2} \left[\frac{1+z^{-1}}{(1-az^{-1})} \right]$$

$$a = 0.9$$

Question No (3)

=> Part B

Design a two pole band pass filter that has the center of the $\omega = \pi/2$ zero in the frequency response characteristic at $\omega = 0$ and $\omega = \pi$ and its magnitude response is

$$\frac{1}{\sqrt{}} \text{ at } \omega = \frac{4\pi}{9}$$

Solution :- The filter must have poles at $P_{1,2} = re$

And zero at $z = 1$ And $z = -1$

consequently the same system function

$$H(z) = G \frac{(z-1)(z+1)}{(z-jr)(z+jr)}$$

$$H(z) = \frac{Gz^2 - 1}{z^2 + r^2}$$

=> The gain factor is determined evaluating the frequency response

$\Rightarrow H(\omega)$ of the filter at $\omega = \pi/2$

$$H = (1/2) = G \frac{2}{1-r^2} = 1$$

$$G = \frac{1-r^2}{2}$$

The value of r is determined by evaluating $H(\omega)$ at $\omega = 4\pi/9$ we have

$$\begin{aligned} |H(4\pi/9)|^2 &= \left(\frac{1-r^2}{4}\right)^2 \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)} \\ &= 1/2 \end{aligned}$$

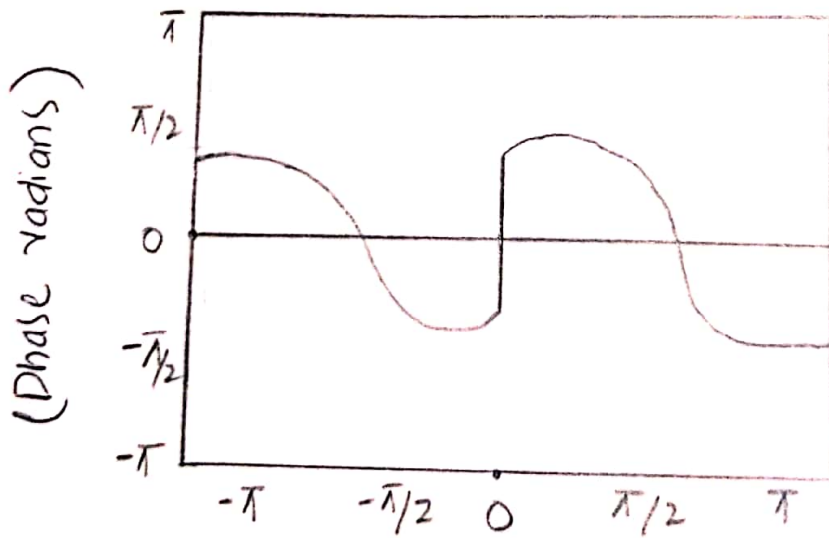
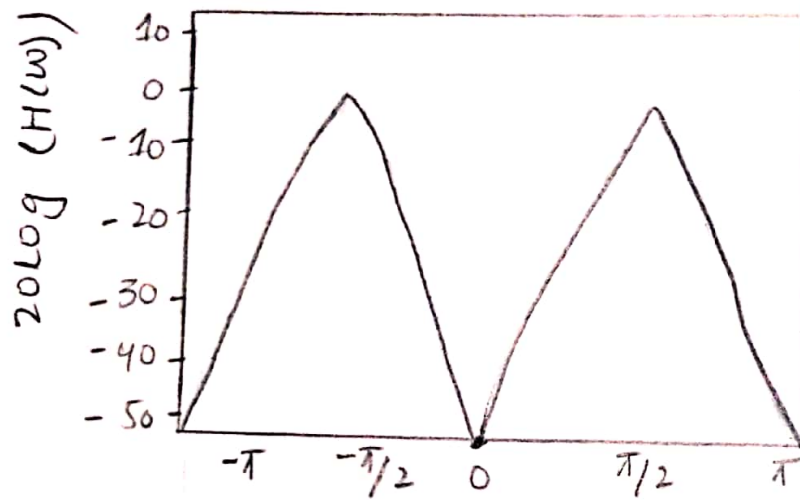
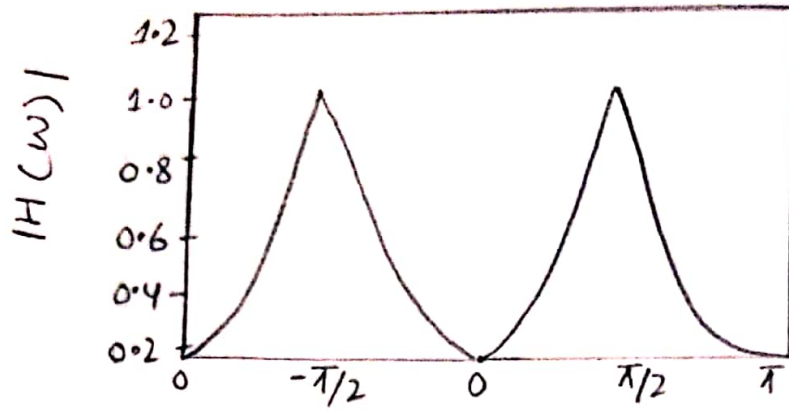
\Rightarrow Equivalently

$$1.94 (1-r^2)^2 = 1 - 1.88r^2 + r^4$$

The value of $r^2 = 0.7$ satisfies this equation there for the system function for the desired filter

$$\text{is } H(z) = 0.15 \frac{1-z^{-2}}{1+0.7z^{-2}}$$

its frequency response is illustrated.



Magnitude and phase response of a simple bandpass filter is

$$H(z) = 0.15 \left[\frac{(1-z)^{-2}}{(1+0.7z^{-2})} \right]$$



Question NO (4)

=> Part (A)

A finite duration sequence of length L is

$$\text{given as } x(n) = \begin{cases} 1, & 0 \leq n \leq L-1 \\ 0, & \text{otherwise} \end{cases}$$

Determine the N point DFT of the sequence for $N \geq L$

Solution :-

The Fourier transform of

this sequence is

$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$X(\omega) = \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}}$$

$$= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}$$

The magnitude and phase of $X(\omega)$ are illustrated for $L=40$

The N point DFT of $x(n)$ is

Simply $X(\omega)$ evaluated at the set of N equally spaced frequencies.

$\Rightarrow WK = 2\pi k/N$

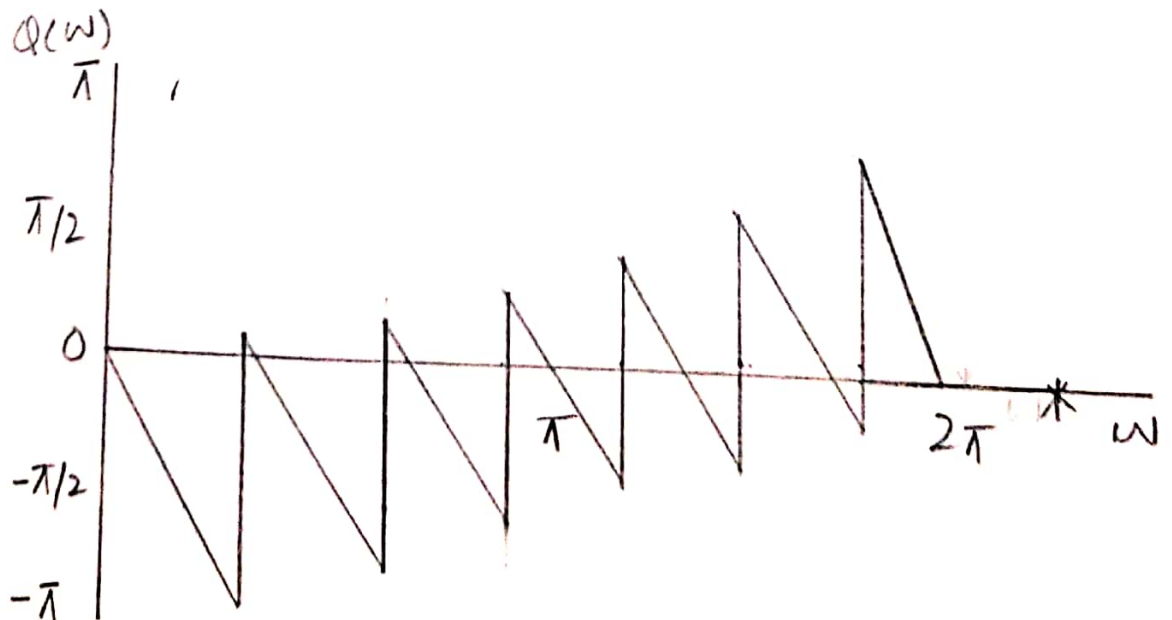
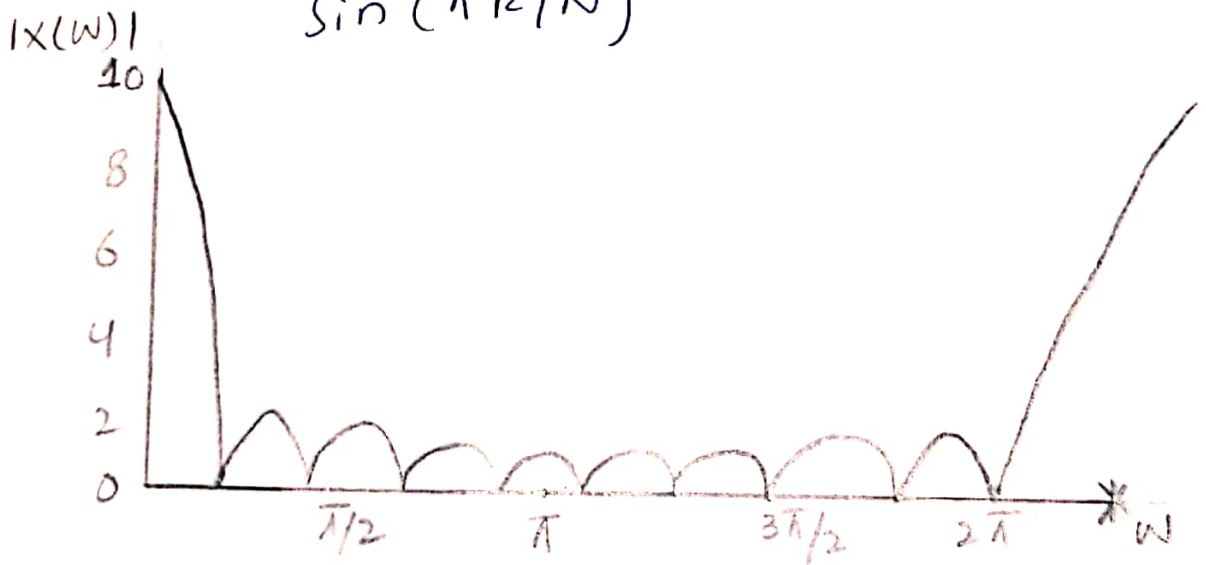
$k = 0, 1, 2, \dots, N-1$

Hence

$$X(k) = \frac{1 - e^{-j2\pi kL/N}}{1 - e^{-j2\pi k/N}}$$

$k = 0, 1, \dots, N-1$

$$X(k) = \frac{\sin(\pi kL/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$



If N is selected such that $N=L$

Then the discrete Fourier transformer becomes

$$X(k) = \begin{cases} L, & k=0 \\ 0, & k=1, 2, \dots, L-1 \end{cases}$$

Thus there is only one non zero value in the DFT this is apparent from observation of $x(\omega)$ since $x(\omega) = 0$ at the frequencies $\omega_k = 2\pi k/L$ $k \neq 0$;

The reader should verify that $x(n)$ can be recovered from $X(k)$ by

performing an L -point IDFT provides a plot of the N -point DFT in magnitude phase for $L=40$

$N=50$ and $N=100$ Now the

special characteristics of the sequence are more clearly evident.

=> Part (B)

Determine the N point DFT of this sequence

for $N \geq L$

Compute the DFT of the four point sequence

$$x(n) = (0, 1, 2, 3)$$

Solution :- The first step is to determine

the matrix W_4 by exploiting the periodicity property of W_4

$$W_N^{k+N/2} = -W_N^k$$

$$W_4 = \begin{bmatrix} W_4^0 & W_4^0 & W_4^0 & W_4^0 \\ W_4^0 & W_4^{0^2} & W_4^1 & W_4^2 \\ W_4^0 & W_4^2 & W_4^1 & W_4^1 \\ W_4^0 & W_4^3 & W_4^6 & W_4^7 \end{bmatrix}$$

$$W_4 = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^1 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^2 & W_4^2 \\ 1 & W_4^3 & W_4^2 & W_4^2 \end{bmatrix}$$

\Rightarrow Then they

$$y_4 = w_4 \lambda_4 = \begin{bmatrix} 6 \\ -2+2j \\ -2 \\ -2-2j \end{bmatrix}$$

The DFT of λ_4 may be determined by conjugating the element in w_4 to obtain w_4^* and then applying the formula.



Thank You