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Subject :- Communication System.

Q#01 part(a).

Ans:-

The SNR of an access point signal, measured at the user device, decrease as to range to the user decrease because the applicable free space loss between the user and the access point reduce signal level. ∴

The same goes for the signals propagating from the user device to the access point. An increase in RF interference from micro wave ovens and cordless phones, which increase the noise level, also decrease

SNR.

SNR directly impacts the performance of a wireless connection. A higher SNR

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value means that the signal strength is stronger in relation to the noise levels, which allows higher data rates and fewer retransmissions - all of which offers better throughput. Of course the opposite is also true. A lower SNR requires wireless device to operate at lower data rates, which decrease throughput. A SNR of 30 dB for ex. may allow an 802.11g client radio and access point to communicate at 24 Mbps; whereas, a SNR of 15 dB may only provide for 6 Mbps.

40 dB SNR = Excellent signal (5 bars); always associated; lightning fast.

25 dB to 40 dB SNR = very good signal (3-4 bars) also associated very fast.

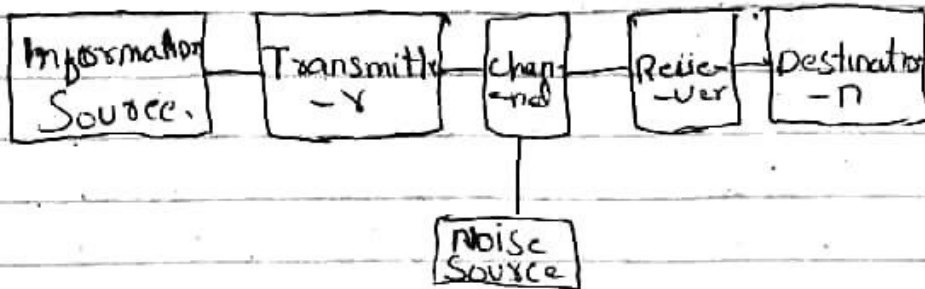
15 dB to 25 dB SNR = Low signal (2 bars); always associated; usually fast.

10 dB - 15 dB SNR = very low signal (1 bar); mostly associated mostly slow.

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5dB to 10dB SNR = No Signal;
not associated; no go.
These values seem consistent
with testing.

Part (b) :-



The block diagram of communication system, including the information source, transmitter, channel, receiver and destination blocks.

(i) Information Source :-

The information source convert this information into physical quantity.

(ii) Transmitter :-

The objective of the transmitter block is to collect the incoming message and to modify it in suitable fashion (if needed), such that, it can be transmitted via the chosen channel to the receiving point.

iii)

Channels -

Channel is the physical medium which connect the Transmitter with that of the Receiver.

iv)

Receiver -

The receiver block Receive the incoming modified version of the message signals from the channel And process it to recreate the original (non-electrical) form of the message signals.

v)

Destination -

The destination is The final block in the communication system which Receives the message signals And process it to comprehend the information present in it.

Part (C) :-

when transmission distance increase the signal source tend to loss, so carrier signals is added along with the message signal (to strengthen the original message signal). After receiving the receiver, the original signal is received.

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By filtering or removing the carrier called demodulation.
part (d):-

If you send digital data directly through the air you will probably interfere with other transmitters. So to separate the different channels the signal is modulated in a given frequency band. Obviously you can do this by digital modulation but due to harmonics you will impact other channels (modulation with a square signal has lots of harmonics). And during your demodulation depending on the other channels your signal will be distorted. Moreover you can suffer from a bandwidth problem of your power amplifier which will be distorted also your transmission.

part (e):-

This is a periodic signal with period $T_0 = 2\pi/\omega_0$. The suitable measure of its size is power. Because it is a periodic signal, we may compute its power by averaging

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Averaging its energy over one period $2\pi/\omega$. However for the sake of generality we shall solve this problem by averaging over infinitely large time interval using Eq:

$$P_g = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} c^2 \cos^2(\omega t + \theta) dt =$$
$$= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} \frac{c^2}{2} (1 + \cos(2\omega t + 2\theta)) dt$$
$$= \lim_{T \rightarrow \infty} \frac{c^2}{2T} \int_{-T/2}^{T/2} dt + \lim_{T \rightarrow \infty} \frac{c^2}{2T} \int_{-T/2}^{T/2} \cos(2\omega t + 2\theta) dt$$

The first term in the right hand side equals $c^2/2$ while the second term is zero because of the integral appearing in this term presents the area under a sinusoid over a very large time interval T with $T \rightarrow \infty$.

This area is at most equal to the area of half the cycle because of cancellation of the positive and negative areas of a sinusoid. The second term is the area multiplied by $c^2/2T$ with $T \rightarrow \infty$ clearly this term is zero and

$$P_g = \frac{c^2}{2}$$

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This shows a well known fact that a sinusoid of amplitude C has a power $C^2/2$ regardless of its angular frequency ($\omega = 0$) and its phase θ . The rms value is $C/\sqrt{2}$. If the signal frequency is zero (dc or a constant signal of amplitude C), the reader can show that the power is C^2 .

Q #02: part (A)

Amplitude modulation-3

- AM Spectrum
- power and bandwidth of a signal
- multiple tone AM

$$x_m(t) \rightarrow x_c(t) \rightarrow x_{AM}(t)$$

$$\downarrow \quad \downarrow \quad \downarrow$$

$$A_2 \cos \omega_m t \quad A_c \cos \omega_c t \quad A_c [1 + m \cos \omega_m t]$$

$$x_{AM}(t) \leftarrow x_m(t) \rightarrow x_c(t) \xrightarrow{E} x_c(t) \cdot x_{AM}(t) \rightarrow x(t)$$

$$x_{PM}(t) = A_c \cos \omega_c t + x_m(t) \cos \omega_c t$$

$$x_1(t) \quad x_2(t)$$

$$x_{AM}(t) = x_1(t) + x_2(t)$$

$$\rightarrow x_m(t) \xrightarrow{MFT} X(\omega_c - \omega_m)$$

$$x_{AM}(t) = A_c \cos \omega_c t + x_m(t) \cos \omega_c t$$

$$= 1 + \dots + \dots + \dots + \dots \rightarrow Net$$

$$\frac{A_c}{2} \left(\frac{e^{j\omega_c t} + e^{-j\omega_c t}}{2} + \frac{x_m(t)}{2} (e^{j\omega_c t} + e^{-j\omega_c t}) \right)$$

Embedding
 $\cos \omega_c t \cdot \frac{1}{2} (e^{j\omega_m t} + e^{-j\omega_m t})$

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$$= \frac{1}{2} \cos(\omega_c t) e^{j\omega_c t} + \frac{1}{2} \cos(\omega_m t) e^{-j\omega_c t}$$

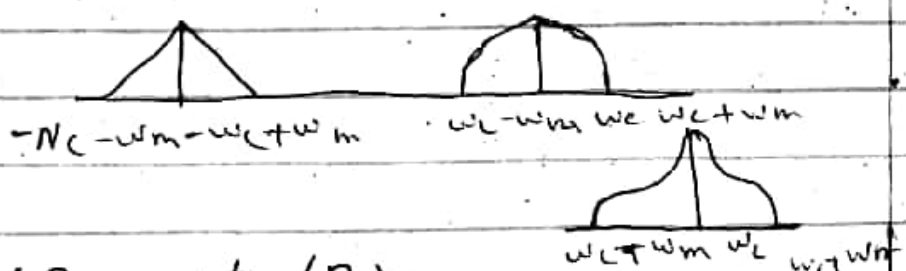
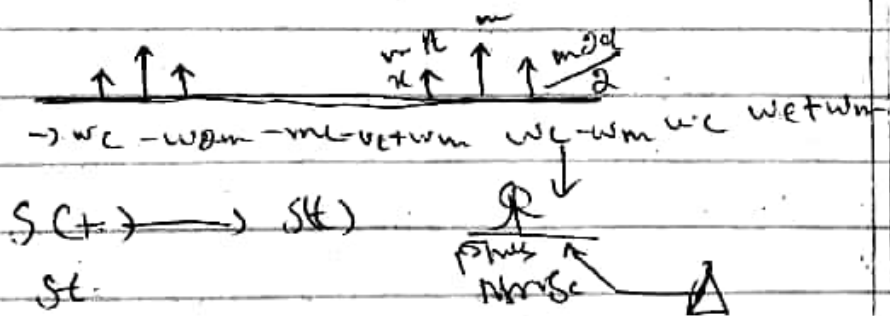
$$= \frac{1}{2} X(\omega_c - \omega_m) + \frac{1}{2} X(\omega_c + \omega_m)$$

$$X_i(t) = \mathcal{F}^{-1} \{ \pi A (\delta(\omega - \omega_c) + \delta(\omega + \omega_c)) \}$$

$$X_{AM}(t) = \mathcal{F}^{-1} \{ \pi A (\delta(\omega - \omega_c) + \delta(\omega + \omega_c)) +$$

$$\frac{1}{2} (X(\omega_c - \omega_m) + X(\omega_c + \omega_m)) \}$$

$\omega_c, \omega_c + \omega_m, \omega_c - \omega_m$



Q#02 part (A):-

Ans $S \cos 2\pi \times 10^6$
 $h = \frac{d}{4} = \frac{c}{4f}$

$S = 20 \text{ km}$
 $f = 10^6$

put the values
 $h = \frac{3 \times 10^8}{4 \times 10^6}$

$h = 75 \text{ meter}$

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$$3 \cos \pi 10^3 t$$

$$h = \frac{c}{4f}$$

$$f = 10^3 \Rightarrow 4 = \frac{c}{4f}$$

$$h = \frac{3 \times 10^5}{4 \times 10^3}$$

$$h = \frac{3 \times 10^5}{4}$$

$$h = 75000 \text{ meters}$$

Question # 03 part (B)

Ans

Given that

$$(E_c) = 7v$$

$$f_c = 1 \text{ MHz}$$

$$B_m = 5v \text{ and } f_m = 5 \text{ KHz}$$

$$\text{D modulation Index} = \frac{E_m}{E_c} = \frac{5v}{7}$$

$$M = 0.50$$

ii) Equation for modulated wave is:

$$s(t) = E_c (1 + m \cos \omega_m t) \cos \omega_c t$$

$$s(t) = 7 [1 + 0.5 \cos \omega_m t] (5 \cos \omega_c t)$$

$$s(t) (2\pi \times 1 \times 10^6 t)$$

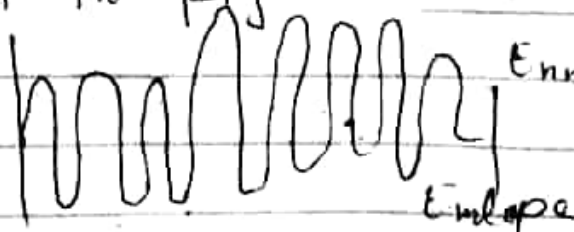
$$s(t) = 10 [4.03 \cos (10\pi \times 10^3 t)$$

$$\cos (2\pi \times 10^6 t)]$$

iii) The modulus wave form has shown in Fig.

$$E_{m\max} = 10.5$$

$$E_{m\min} = 7v$$



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ii) Spectrum of modulated
 $f_{USB} = f_c + f_m = 10^6 + 5 \times 10^3$

$$= 1000 \times 10^3 + 5 \times 10^3$$

$$= 1000 \times 10^3 + 5 \times 10^3$$

1000 kHz

$$f_{LSB} = f_c - f_m = 1000 \text{ kHz}$$

Am of each same signal

$$\frac{m}{2} \times I_r C$$

$$= \frac{0.5}{2} \times 7 = \boxed{1.75 \text{ V}}$$

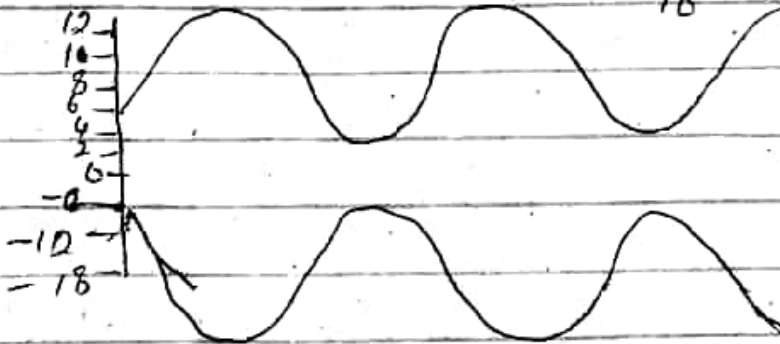
Q# 03 part (A)

$$e(t) = 2 \sin \omega t$$

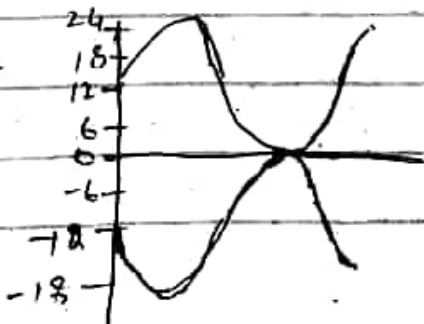
$$e_m(t) = ?$$

(i) $A_m = 6$, $A_c = 12$ ($A_c > A_m$)

$m < 100\%$



ii) $A_m = 12$, $A_c = 12$ ($A_c = A_m$)



$$m = 100\%$$

