



	(b)	Design a two-pole bandpass filter that has the center of its passband at $\omega = \pi/2$ , zero in its frequency response characteristics at $\omega = 0$ and $\omega = \pi$ and its magnitude response in $\frac{1}{\sqrt{2}}$ at $\omega = 4\pi/9$ .	
Q 4	(c)	A finite duration sequence of Length L is given as $x(n) = \{1, \quad 0 \leq n \leq L - 1, \quad \textit{otherwise}$ Determine the N- point DFT of this sequence for $N \geq L$	Marks 4
	(d)	Compute the DFT of the four-point sequence $x(n) = (0 \ 1 \ 2 \ 3)$	

Q No 1

(a)

Sol: The characteristic equation is

$$\lambda^2 - 4\lambda + 4 = 0$$

$$\lambda = 2, 2$$

$$y_h(n) = c_1 2^n + c_2 n 2^n$$

The particular solution is

$$y_p(n) = k (-1)^n u(n)$$

$$k (-1)^n u(n) - 4k (-1)^{n-1} y(n-1) + 4k (-1)^{n-2} y(n-2) =$$

$$(-1)^n u(n) - (-1)^{n-1} y(n-1)$$

$$\text{For } n=2 \quad k(1+4+4) \Rightarrow 2 \Rightarrow k$$

$$\Rightarrow k = \frac{2}{9} \quad \text{total solution is}$$

$$y(n) = \left[ c_1 2^n + c_2 n 2^n + \frac{2}{9} (-1)^n \right] u(n)$$

For initial condition we have  $y(0)$

$$y(0) = 1, \quad y = 2 \quad \text{then}$$

$$c_1 + \frac{2}{9} = 1$$

$$\Rightarrow c_1 = \frac{7}{9}$$

$$2c_1 + 2c_2 - \frac{2}{9} = 2$$

(2)

$$\Rightarrow C_2 = \frac{1}{3}$$

Q No 1  
part (b)

Sol :- The characteristic equation is

$$d^2 - 0.7d + 0.1 = 0$$

$$d = \frac{1}{2}, \frac{1}{5}$$

$$y_n(n) = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{1}{5}\right)^n$$

with  $x(n) = y(n)$  we have

$$y(0) = 2$$

$$y(1) = 0.7y(0) = 0 \Rightarrow (1) = 1.4$$

Hence  $C_1 + C_2 = 2$  and

$$\frac{1}{2}C_1 + \frac{1}{5}C_2 = 1.4 = \frac{7}{5}$$

$$\Rightarrow C_1 + \frac{2}{5}C_2 = \frac{14}{5}$$

the evaluation yield

$$C_1 = \frac{10}{3}, C_2 = -\frac{4}{3}$$

$$h(n) = \frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \Big] u(n)$$

(3)

The set response is

$$O(n) = \sum_{k=0}^n h(n-k)$$

$$\frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) - \frac{4}{3} \left(\frac{1}{5}\right)^n (5^{n+1} - 1)$$

$u(n)$ .



(4)

Q No 2

(a) :-

Sol :-

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

$$X(z) = \frac{1}{(1+2z^{-1})(1-z^{-1})^2}$$

$$X(z) = \frac{1}{4(1+2z^{-1})} + \frac{3}{4} \frac{1}{1-2z^{-1}} + \frac{1}{2} \frac{z^{-1}}{(1-z^{-1})^2}$$

By applying inverse transform.

$$X(n) = \frac{1}{8} (-1)^n u(n) - \frac{3}{8} u(n) + \frac{1}{2} n u(n)$$
$$= \left[ \frac{1}{8} (-1)^n + \frac{3}{8} + \frac{n}{2} \right] u(n)$$

Q No 2

(b)

Sol:-

First we eliminate the negative power by multiplying both numerator and denominator by  $z^2$  thus

$$X(z) = \frac{z^2}{z^2 - 1.5z + 0.5}$$

The poles are  $X(z)$  are  $p_1=1$  and  $p_2=0.5$

$$\frac{X(z)}{z} = \frac{z}{(z-1)(z-0.5)} = \frac{A_1}{z-1} + \frac{A_2}{z-0.5}$$

To determine  $A_1$  and  $A_2$  is to multiply the equation by denominator term  $(z-1)(z-0.5)$

$$z = (z-0.5)A_1 + (z-1)A_2$$

$z=p_1=1$

$$1 = (1-0.5)A_1$$

thus we obtain the result  $A_1=2$  and set  $z=p_2=0.5$  so eliminating  $A_1$  we have

$$0.5 = (0.5-1)A_2$$

$$A_2 = -1$$

So the result of partial fraction is

$$X(z) = \frac{z}{z-1} - \frac{1}{z-0.5}$$

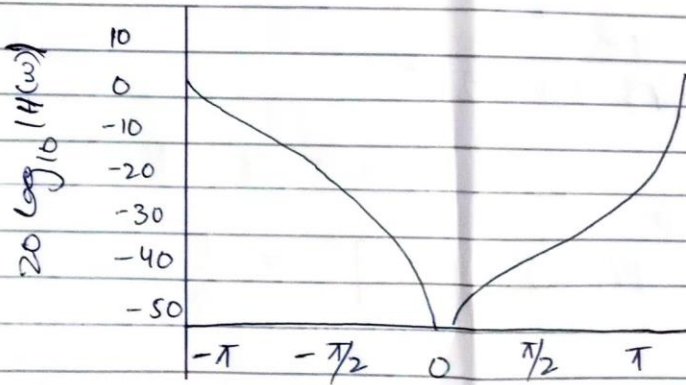
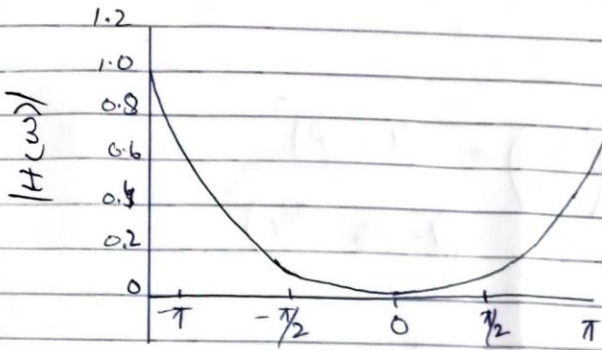
b

Q No 3 (a) :-

Sol :- At  $\omega = 0$  we have

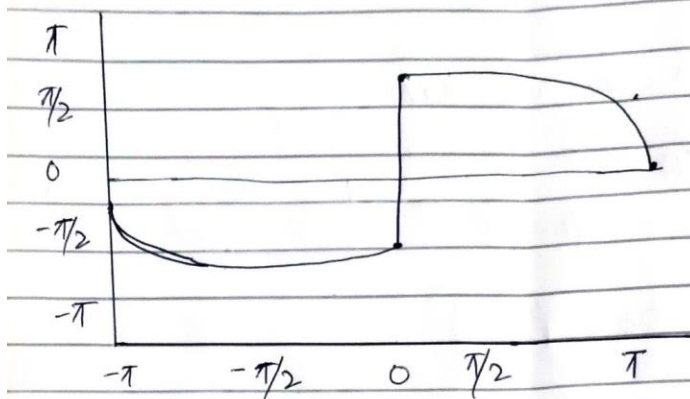
$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

$$\text{Hence } b_0 = (1-p)^2$$





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$$\text{At } \omega = \pi/4 \quad H\left(\frac{\pi}{4}\right) = \frac{(1-p)^2}{(1-pe^{-j\pi/4})^2}$$

$$= \frac{(1-p)^2}{1-p(\cos(\pi/4) + jp \sin(\pi/4))^2}$$

$$= \frac{(1-p)^2}{(1-p/\sqrt{2} + jp/\sqrt{2})^2}$$

$$= \frac{(1-p)^4}{[(1-p/\sqrt{2})^2 + p^2/2]^2} = \frac{1}{2}$$

Q No 3

Part (b):-

Sol: By the filter requirements:-Poles  $P_{1,2} = re^{\pm j\pi/2}$  pass band centerZeros  $Z_{1,2} = \pm 1$  Stopband Center.

$$\Rightarrow H(z) = G \frac{(z-1)(z+1)}{(z-jr)(z+jr)} = G \frac{z^2-1}{z^2+r^2}$$

By the filter requirement

$$H\left(\frac{\pi}{2}\right) = G \frac{-2}{-1+r^2} = 1$$

$$= G \frac{1-r^2}{2}$$

To set  $r$  use  $H\left(\frac{4\pi}{9}\right) = 1/\sqrt{2}$ 

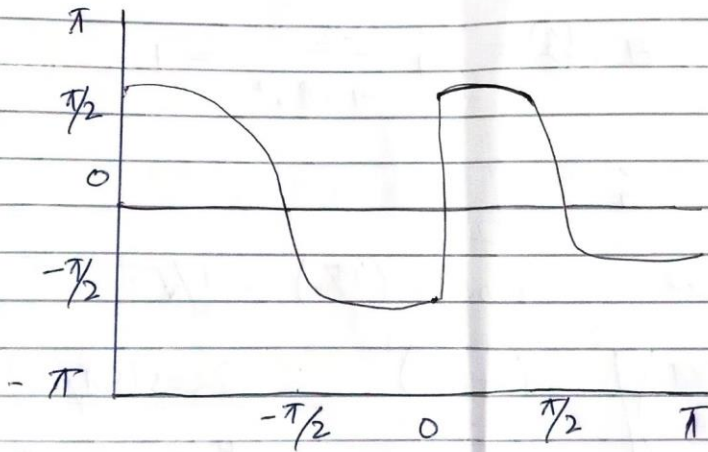
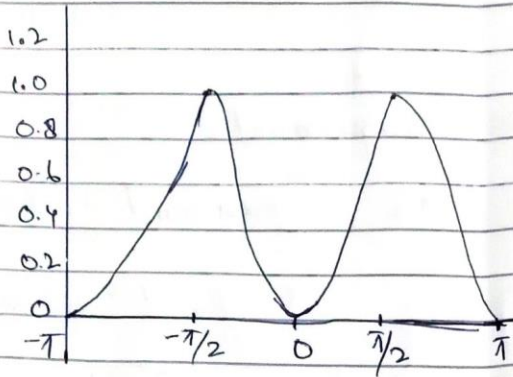
$$\left| H\left(\frac{4\pi}{9}\right) \right|^2 = \frac{(1-r^2)^2}{4} \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)}$$

$$= \frac{1}{2}$$

Evaluating gives  $r^2 = 0.7$  therefore

$$H(z) = 0.15 \frac{1-z^{-2}}{1+0.7z^{-2}}$$

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Q No 4

(a)

Sol: The Fourier transform of this sequence is

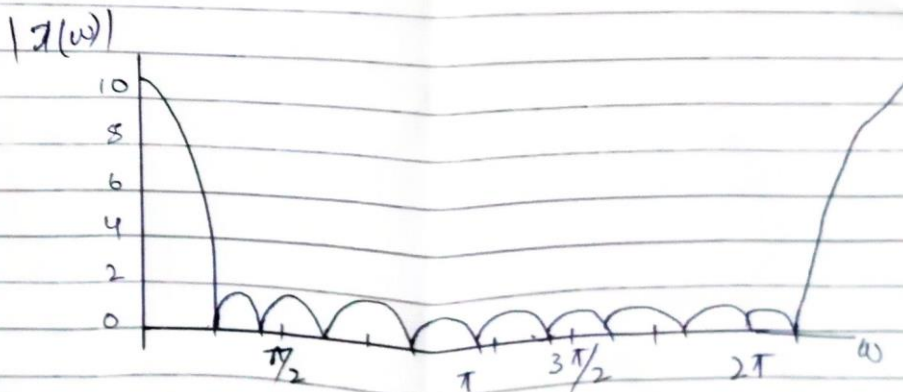
$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} =$$

$$= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}$$

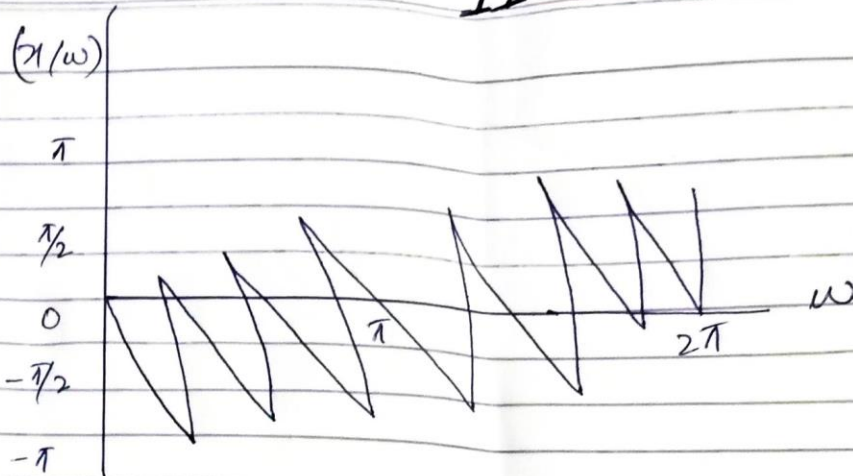
$$X(k) = \frac{1 - e^{-j2\pi kL/N}}{1 - e^{-j2\pi k/N}} \quad k = 0, 1, \dots, N-1$$

$$= \frac{\sin(\pi kL/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$





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if  $N$  is selected such that  $N=L$  then the DFT becomes

$$X(k) = \begin{cases} L & k=0 \\ 0 & k=1, 2, \dots, L-1 \end{cases}$$

thus there is only one nonzero value in DFT.



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Q No 4

b) Sol:-  $\omega_N^{k+N/2} = \omega_N^k$

Then Matrix can be extend by

$$\omega_N = \begin{bmatrix} \omega_N^0 & \omega_N^0 & \omega_N^2 & \omega_N^{N-1} \\ \omega_N^0 & \omega_N^2 & \omega_N^4 & \omega_N^{2(N-1)} \\ \omega_N^0 & \omega_N^2 & \omega_N^4 & \omega_N^{N-1} \\ \omega_N^0 & \omega_N^{2(N-1)} & \omega_N^{N-1(N-1)} & \omega_N^{N-1} \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & \omega_N^1 & \omega_N^2 & \omega_N^3 \\ 1 & \omega_N^2 & \omega_N^0 & \omega_N^4 \\ 1 & \omega_N^1 & \omega_N^3 & \omega_N^4 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

Then  $X_4 = \omega_N \times 4 = \begin{bmatrix} 4 \\ -2 + 2j \\ -2 \\ -2 - 2j \end{bmatrix}$  Ans