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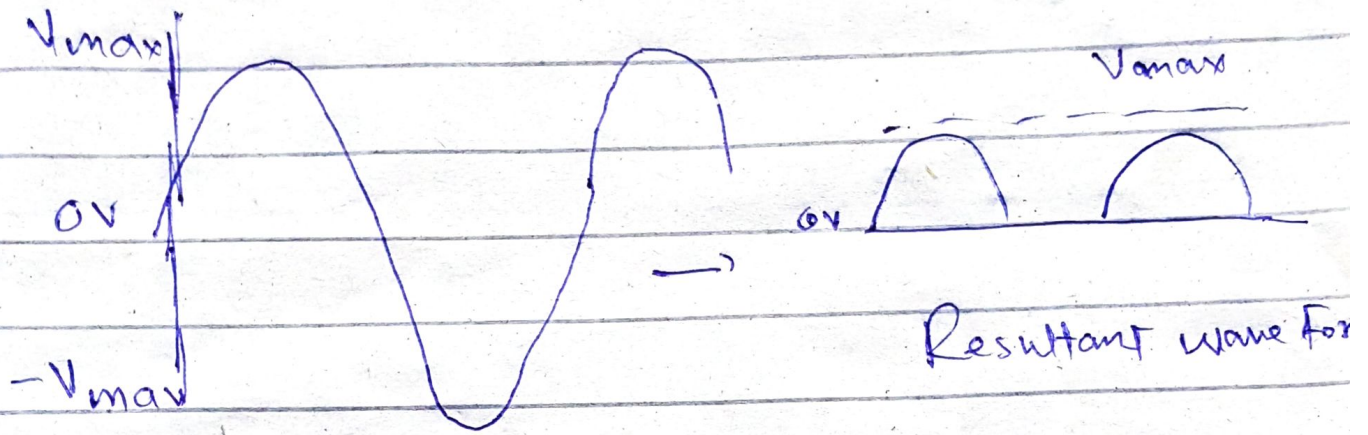
## Difference b/w s- $\phi$ half wave and Full wave bridge rectifier

① Half wave rectifier which convert only one half of the AC cycle into pulsating DC while Full wave rectifier is an electronic circuit which convert entire cycle AC into Pulsating DC.

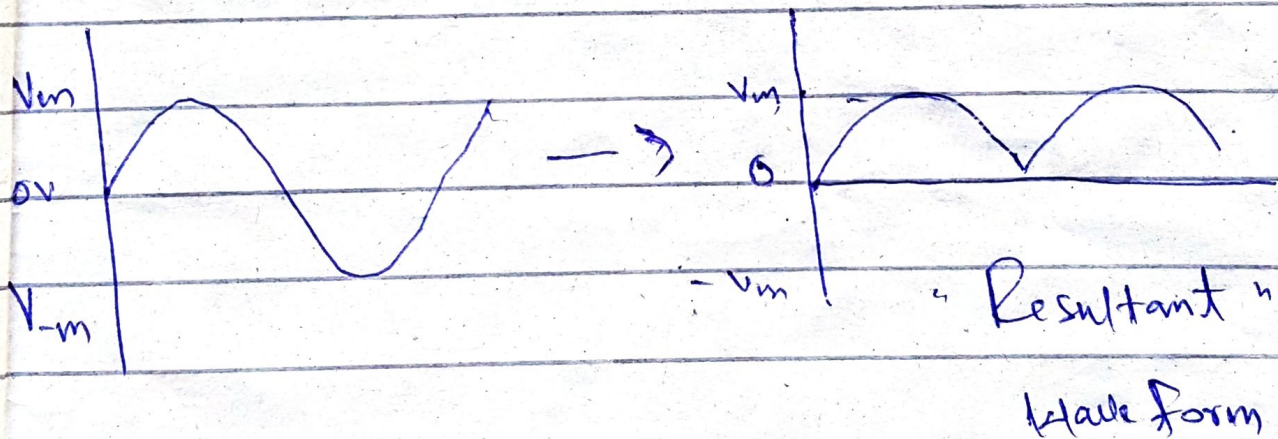
② Half wave utilize only half of the AC cycle for the conversion while Full wave utilize Full wave of the AC cycle.

③ Half wave is unidirectional, the conduction is one direction only either convert positive or negative, that why called Half wave rectifier while the Full wave rectifier is bidirectional, it conduct for positive half as well as negative half of the cycle.

(4) out put wave form of 1- $\phi$  single phase half wave rectifier



and single phase full wave bridge rectifier



(5) The number of diode in half wave rectifier is 1 while in bridge rectifier is 4.

Similarities b/w Single phase

Half wave and Full wave Bridge

rectifier

- ① Peak inverse voltage of Single half wave and full wave rectifier are same which is  $V_m$  and same in both rectifier
- ② Both utilize the Single phase for the operation.

2) 1- $\phi$  Uncontrolled and controlled rectifier (differences and similarities)

- ① Uncontrolled they are naturally turn on whenever a positive voltage is applied between its terminal and when you stop by applying it negative voltage

while in controll rectifier, the conduction start at any angle in positive half cycle namely 0 to 180 degree. Once the conduction start can be turn on and off.

Q.2

Solution

$$V_m = 26V$$

$$R = 12\Omega$$

We know that

For half wave

1)  $V_{dc}$

$$\Rightarrow \frac{V_m}{\pi} \quad \text{--- (1)}$$

where  $V_m = 26V$

$$\pi = 3.14$$

Put in equation (1)

$$\frac{26}{3.14} = 8.2V$$

For Full wave bridge

$$\frac{2V_m}{\pi} \quad \text{putting value}$$

$$\frac{2(26)}{3.14} \Rightarrow 16.5 \text{ V}$$

②  $I_{dc}$

For half wave

$$I_{dc} = \frac{V_m}{\pi R}$$

$$\Rightarrow \frac{26}{(3.14)(13)} = 0.64 \text{ A}$$

For Full wave

$$I_{dc} = \frac{I_m}{\pi} \quad \text{where } I_m = \frac{V_m}{R}$$

$$\Rightarrow \frac{2}{3.14} \Rightarrow 0.6 \text{ A}$$

$$\Rightarrow \frac{26}{13} = 2 \text{ A}$$

For half wave ③  $V_{rms}$

$$V_{rms} = \frac{V_m}{2}$$

$$\Rightarrow \frac{26}{2} = 13 \text{ V}$$

Put in above

For Full wave

$$V_{rms} = \frac{1}{\sqrt{2}} V_s$$

$$V_{rms} = 26.02 \text{ V}$$

$$V_s = \frac{V_m}{\sqrt{2}}$$

$$\Rightarrow \frac{26}{\sqrt{2}}$$

$$\Rightarrow 18.4 \text{ V}$$

## 4) Irms

4) Irms For half wave and Full wave  
bridge

For half wave

$$I_{rms} = \frac{V_m}{2R}$$

$$\Rightarrow \frac{26}{2(13)}$$

$$\Rightarrow 1A$$

For Full wave

$$\Rightarrow \frac{I_m}{2} \quad \text{--- * sign in$$

$$\text{where } I_m = \frac{V_m}{R}$$

$$\Rightarrow \frac{26}{13}$$

$$\Rightarrow 2A$$



Put in equation  $\star$  we get

$$\Rightarrow \frac{2}{2} = 1 \text{ A}$$

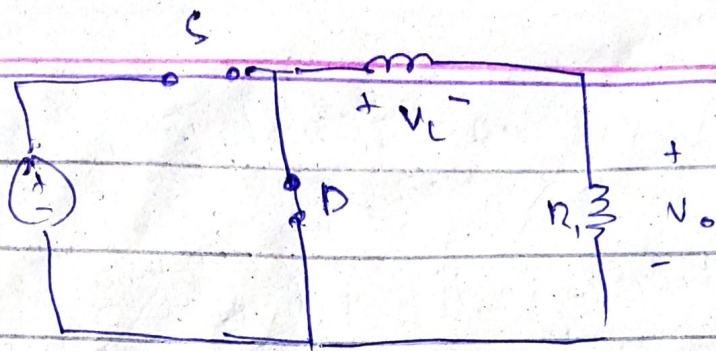
(5)

I would like to refer the uncontrolled Full wave bridge rectifier because the efficiency of the Full wave bridge rectifier is better than in half wave rectifier and output Frequency also greater than half wave rectifier.

Q.3)

## Buck chopper

- 1) Output voltage is less than input voltage.
- 2) The thyristor in the circuit acts as a switch.
- 3) When thyristor is ON, supply voltage appears across the load.
- 4) When thyristor is OFF, the voltage across the load will be zero.
- 5) Practical arrangement includes an inductor  $L$  and diode which are used to eliminate current pulsation providing a smooth DC current.
- 6) With  $S$  closed,  $D$  is OFF and it remains OFF as long as  $S$  is ON.
- 7) The i/p current builds up exponentially and flows through  $L$  and load.
- 8)  $V_o$  equal  $V_i$ .
- 9) With  $S$  OFF or OPEN, the current through  $L$  decays to zero.
- 10) This causes inductive voltage with opposite polarity across  $L$ .



(Circuit when Switch is opened)

Data:

$$V_{in} = 50V$$

$$D = 26\%$$

$$R = 13\Omega$$

$$\text{Frequency} = 20kHz$$

1)  $V_{out}$

We know that

$$V_{out} = \alpha V_s \quad \text{--- (1) or } V_{out} = D \times V_s$$

Here  $\alpha = \text{Duty cycle}$

$$\text{which } 26\% = 0.26$$

Put in equation (1) we get

$$\Rightarrow (0.26)(50)$$

$$\Rightarrow 13V$$

2)  $I_{out}$

$$I_o = \frac{V_o}{R}$$

$$\Rightarrow \frac{13}{13}$$

$$\Rightarrow 1A$$

3)  $I_{in}$

We know that

$$I_o = \frac{I_i}{\alpha \text{ or } d}$$

$$I_i = I_o \times d$$

where  $\alpha$  duty cycle or  $d$  is duty cycle

$$I_i = 1 \times 0.26$$

$$I_i = 0.26A$$

4) Inductor

We know that

$$L = \frac{T_{OFF} \times R}{2}$$

10% = 10

Let suppose  $T_{OFF} = 0.004$

$$L = \frac{0.004 \times 1341}{2}$$

$$L = 0.002684$$

10% (2)

Q.4)

## Boost chopper

- 1) The output voltage is more than the input voltage by several times
  - a) L is used to provide smooth i/p current
  - b) The SCR (S) acts as the switch which works in the PWM mode
  - c) With S ON, the L is connected to the supply
  - d) Load voltage  $V_L$  jumps instantaneously to  $V_s$ , but current through L increases linearly and stores energy
  - e) When S is open, the current collapses and energy stored in L is transferred to C through D.
  - f) The induced voltage across the inductor reverses and adds to the source voltage, increasing the o/p voltage.
  - g) The current that was flowing through S now flows through L, D and C to the load.

9) Energy stored in the inductor is released to the load.

10) When S closed again, D become reverse biased, the capacitor energy supplied the load voltage and cycle again

$$V_o = V_i + V_L$$

11)  $V_o$  is always higher than  $V_i$  as polarity of inductor voltage  $V_L$  is same as  $V_i$

12) If inductor  $L$  is very large, source current  $I_i$  is ripple free and considered constant.

$$(W_{ON} = V_i I_i T_{ON})$$

13) Assuming  $C$  to be large enough to neglect the voltage ripple,  $V_o$  is considered constant

$$W_{OFF} = (V_o - V_i) * I_i * T_{OFF}$$

14) Since losses are neglect, the energy stored transferred during  $T_{OFF}$  by  $L$  must be equal to energy gained during  $T_{ON}$ .

$$W_{ON} = W_{OFF} = V_i I_i T_{ON} = (V_o - V_i) * I_i$$

\* OFF

$$V_o = V_i \left( 1 + \frac{T_{ON}}{T_{OFF}} \right) = V_i \left( \frac{T}{T - T_{ON}} \right)$$

$$\Rightarrow V_i \left( \frac{1}{1 - \frac{T_{ON}}{T}} \right) = V_i \left( \frac{1}{1 - d} \right)$$

Thus  $V_o$  is always greater than  $V_i$

$$\rightarrow P_i = P_o \rightarrow V_i I_1 = \frac{V_o^2}{R} \rightarrow I_1 = \frac{V_o^2}{V_i} * \frac{1}{R}$$

$$\rightarrow I_o = I_1 * \frac{T_{OFF}}{T} \Rightarrow I_o = I_1 (1-d)$$

$$\rightarrow P_o = P_i \Rightarrow V_i I_1 = \frac{V_o^2}{R} = \frac{V_i^2}{(1-d)^2} * \frac{1}{R}$$

$$\rightarrow I_1 = \frac{V_i}{(1-d)^2} * \frac{1}{R}$$

$$\rightarrow I_L = \frac{I_{max} + I_{min}}{2} = I_1$$

$$I_{max} + I_{min} = 2 * I_1$$



- Voltage across L is

$$V_L = V_s = L \frac{di}{dt}$$

$$\Delta I_1 = \frac{V_L \times T_{ON}}{L} \Rightarrow I_{MAX} - I_{MIN} \Rightarrow \frac{V_i T_{ON}}{L}$$

Again  $I_{MAX} + I_{MIN} = 2 \times I_1$

Solving

$$I_{MAX} = V_i \left[ \frac{1}{R(1-d)^2} + \frac{T_{ON}}{2L} \right]$$

$$I_{MIN} = V_i \left[ \frac{1}{R(1-d)^2} - \frac{T_{ON}}{2L} \right]$$

$$I_{P-D} = I_{MAX} - I_{MIN} = \frac{V_i T_{ON}}{L}$$

For continuous current mode

$$I_{MIN} = V_i \left[ \frac{1}{R(1-d)^2} - \frac{T_{ON}}{2L} \right]$$

$$\Rightarrow \frac{1}{R(1-d)^2} = \frac{T_{ON}}{2L} \Rightarrow L = \frac{RT_{ON}(1-d)^2}{2}$$

(Boost)

Data

$$V_{in} = 50V$$

$$\text{duty cycle } D = 26\%$$

$$\text{Resistor, } R = 13\Omega$$

$$\text{Frequency, } F = 20\text{kHz}$$

1)  $V_{out}$

We know that

$$V_o = V_i \left( \frac{1}{1-d} \right)$$

$$V_o = 50 \left( \frac{1}{1-0.26} \right)$$

$$V_o = 50 \left( \frac{1}{0.74} \right)$$

$$V_o = 67.56V$$

2)

$I_{out}$

$$I_o = I_1 (1 - d) \quad \text{--- (1)}$$

First we have to find  $I_{in}$

3)  $I_{in}$

$$I_1 = \frac{V_i}{(1-d)^2} * \frac{1}{R}$$

$$\Rightarrow \frac{50}{(1-0.26)^2} * \frac{1}{13}$$

$$\Rightarrow \frac{50}{0.5476} * \frac{1}{13}$$

$$\Rightarrow \frac{50}{7.1188}$$

$\Rightarrow 7.04 \text{ A}$  Put this value  
equation (1)

$$I_0 = I_i (1 - d)$$

$$I_0 = 7(1 - 0.26)$$

$$I_0 = 7(0.74)$$

$$I_0 = 5.18 \text{ A}$$

#### 4) Inductor

$$L = \frac{R T_{\text{ON}} (1 - d)^2}{2} \quad \text{--- (1)}$$

$$d = \frac{T_{\text{ON}}}{T}$$

$$T = \frac{1}{20 \times 1000}$$

$$T = 20000$$

$$T_{\text{ON}} = d \times T \quad \text{ON Next page}$$

~~$$T_{\text{ON}} = 0.26 \times 20000 (0.00005)$$~~

~~$$T_{\text{ON}} = 5200 \text{ put in equation (1)}$$~~

~~$$L = \frac{13 \times 5200 (1 - 0.26)^2}{2}$$~~

~~$$L = \frac{13 \times 5200 (0.5476)}{2}$$~~

$$T_{ON} = 0.26 \times 0.00055$$

$$T_{ON} = 0.00013 \text{ (put in equation 1)}$$

$$L = \frac{R T_{ON} (1-d)^2}{2}$$

$$L = \frac{13 (0.00013) (1-0.26)^2}{2}$$

$$L = 0.00046 \text{ H}$$

Q.5

## Buck-Boost chopper:

- It combines the concept of both step-up and step-down choppers
- The output voltage is either higher or lower than input voltage
- The output voltage polarity also be reserved.
- The Switch is either an SCR or GTO or IGBT
- When S is ON, D is reverse bias and  $I_D$  is zero
- While S is OFF, the Source is disconnected
- The current through inductor does not change instantaneously and it forward biases the diode, providing path for the load
- With S = ON ( $T_{ON}$ );  $W_{ON} = V_1 * I_1 * T_{ON}$   
With S = OFF ( $T_{OFF}$ );  $W_{ON} = V_1 * I_1 * T_{OFF}$
- Ignoring losses,  $W_{ON} = W_{OFF} \Rightarrow V_1 * I_1 * T_{ON}$   
 $\Rightarrow V_1 * I_1 * T_{OFF}$
- $V_0 = V_1 \frac{dT}{(1-d)T} = V_1 \frac{d}{(1-d)}$

$$I_L = \frac{I_{\max} + I_{\min}}{2} ; I_1 = I_L d = \left( \frac{I_{\max} + I_{\min}}{2} \right) d$$

$\times d$

The average output  $P_o = V_o I_o = \left( \frac{I_{\max} + I_{\min}}{2} \right) V_o$

$\times dV \Rightarrow P_o = \frac{V_o^2}{R}$

$$\rightarrow I_{\max} + I_{\min} = \frac{2dV_1}{R(1-d)^2} ; I_{\max} - I_{\min} = \frac{V_1 d T}{L}$$

$$\rightarrow I_{\max} = V_1 \left[ \frac{1}{R(1-d)^2} + \frac{T}{2L} \right] d ;$$

$$I_{\min} = V_1 \left[ \frac{1}{R(1-d)^2} - \frac{T}{2L} \right] d ;$$

For continuous current condition

$$I_{\min} > 0 = V_1 \left[ \frac{1}{R(1-d)^2} - \frac{T}{2L} \right] d$$

$$\Rightarrow L = \frac{RTd}{2} (1-d)^2$$

Data)

$$V_{in} = 50V$$

$$V_{out} = ~~12\%~~ 26\%$$

$$\text{Resistor, } R = 13\Omega$$

$$\text{Frequency, } f = 20\text{KHz}$$

a) Duty cycle (D)

We know that

$$\frac{V_o}{V_i} = \frac{-D}{1-D}$$

$$V_o = +V_i \frac{d}{(1-d)}$$

$$0.26 = +50 \frac{d}{1-d}$$

$$(0.26)(1-d) = 50d$$

$$0.26 - 0.26d = 50d$$

$$0.26 = 50d + 0.26d$$

$$0.26 = 50.26d$$

P.T.O



$$\frac{0.26}{50.26} = \frac{50.26d}{50.26}$$

∴ d = 0.0051

$$d = 0.0051$$

2)  $I_{out}$

We know that

$$I_{max} + I_{min} = \frac{2dV_i}{R(1-d)^2}$$

retubor (1)

$$I_{max} + I_{min} = \frac{2(0.0051)(50)}{13(1-0.0051)^2}$$

$$\Rightarrow \frac{0.51}{12.9}$$

$$\Rightarrow 0.039 \text{ — (1)}$$

We know that

$$I_{out} = \frac{I_{max} + I_{min}}{2} \text{ Put eqn (1) value}$$

$$\Rightarrow \frac{0.039}{2} \Rightarrow 0.0195A$$

$$3) I_i = ?$$

$$I_i = I_d$$

$$\Rightarrow 0.0195 \times 0.0051$$

$$\Rightarrow \cancel{0.00099}$$

$$\Rightarrow 0.000099 \text{ A}$$

#### 4) Inductor

$$L = \frac{RTd(1-d)^2}{2}$$

Putting value where  $T = \frac{1}{F}$

$$L = \frac{13(0.00005)(1-0.0051)^2}{2} \quad \left. \begin{array}{l} T = \frac{1}{20 \times 1000} \\ T = 0.00005 \end{array} \right\}$$

$$L = \frac{13(0.00005)(0.98)}{2}$$

$$L = 0.00031 \text{ H}$$