

IQRA NATIONAL UNIVERSITY



Digital Signal Processing **Final Assignment Spring 2020**

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Question No (1)⇒ Part (A)

Determine the response $y(n)$, $n \geq 0$, of the system describe by the second order difference equation.

$$y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$$

To the input $x(n) = (-1)^n u(n)$

And the initial condition are

$$y(-1) = y(-2) = 0 ?$$

Solution:- $\lambda^2 - 4\lambda + 4 = 0$

$$\lambda = 2, 2 \quad \text{Hence}$$

$$y_h(n) = C_1 2^n + C_2 n 2^n$$

The particular solution is

$$y_p(n) = K (-1)^n u(n)$$

Substituting this solution into the difference equation we obtain

$$\begin{aligned} \Rightarrow K (-1)^n u(n) - 4K (-1)^{n-1} u(n-1) + 4K (-1)^{n-2} \\ u(n-2) = \\ (-1)^n u(n) - (-1)^{n-1} u(n-1) \end{aligned}$$

$$\Rightarrow \text{For } n=2, \quad K(1+4+4) = 2$$

$K = 2/9$ The total solution is

$$y(n) = \left[C_1 2^n + C_2 n 2^n + \frac{2}{9} (-1)^n \right] u(n)$$

From the initial conditions

$$\text{we obtain, } y(0) = 1, \quad y(1) = 2$$

Then

$$C_1 + \frac{2}{9} = 1$$

$$\Rightarrow C_1 = 7/9$$

$$\Rightarrow 2C_1 + 2C_2 - \frac{2}{9} = 2$$

$$\Rightarrow C_2 = \frac{1}{3}$$



Question (1)⇒ Part (B)

Determine the impulse response and unit step response of systems described by the difference equation

$$y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$$

Solution:- The characteristic equation is

$$\lambda^2 - 0.7\lambda + 0.1 = 0$$

$$\lambda = \frac{1}{2}, \frac{1}{5} \quad \text{Hence}$$

$$y_h(n) = c_1 \left(\frac{1}{2}\right)^n + c_2 \left(\frac{1}{5}\right)^n$$

with $x(n) = \delta(n)$ we have

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0$$

$$\Rightarrow y(1) = 1.4$$

$$\text{Hence, } c_1 + c_2 = 2$$

And,

$$\frac{1}{2}c_1 + \frac{1}{5}c_2 = 1.4$$

$$1.4 = \frac{7}{5}$$

$$\Rightarrow c_1 + \frac{2}{5} c_2 = 14/5$$

These equations yield

$$c_1 = \frac{10}{3}, \quad c_2 = -4/3$$

$$h(n) = \left[\frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

The step response is

$$S(n) = \sum_{k=0}^n h(n-k)$$

$$S(n) = \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$S(n) = \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$S(n) = \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{1}{3}$$

$$\left(\frac{1}{5}\right)^n (5^{n+1} - 1) u(n).$$



Question (2)⇒ Part (A)

Determine the causal signal $x(n]$ having the Z-transform

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

Solution:- Taking inverse and Z-transform

$$\frac{A}{(1-2z^{-1})} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$A=4, \quad B=-3, \quad C=-1$$

Hence

$$x(n) = [4(2)^n - 3 - n] u(n)$$



Question NO (2)⇒ Part (B)

Evaluate the inverse Z transform using the inverse complex inversion integral

$$x(z) = \frac{1}{1 - az^{-1}} \quad |z| > |a|$$

Solution:- using the complex inversion integral.

we have

$$x(n) = \frac{1}{2\pi j} \oint_C \frac{z^{n-1}}{1 - az^{-1}} dz = \frac{1}{2\pi j} \oint_C \frac{z^n dz}{z - a}$$

where C is circle at radius greater than $|a|$ we shall evaluate this integral using (3,4,2) with $f(z) = z^n$

we distinguish two cases:

1 - If $n \geq 0$, $f(z)$ has radius only zeros and hence no poles inside C . The only pole inside C is

$$z = a$$

Hence $x(n) = f(z_0) = a^n \quad n \geq 0$

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\Rightarrow 2 - if $n < 0$ $f(z) = z^n$ has an
nth-order

pole at $z=0$, which is also
inside C . Thus there are contributions
from both poles. For $n = -1$ we have

$$x(-1) = \frac{1}{2\pi j} \oint_C \frac{1}{z^2(z-a)} dz = \frac{1}{z-a} \Big|_{z=0} + \frac{1}{z} \Big|_{z=ma} = 0$$

if $n = -2$, we have

$$x(-2) = \frac{1}{2\pi j} \oint_C \frac{1}{z^2(z-a)} dz = \frac{d}{dz} \left(\frac{1}{z-a} \right) \Big|_{z=0} + \frac{1}{z^2} \Big|_{z=ma} = 0$$

By continuing in same way we can

show that $x(n) = 0$

for $n < 0$, thus

$$x(n) = a^n u(n)$$



Question No (3)⇒ Part (A)

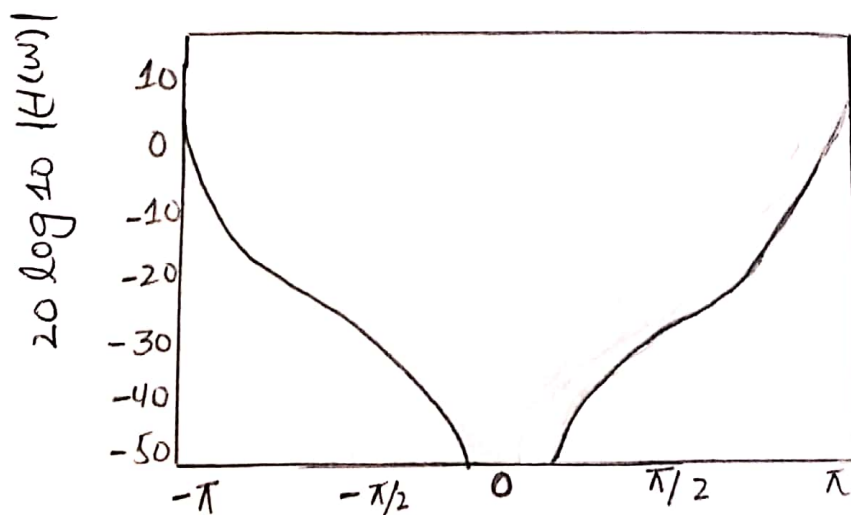
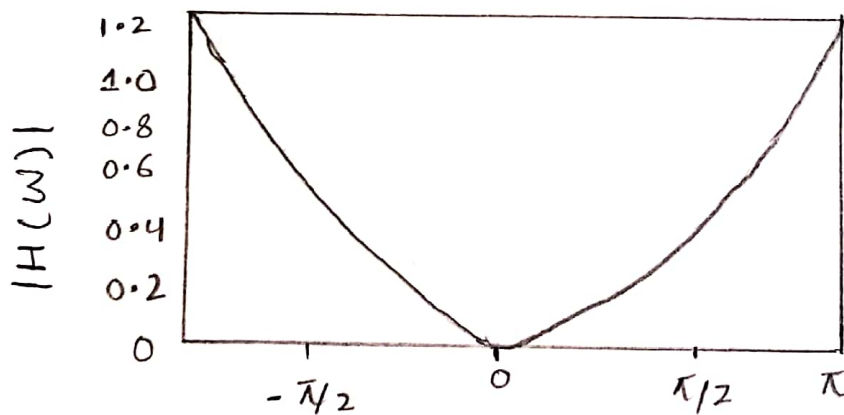
$$H(z) = \frac{b_0}{(1 - pz^{-1})^2}$$

Determine b_0 and p such that $H(w)$ satisfies the condition $H(0) = \frac{1}{2}$.

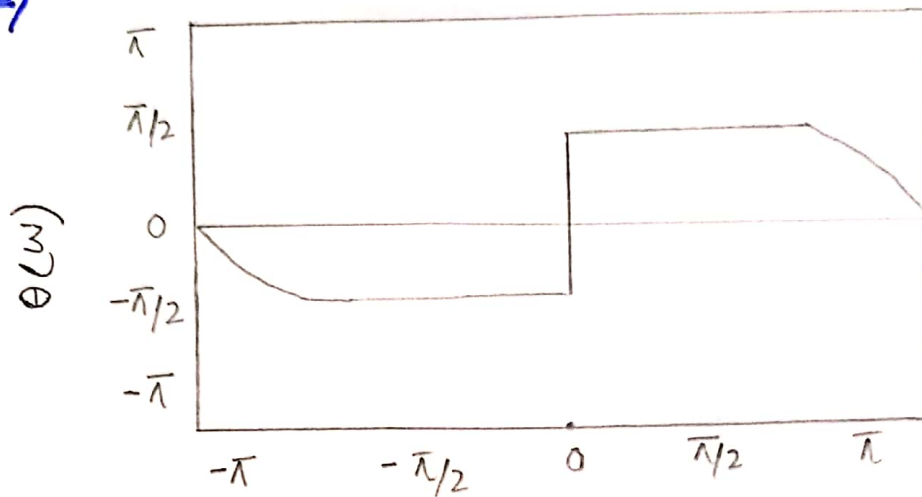
Solution:- At $w=0$ we have

$$H(0) = \frac{b_0}{(1-p)^2} = \frac{1}{2}$$

$$\text{Hence } b_0 = (1-p)^2$$



→



$$\text{At } \omega = \pi/4$$

$$H(\pi/4) = \frac{(1-p)^2}{(1 - p e^{-j\pi/4})^2}$$

$$= \frac{(1-p)^2}{[1 - p \cos(\pi/4) + j p \sin(\pi/4)]^2}$$

$$= \frac{(1-p)^2}{[1 - p/\sqrt{2} + j p/\sqrt{2}]^2}$$

$$= \frac{(1-p)^4}{[(1 - p/\sqrt{2})^2 + p^2/2]^2}$$

Hence

$$= \frac{(1-p)^4}{[(1 - p/\sqrt{2})^2 + p^2/2]^2}$$

$$= 1/2$$

⇒ Equivalently

$$\sqrt{2} (1-p)^2 = 1 + p^2 - \sqrt{2}p$$

The system function for desired filter

$$H(z) = \frac{0.46}{(1 - 0.32z^{-1})^2}$$

Question No (3)

⇒ part (B)

Design a two pole bandpass filter that has the center of the $\omega = \pi/2$ zero in the frequency response characteristic at $\omega = 0$ and $\omega = \pi$ and its magnitude response is $\frac{1}{\sqrt{2}}$ at $\omega = \frac{4\pi}{9}$.

Solution:- The filter must have poles at $p_{1,2} = re$

And zero at $z = 1$ and $z = -1$

Consequently, the same system function is.

$$\Rightarrow H(z) = G \frac{(z-1)(z+1)}{(z-jr)(z+jr)}$$

$$\Rightarrow H(z) = \frac{G z^2 - 1}{z^2 + r^2}$$

The gain Factor is determined evaluating the frequency response $H(\omega)$ of the filter at $\omega = \pi/2$

$$H = \left(\frac{\pi}{2} \right) = G \frac{2}{1-r^2} = 1$$

$$G = \frac{1-r^2}{2}$$

The value of r is determined by evaluating $H(\omega)$ at $\omega = 4\pi/9$

we have

$$\begin{aligned} |H(4\pi/9)|^2 &= \frac{(1-r^2)^2}{4} \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)} \\ &= 1/2 \end{aligned}$$

\Rightarrow Equivalently

$$1 \cdot 94 (1-r^2)^2 = 1 - 1.88 r^2 + r^4$$

The value of $r^2 = 0.7$ satisfies this equation therefor the System function for the desired filter is.

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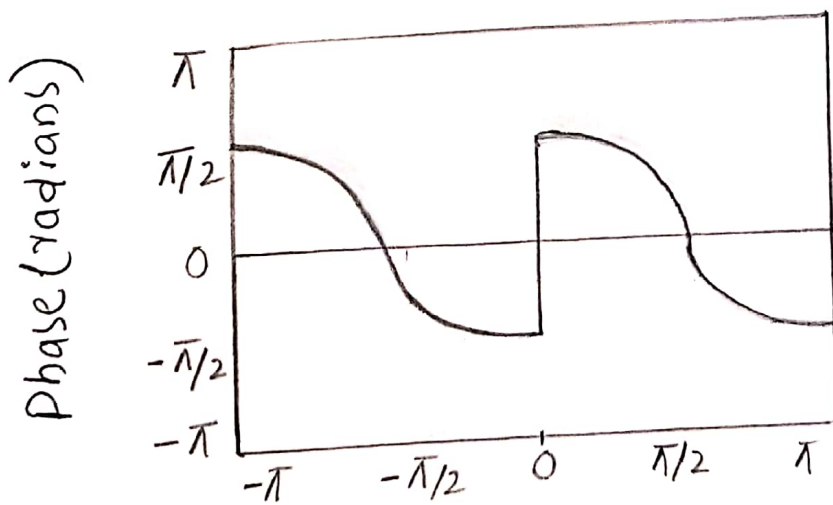
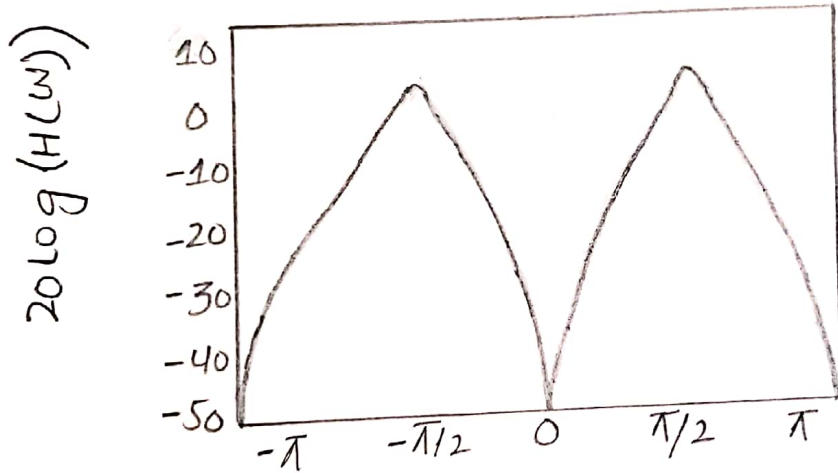
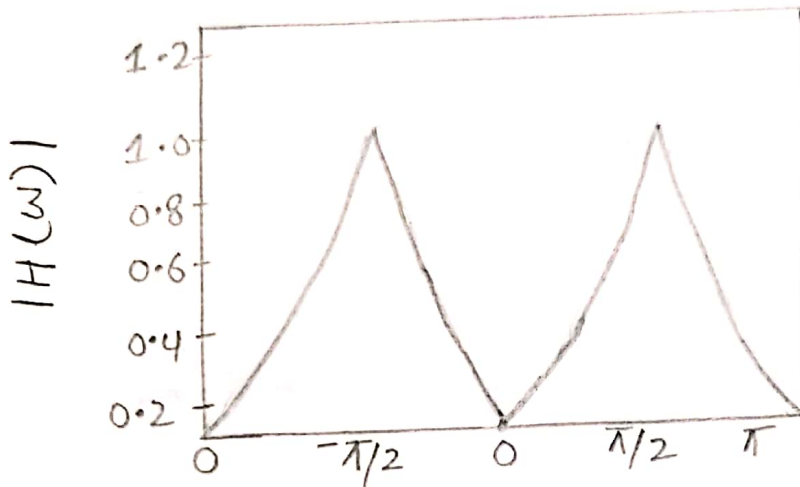
$H(z)$

$$= 0.15$$

$$\frac{1-z^{-2}}{1+0.7z^{-2}}$$

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its frequency response is illustrated



\Rightarrow Magnitude and phase response of a simple bandpass filter is $H(z) = 0.15 \left[\frac{1-z^{-2}}{1+0.7z^{-2}} \right]$

Question No (4)=> Part (A)

A finite duration sequence of length L given is

$$x[n] = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{otherwise} \end{cases}$$

Determine the N -point DFT $N \geq L$.

Solution :- The Fourier Transform of this sequence is

$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$X(\omega) = \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}}$$

$$= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}$$

The magnitude and phase of $X(\omega)$ are illustrated for $L=10$. The N -point DFT of $x(n)$ is simply $X(\omega)$ evaluated at the set of N equally spaced frequencies.

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$$\Rightarrow \omega_k = 2\pi k/N,$$

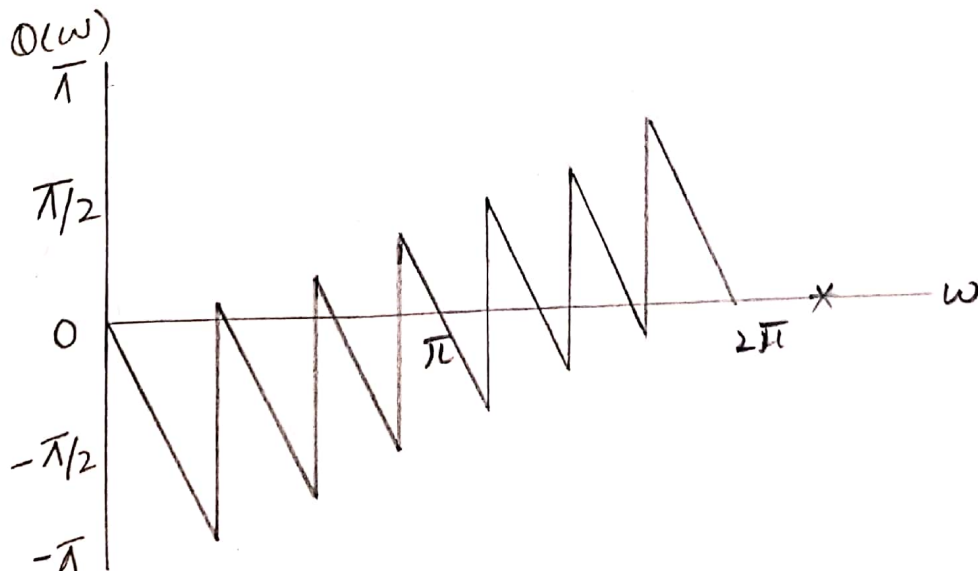
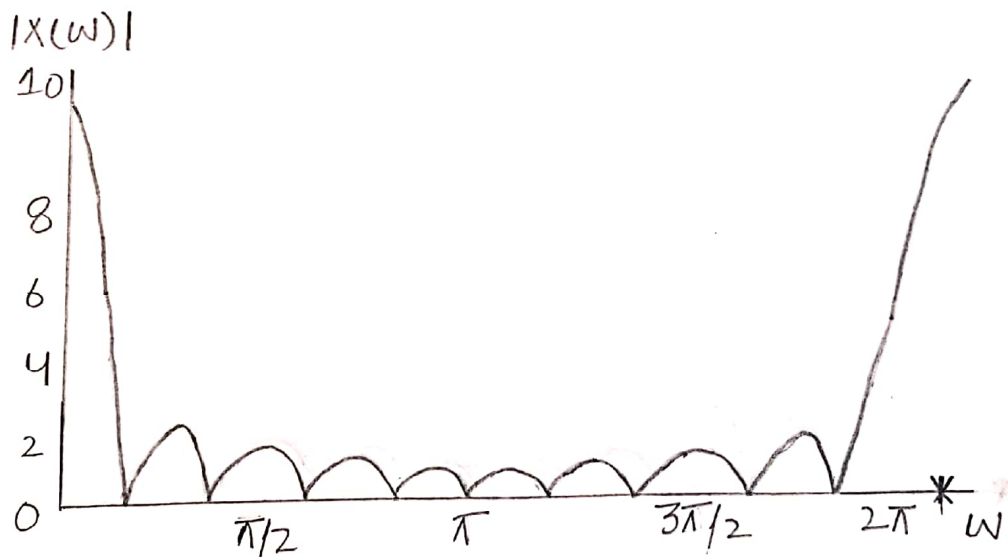
$$k = 0, 1, 2, \dots, N-1$$

Hence,

$$X(k) = \frac{1 - e^{-j2\pi kL/N}}{1 - e^{-j2\pi k/N}}$$

$$k = 0, 1, \dots, N-1$$

$$X(k) = \frac{\sin(\pi kL/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$



→ If N is selected such that $N=L$
then the Discrete Fourier Transform
becomes,

$$X(K) = \begin{cases} L, & K=0 \\ 0, & K=1, 2, \dots, L-1 \end{cases}$$

Thus there is only one non zero
value in the DFT This is apparent
from observation of $X(\omega)$, since
 $X(\omega) = 0$ at the frequencies $\omega_k =$
 $2\pi k/L$
 $k \neq 0$;

The reader should verify that $x(n)$
can be recovered from $X(K)$ by
performing an L -point IDFT
provides a plot of the N -point
DFT in magnitude phase for $L=10$
 $N=50$ and $N=100$ Now the
spectral characteristics of the
sequence are more clearly
evident. \longleftrightarrow

Question No (4)=> Part (B)

Perform the circular convolution of the following two sequence.

$$x_1(n) = \left\{ \underset{\uparrow}{2}, 1, 2, 1 \right\}$$

$$x_2(n) = \left\{ \underset{\uparrow}{1}, 2, 3, 4 \right\}$$

Solution:-

Each sequence consist of four non zero points for the purpose of illustrating the operations involved in circular convolution it is desired to graph each sequence as points on a circle. Thus the sequences $x_1(n)$ and $x_2(n)$ are graphed as illustrated we note that the sequences are graphed in a counterclockwise direction on a circle.

Now, $x_3(m)$ is obtained by circularly convolving $x_1(n)$ with

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\Rightarrow $x_1(n)$ with $x_2(n)$ as specified (17)

Beginning with $m=0$ we have

$$x_3(m) = \sum_{n=0}^3 x_1(n) x_2[(1-n)]_4$$

$x_2(-n)_4$ is simply the sequence $x_2(n)$

folded and graphed on a circle

The product sequence is obtained by

multiplying $x_1(n)$ with $x_2(-n)_4$ point

by point. Finally we sum the values

in the product sequence to obtain

$$x_3(0) = 14$$

For $m=1$ we have

$$x_3(1) = \sum_{n=0}^3 x_1(n) x_2(1-n)_4$$

it is easily verified that $x_2(1-n)_4$

is simply the sequence $x_2(-n)_4$

rotated counter clockwise by one

unit in the time.

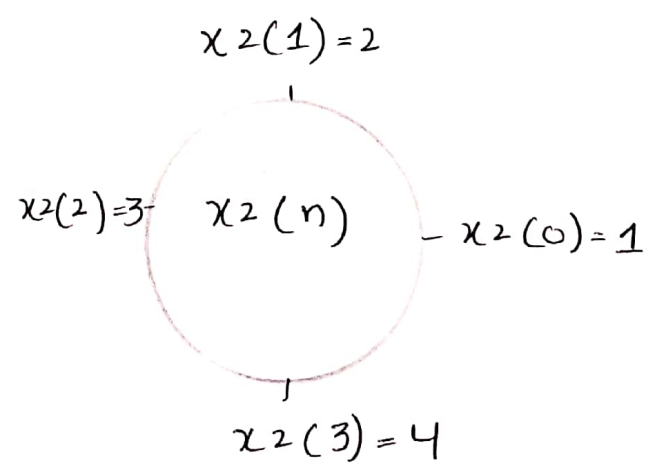
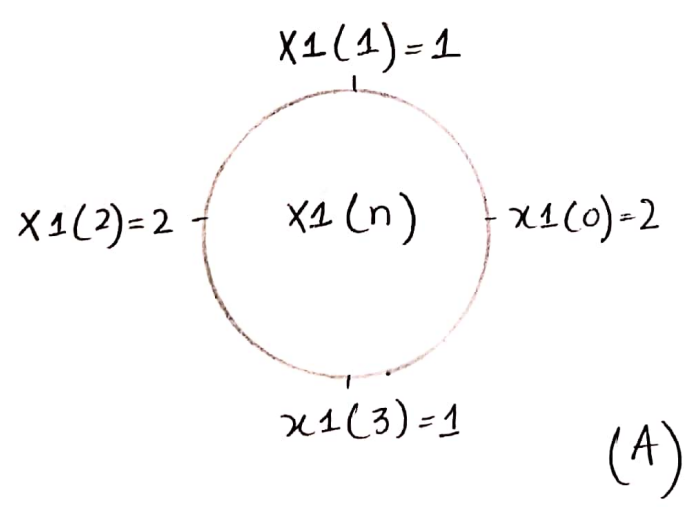
\Rightarrow This rotated sequence multiplies $x_1(n)$ to yield the product sequence also finally we sum the values in the product sequence to obtain $x_3(1)$ Thus.

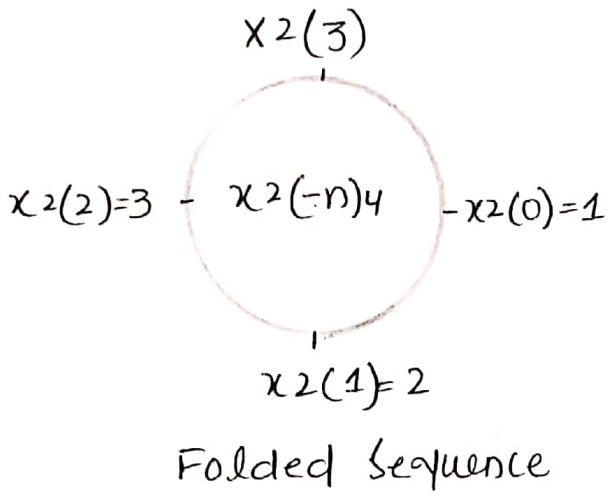
$$x_3(1) = 16$$

For $m=2$ we have

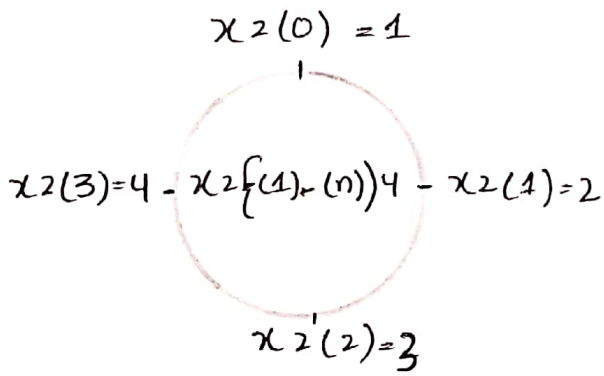
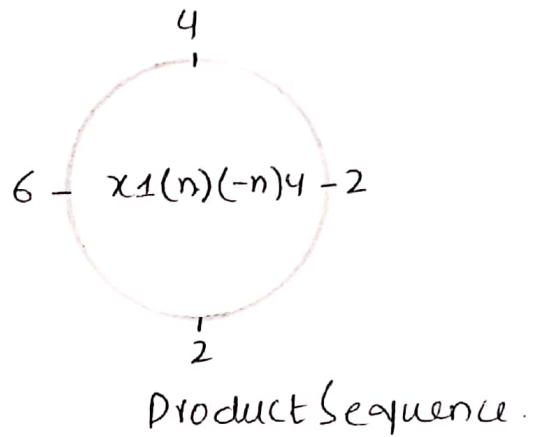
$$x_3(2) = \sum_{n=0}^3 x_1(n) x_2(2-n)$$

Now $x_2(2-n)$ is the folded sequence.

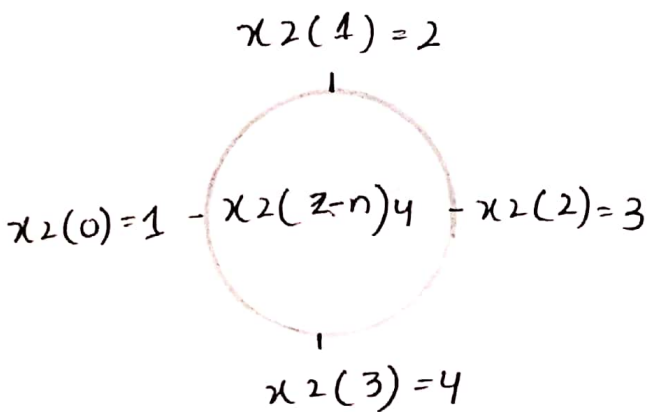
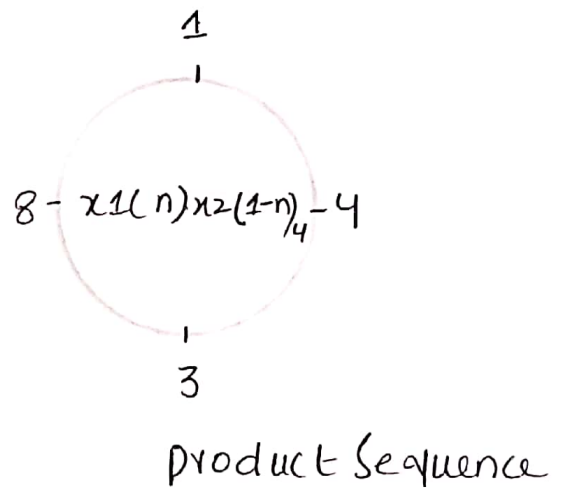




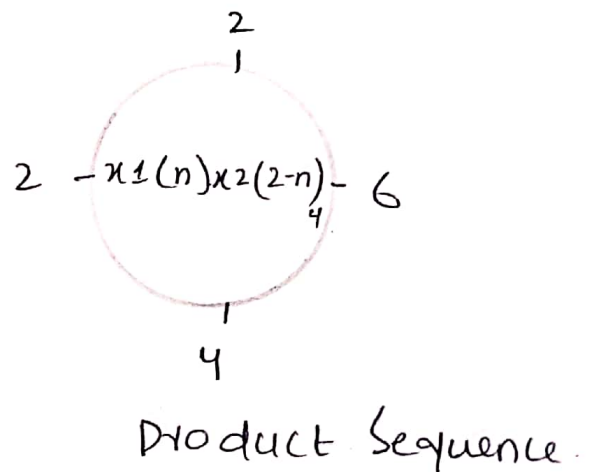
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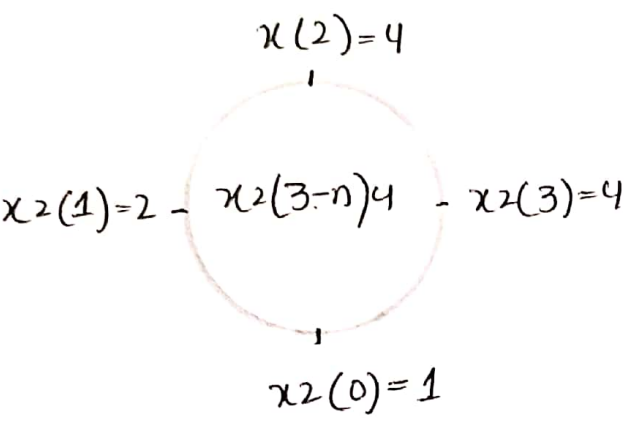


(C)



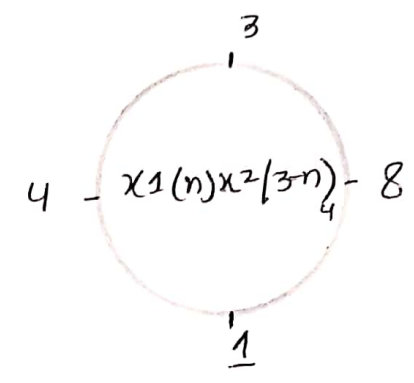
(D)





(E)

=> Folded Sequence rotated by three unit in time



=> Product Sequence.

Thank You