

Department of Electrical Engineering

Final Exam Assignment

Date: 27/06/2020

Course Details

Course Title: Digital Signal Processing
Instructor: Pir Meher Ali Shah

Module: 6th
Total Marks: 50

Student Details

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Q1.	(a)	Determine the response $y(n)$, $n \geq 0$, of the system described by the second order difference equation $y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$ To the input $x(n) = (-1)^n u(n)$. And the initial conditions are $y(-1) = y(-2) = 0$.	Marks 7
			CLO 2
Q1.	(b)	Determine the impulse response and unit step response of the systems described by the difference equation. $y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$	Marks 7
			CLO 2
Q2.	(a)	Determine the causal signal $x(n)$ having the z-transform $X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$ (Hint: Take inverse z-transform using partial fraction method)	Marks 6
			CLO 2
Q3.	(b)	Evaluate the inverse z- transform using the complex inversion integral $X(z) = \frac{1}{1-az^{-1}} \quad z > a $	Marks 6
			CLO 2
Q.3	(a)	A two- pole low pass filter has the system response $H(z) = \frac{b_0}{(1-pz^{-1})^2}$ Determine the values of b_0 and p such that the frequency response $H(\omega)$ satisfies the condition $H(0) = 1$ and $\left H\left(\frac{\pi}{4}\right)\right ^2 = \frac{1}{2}$.	Marks 6
			CLO 3

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Q: 1 (a)

Ans.

Sol: $y(n) - 4y(n-1) + 4y(n-2)$
 $= n(n) - n(n-1)$

The characteristics eqn is

$$\lambda^2 - 4\lambda + 4 = 0$$
$$\lambda = 2, 2$$

Hence

$$y_h(n) = C_1 2^n + C_2 n 2^n$$

Particular Solution is

$$y_p(n) = K(-1)^n u(n)$$

Substituting this solution into the difference equation we obtained:

$$K(-1)^n u(n) - 4K(-1)^{n-1} u(n-1) + 4K(-1)^{n-2} u(n-2)$$
$$= (-1)^n u(n) - (-1)^{n-1} u(n-1)$$

for

$$n=2, K(1+4+4) = 2$$

$$K = \frac{2}{9} \text{ solution is}$$

$$y(n) = \left[C_1 2^n + C_2 n 2^n + \frac{2}{9} (-1)^n \right] u(n) \quad \text{--- (i)}$$

From the initial conditions,
we obtain $y(0) = 1$ $y(1) = 2$ then

$$C_1 + \frac{2}{9} = 1$$

$$C_1 = \frac{7}{9}$$

$$2C_1 + 2C_2 - \frac{2}{9} = 2$$

$$C_2 = \frac{1}{3}$$

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So put in eqn (i)

$$y(n) = \left[\frac{7}{9} 2^n + \frac{1}{3} n 2^n + \frac{2}{9} (-1)^n \right] u(n)$$

Q.1 (b)

Ans:

Solutions:

$$y(n) - 0.7y(n-1) + 0.1y(n-2) = 2n(n) - 2(n-2)$$

The characteristic eqn is
 $d^2 - 0.7d + 0.1 = 0$

$d = 1/2, 1/5$ Hence

$$y_h(n) = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{1}{5}\right)^n \quad \text{--- (i)}$$

with $x(n) = \delta_n$ we have

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0$$

$$y(1) = 1.4$$

$$\text{Hence } C_1 + C_2 = 2 \quad \text{--- (2)}$$

$$\frac{1}{2} C_1 + \frac{1}{5} C_2 = \frac{7}{5} = 1.4$$

$$C_1 + \frac{2}{5} C_2 = \frac{14}{5}$$

So equation is

$$C_1 = \frac{10}{3}$$

$$C_2 = -\frac{4}{3}$$

So put C_1 & C_2 in
eqn (i)

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$$h(n) = \left[\frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

The step response is

$$S(n) = \sum_{k=0}^n h(n-k)$$

$$= \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$\left. \begin{aligned} &= \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{4}{3} \\ &\left(\frac{1}{5}\right)^n (5^{n+1} - 1) u(n) \end{aligned} \right\}$$

Required Ans

Q.2 9

Ans

Solution:

$$x(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

By partial fraction method

$$\frac{1}{(1-2z^{-1})(1-z^{-1})} = \frac{A}{(1-2z^{-1})} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$1 = \frac{A(1-z^{-1})^2 + B(1-2z^{-1}) + Cz^{-1}(1-2z^{-1})}{(1-2z^{-1})(1-z^{-1})^2}$$

$$1 = \frac{A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1})}{(1-2z^{-1})(1-z^{-1})^2} \quad \text{--- (i)}$$

Put $z=1$

$$1 = A(1-1)^2 + B(1-2(1))(1-1) + C(1)(1-2(1))$$

$$1 = A(0) + B(-1)(0) + C(1)(-1)$$

$$1 = 0 + 0 - C$$

$$1 = -C$$

$$C = -1$$

Put $z=2$ in eqn (i)

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$$1 = A \left(1 - \frac{1}{2}\right)^2 + B \left(1 - \frac{2}{2}\right) \left(1 - \frac{1}{2}\right) + C \left(\frac{1}{2}\right) \left(1 - \frac{2}{2}\right)$$

$$1 = A \left(\frac{1}{2}\right)^2 + B(0) \left(\frac{1}{2}\right) + C \left(\frac{1}{2}\right) (1-1)$$

$$\frac{1}{4} = \frac{A}{4} + B(0) + C(0)$$

$$1 = \frac{A}{4} + 0 + 0$$

$$A = 4$$

Put $z = \frac{2}{3}$ in eqn (i)

$$1 = A \left(1 - \frac{1}{3}\right)^2 + B \left(1 - \frac{2}{3}\right) \left(1 - \frac{1}{3}\right) + C \left(\frac{1}{3}\right) \left(1 - \frac{2}{3}\right)$$

$$1 = A \left(\frac{4}{9}\right) + B \left(\frac{1}{3}\right) \left(\frac{2}{3}\right) + C \left(\frac{1}{3}\right) \left(\frac{1}{3}\right)$$

$$1 = \frac{4A}{9} + \frac{2}{9}B + \frac{1}{9}C$$

$$1 + \frac{1}{9} - \frac{16}{9} = \frac{2}{9}B$$

Taking L.C.M

$$\frac{9+1-16}{9} = \frac{2}{9}B$$

$$\frac{-6}{9} \times \frac{9}{2} = B$$

$$B = -3$$

Hence, $n[n] = 4(2)^n - 3 - n] u(n)$

Q: 2 (b)

Ans:

Solution: $x(z) = \frac{1}{1-az^{-1}} \quad (|z| > |a|)$

we have

$$x(n) = \frac{1}{2\pi j} \oint_C \frac{z^{n-1}}{1-az^{-1}} dz$$

$$= \frac{1}{2\pi j} \oint_C \frac{z^n dz}{z-a}$$

Where C is a circle at radius greater than $|a|$. we should evaluate this integral using $f(z) = z^n$.

(1) if $n < 0$, $f(z) = z^n$ has an n th-order pole at $z=0$ which is also inside C . thus there are contributions from both poles for $n = -1$ we have.

$$\begin{aligned} x(-1) &= \frac{1}{2\pi j} \oint_C \frac{1}{z(z-a)} dz \\ &= \frac{1}{z-a} \Big|_{z=0} + \frac{1}{z} \Big|_{z=a} = 0 \end{aligned}$$

(2) if $n \geq 0$, $f(z)$ has only zeros and hence no poles inside C . the only pole inside C is $z=a$ Hence

$$x(n) = f(z_0) = a^n \quad n \geq 0$$

if $n = -2$ we have

$$h(-2) = \frac{1}{2\pi i} \oint_C \frac{1}{z^2(z-a)} dz$$

$$= \frac{d}{dz} \left(\frac{1}{z-a} \right) \Big|_{z=0} + \frac{1}{z^2} \Big|_{z=a} = 0$$

By continuing in the same way we can show that $h(n) = 0$ if $n < 0$, thus

$$h(n) = a^n u(n)$$

Q.3 (c)

Ans: Solution: At $w=0$ we have

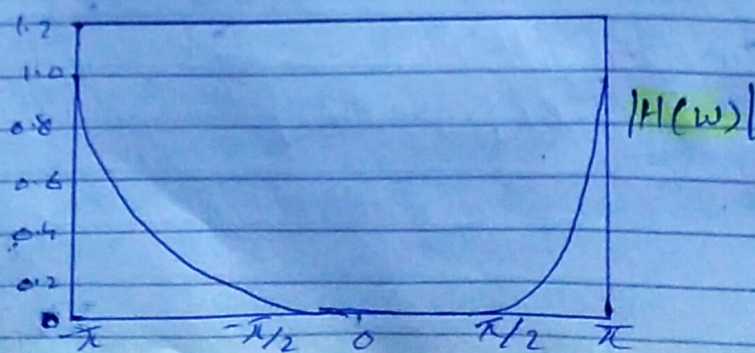
$$H(z) = \frac{b_0}{(1-pz^{-1})^2}$$

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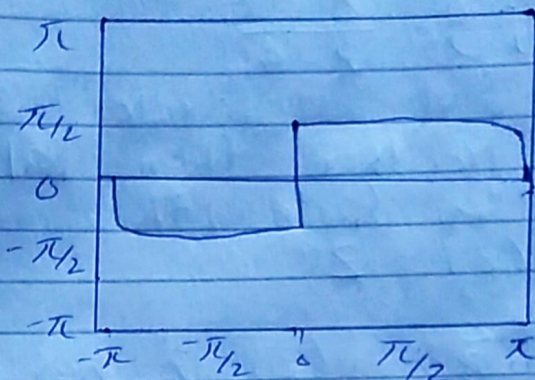
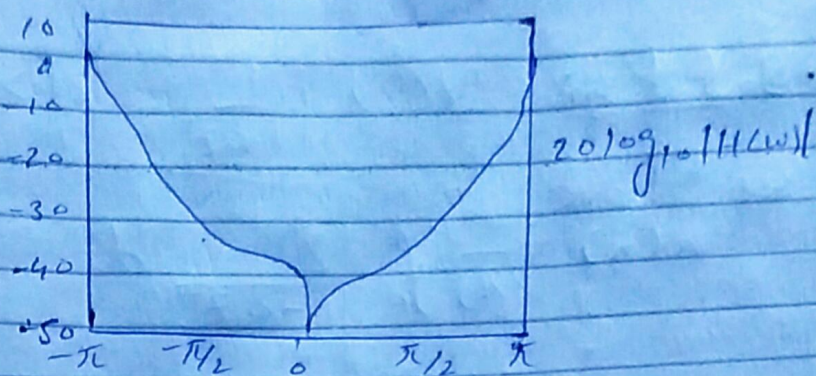
$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

Hence

$$b_0 = (1-p)^2$$



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At $\omega = \pi/4$.

$$\begin{aligned}
 H(\pi/4) &= \frac{(1-p)^2}{(1-pe^{-j\pi/4})^2} \\
 &= \frac{(1-p)^2}{(1-p(\cos(\pi/4) + j\sin(\pi/4)))^2} \\
 &= \frac{(1-p)^2}{(1-p/\sqrt{2} + jp/\sqrt{2})^2}
 \end{aligned}$$

Hence $\frac{(1-p)^2}{[(1-p/\sqrt{2})^2 + p^2/2]} = \frac{1}{2}$

Required Answer.

Q.3 (B)

Ans:

Solution:

$$P(z) = r e^{j4\pi/2}$$

∴ zeros at $z=1$ & $z=-1$

consequently, the system function is

$$H(z) = G \frac{(z-1)(z+1)}{(z-jr)(z+jr)}$$

$$= G \frac{z^2 - 1}{z^2 + r^2}$$

The gain factor is determined by evaluating the frequency response $H(\omega)$ of the filter at $\omega = \pi/2$, thus we have

$$H(\pi/2) = G \frac{2}{1-r^2} = 1$$

$$G = \frac{1-r^2}{2}$$

The value of r is determined by evaluating $H(\omega)$ at $\omega = 4\pi/9$. thus we have

$$\begin{aligned} |H(\frac{4\pi}{9})|^2 &= \frac{(1-r^2)^2}{4} \frac{2-2\cos(8\pi/9)}{1+r^2+2r^2\cos(8\pi/9)} \\ &= \frac{1}{2} \end{aligned}$$

or equivalently.

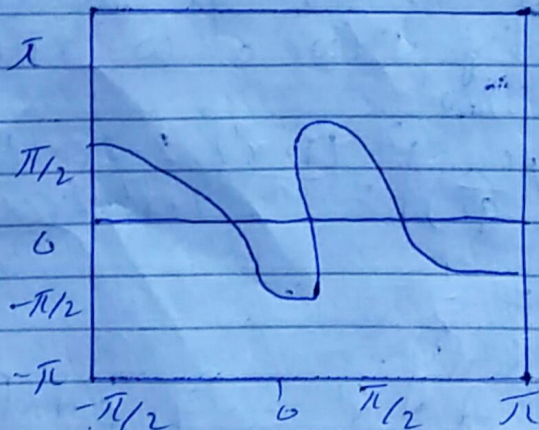
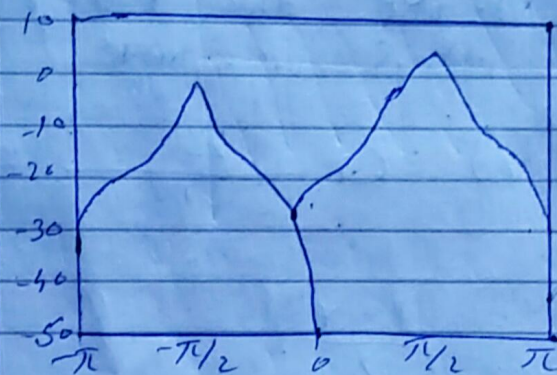
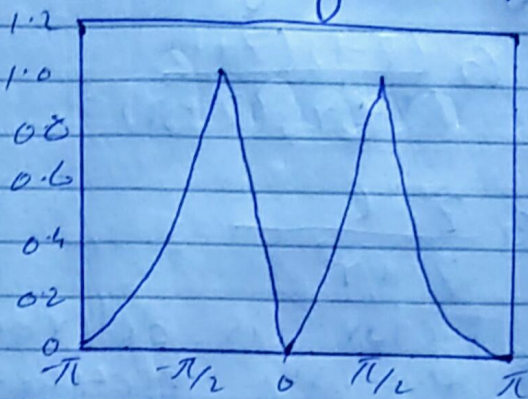
The value of $r^2 = 0.7$ satisfies this equation. therefore the system function

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For the desired filter is

$$H(z) = \frac{0.15(1 - z^{-2})}{1 + 0.7z^{-2}}$$

its frequency response is illustrated



Q.4 (a)

Ans: The Fourier transform of this sequence is

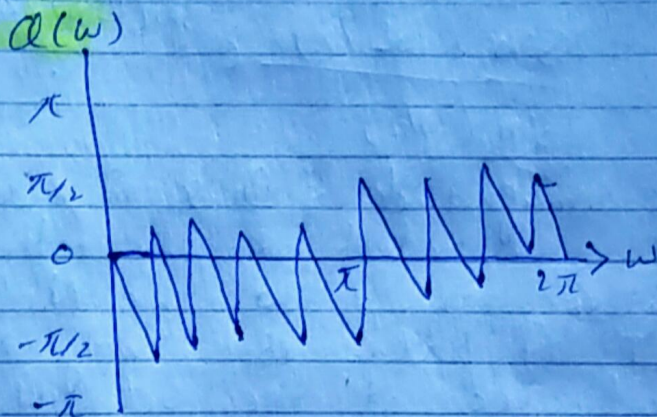
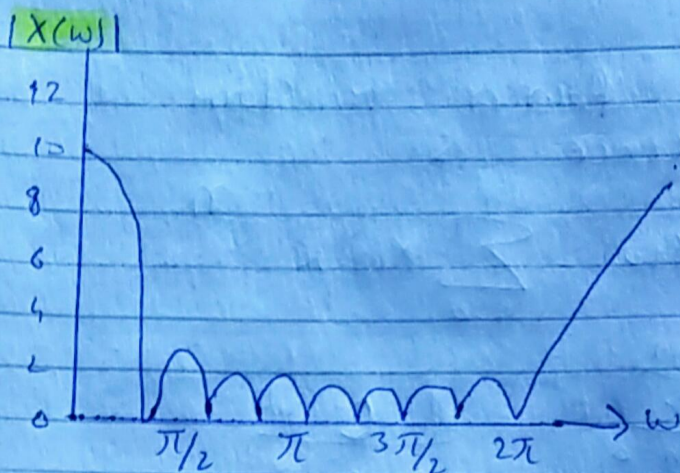
$$\begin{aligned}
 X(\omega) &= \sum_{n=0}^{L-1} x(n) e^{-j\omega n} \\
 &= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} \\
 &= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}
 \end{aligned}$$

The magnitude & Phase of $x(\omega)$ are illustrated in the following figure for $L=10$. The N -Point DFT of $x(n)$ is simply $X(\omega)$ evaluated at the set of N equally spaced frequencies
 $\omega_k = 2\pi k/N$
 $k=0, 1, \dots, N-1$ Hence.

$$X(k) = \frac{1 - e^{-j2\pi kL/N}}{1 - e^{-j2\pi k/N}}$$

$k=0, 1, \dots, N-1$

$$= \frac{\sin(\pi kL/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$



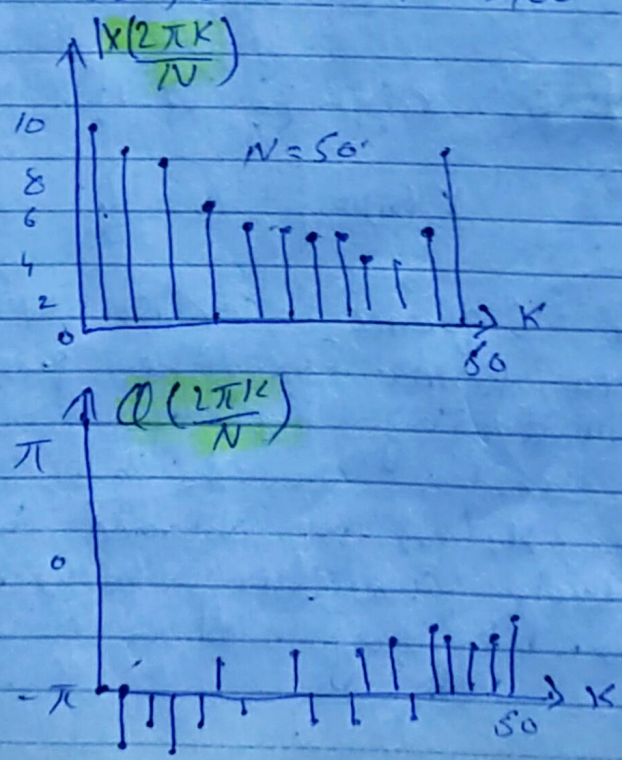
if N is selected such that $N=L$ then the DFT becomes

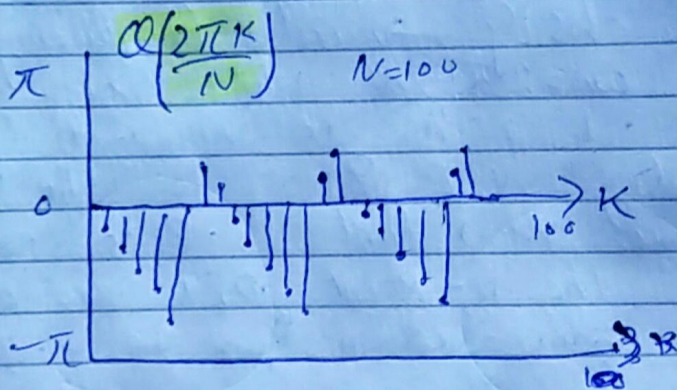
$$X(K) = \begin{cases} L, & K=0 \\ 0, & K=1, 2, \dots, L-1 \end{cases}$$

thus there is only one nonzero value in the DFT. This is apparent from observation of $X(\omega)$. Since $X(\omega) = 0$ at the frequencies $\omega_k = 2\pi k/L$, $k \neq 0$ the reader should verify that $x(n)$ can be recovered from $X(K)$ by performing an L -point IDFT.

Although the L -Point DFT is sufficient to uniquely represent the sequence $x(n)$ in the frequency domain, it is apparent that it does not provide sufficient detail to yield a good picture of the spectral characteristics of $x(n)$ if we wish to have a better pic we must evaluate (interpolate) $X(\omega)$ at more closely spaced frequency say $\omega_k = 2\pi k/N$ where $N > L$ in effect L points to N point by appending $N-L$ zeros to the sequence $x(n)$ that is zero padding then the N -point DFT provides finer interpolation than the L -point DFT.

$N = 50, L = 10 \quad \& \quad N = 100$





Q:4 (b)

Ans:

Solutions $x_1(n) = \{2, 1, 2, 1\}$

$x_2(n) = \{1, 2, 3, 4\}$

Each sequence consist of four nonzero points. For the purposes of illustrating the operations involved in circular convolution.

Now $x_3(m)$ is obtained by circularly convolving $x_1(n)$ with $x_2(n)$. Beginning with $m=0$ we have

$$N_3(0) = \sum_{n=0}^3 n_1(n) n_2(-n)_4$$

$n_2(-n)_4$ is simply the sequence $n_2(n)$ folded & graphed on a circle as illustrated in the graph.

$$N_3(0) = 14$$

For $m=1$ we have

$$N_3(1) = \sum_{n=0}^3 n_1(n) n_2((1-n)_4)$$

it is easily verified that $n_2((1-n)_4)$ is simply the sequence $n_2(-n)_4$ rotated counter-clockwise.

$$N_3(1) = 16$$

For $m=2$ we have

$$N_3(2) = \sum_{n=0}^3 n_1(n) n_2(2-n)_4$$

Now $n_2(2-n)_4$ is the folded sequence.

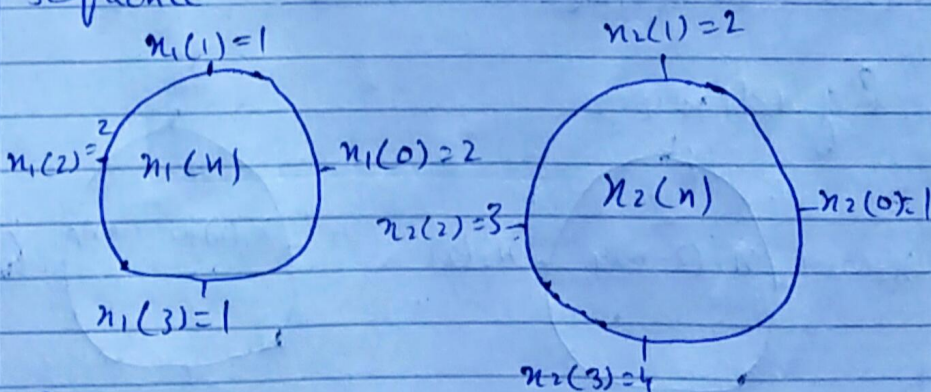


Fig (a)

Fig (b)

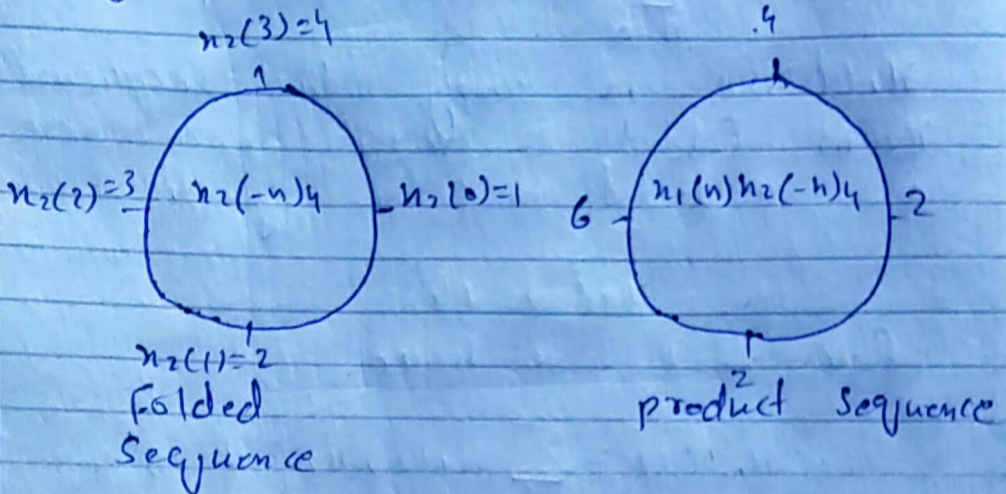


Fig (c)

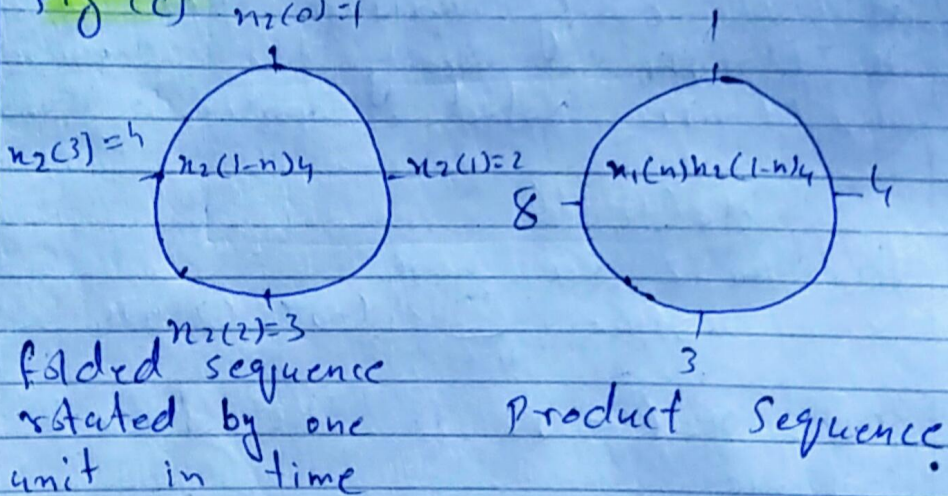
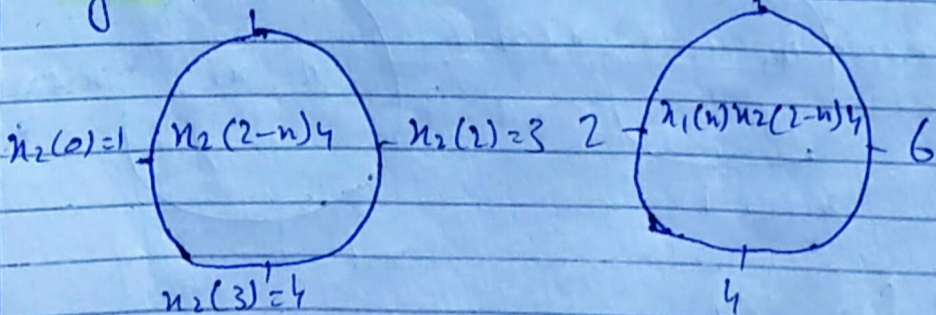


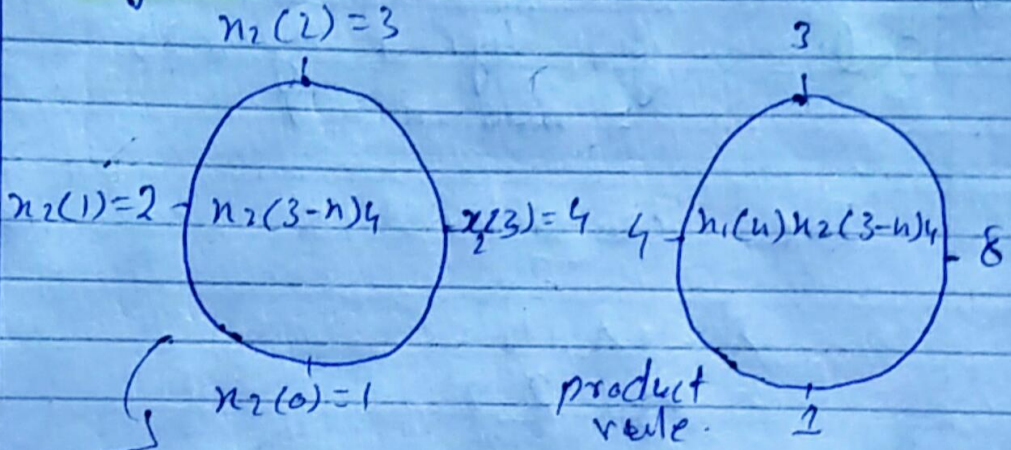
Fig (d)



Folded sequence rotated by two units in time

Product Sequence

Fig (c)



folded sequence rotated by three units in time