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## Question 1

- a) Determine the response  $y(n)$ ,  $n \geq 0$  of the system described by the second order difference equation:

$$y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$$

To the input  $x(n] = (-1)^n u(n)$ . And the initial conditions are  $y(-1) = y(-2) = 0$ .

**Solution:**

$$\lambda^2 - 4\lambda + 4 = 0$$

$$\lambda = 2, 2 \quad \text{Hence}$$

$$y_h(n) = C_1 2^n + C_2 2^n$$

The particular solution is

$$y_p(n) = k(-1)^n u(n)$$

Substituting this solution into the difference equation we obtain

$$k(-1)^n u(n) - 4k(-1)^{n-1} u(n-1) + 4k(-1)^{n-2} u(n-2) =$$

$$(-1)^n u(n) - (-1)^{n-1} u(n-1)$$

$$\text{For } n=2, k(1+4+4) = 2$$

$k = 2/9$  The total solution is

$$y(n) = [C_1 2^n + C_2 2^n + 2/9 (-1)^n] u(n)$$

From the initial conditions

we obtain,  $y(0) = 1$ ,  $y(1) = 2$

Then

$$c_1 + 2/9 = 1$$

$$\Rightarrow c_1 = 7/9$$

$$\Rightarrow 2c_1 + 2c_2 - 2/9 = 2$$

$$\Rightarrow c_2 = 1/3 \text{ Ans.}$$

b) Determine the impulse response and unit step response of the system described by the difference equation.

$$y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$$

Solution:

The characteristic equation is

$$\lambda^2 - 0.7\lambda + 0.1 = 0$$

$$\lambda = 1/2, 1/5 \text{ Hence}$$

$$y_h(n) = c_1 (1/2)^n + c_2 (1/5)^n$$

with  $x(n) = \delta(n)$  we have

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0$$

$$\Rightarrow y(1) = 1.4$$

Hence,  $c_1 + c_2 = 2$

And,

$$1/2 c_1 + 1/5 = 1.4$$

$$1.4 = 7/5$$

$$c_1 + 2/5 c_2 = 14/5$$

These equations yield

$$c_1 = 10/3, c_2 = -4/3$$

$$h(n) = \left[ \frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

The step response is

$$s(n) = \sum_{k=0}^n h(n-k)$$

$$s(n) = \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$s(n) = \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$s(n) = \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{4}{3} \left(\frac{1}{5}\right)^n$$

$$\left(5^{n+1} - 1\right) u(n) \text{ Ans.}$$

## QUESTION 2

- a) Determine the causal signal  $x(n]$  having the Z-transform

$$X(z) = \frac{4}{(1-2z^{-1})(1-z^{-1})^2}$$

Solution:

Taking inverse and Z-transform

$$\frac{A}{[1-2z^{-1}]} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$A=4, B=-3, C=-1$$

Hence

$$x(n) = [4(2)^n - 3 - n] u(n) \quad \text{Ans}$$

b) Perform the circular convolution of the following two sequences. Solve the problem step by step.

$$x_1(n) = \{ \underset{\uparrow}{2}, 1, 2, 1 \}$$

$$x_2(n) = \{ \underset{\uparrow}{1}, 2, 3, 4 \}$$

Solution:

Each sequence consist of four non zero points for the purpose of illustrating the operations involved in circular convolution it is decided to graph each sequence as points on a circle. Thus the sequences  $x_1(n)$  and  $x_2(n)$  are graphed as illustrated

We note that the sequences are graphed in a counterclockwise direction on a circle.

Now,  $x_3(m)$  is obtained by circularly convolving  $x_1(n)$  with  $x_2(n)$  as specified. Beginning with  $m=0$ , we have

$$x_3(m) = \sum_{n=0}^3 x_1(n) x_2[(-n)]_4$$

$x_2(-n)_4$  is simply the sequence  $x_2(n)$  folded and graphed on a circle. The product sequence is obtained by multiplying  $x_1(n)$  with  $x_2(-n)_4$  point by point. Finally we sum the values in the product sequence to obtain

$$x_3(0) = 14$$

For  $m=1$ , we have

$$x_3(1) = \sum_{n=0}^3 x_1(n) x_2(1-n)_4$$

It is easily verified that  $x_2(1-n)_4$  is simply the sequence  $x_2(-n)_4$  rotated counterclockwise by one unit in the time.

$\Rightarrow$  This rotated sequence multiplies  $x_1(n)$  to yield the product sequence. Also finally we sum the values in the product sequence to obtain  $x_3(1)$ . Thus

$$x_3(1) = 16$$

For  $m=2$  we have

$$x_3(2) \sum_{n=0}^3 x_1(n) x_2(2-n)4$$

Now  $x_2(2-n)4$  is the folded sequence.

$$x_1(0) = 1$$

$$x_2(0) = 2$$

$$x_1(2) = 2 \quad \text{circled } x_1(n) \quad x_1(0) = 2$$

$$x_2(2) = 3 \quad \text{circled } x_2(n) \quad x_2(0) = 1$$

$$x_1(3) = 1$$

$$x_2(3) = 4$$

(A)

$$x_2(3)$$

$$x_2(2) = 3 \quad \text{circled } x_2(n) \quad x_2(0) = 1$$

$$4$$

$$6 \quad \text{circled } x_1(n) \quad x_2(n)4 \quad 2$$

$$x_2(1) = 2$$

$$2$$

Folded sequences

Product sequence

(B)

$$x_2(0) = 1$$

$$x_2(3) = 4 \quad \text{circled } x_2(n) \quad x_2(1) = 2$$

$$1$$

$$8 \quad \text{circled } x_1(n) \quad x_2(n)4$$

$$x_2(2) = 3$$

$$4$$

$$3$$

Folded sequence  
rotated by ~~two~~  
one unit in time

Product  
sequence

(C)

$$x_2(1) = 2$$

$$x_2(0) = 1 \quad x_2(1) = 2 \quad x_2(2) = 3$$

$$x_2(3) = 4$$

$$2 \quad \begin{matrix} 2 \\ x_2(n) \\ x_2(2-n) \\ 4 \\ 4 \end{matrix} \quad 6$$

Folded sequence rotated  
by two unit in time (D)

Product  
sequence

$$x_2(2) = 4$$

$$x_2(1) = 2 \quad x_2(3) = 4 \quad x_2(3) = 4$$

$$x_2(0) = 1$$

$$4 \quad \begin{matrix} 3 \\ x_2(n) \\ x_2(3-n) \\ 1 \end{matrix} \quad 8$$

Folded sequence rotated  
by three unit in time

Product  
sequence

## QUESTION 3

- a) A two-pole low pass filter has the system response:

$$H(z) = \frac{b_0}{(1 - pz^{-1})^2}$$

Determine the value of  $b_0$  and  $p$  such

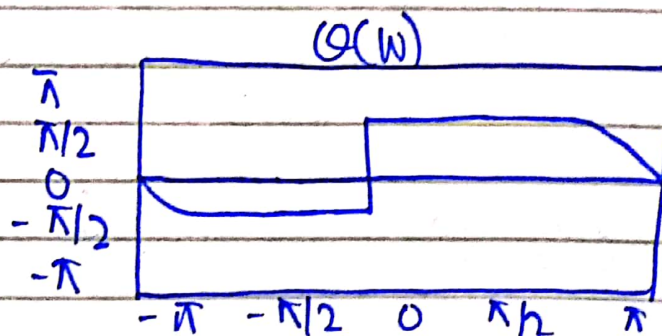
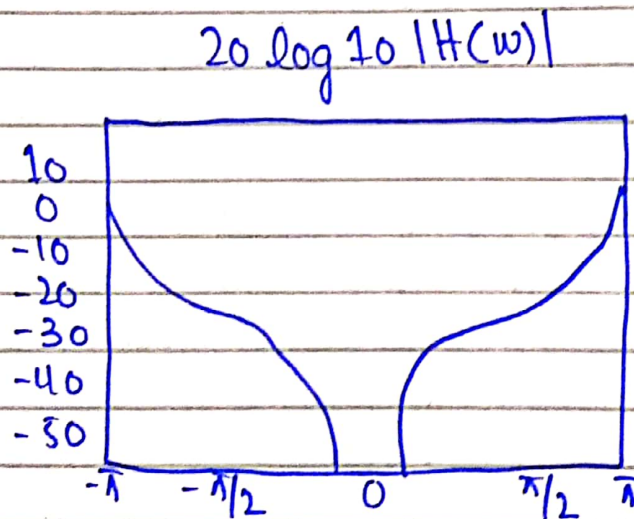
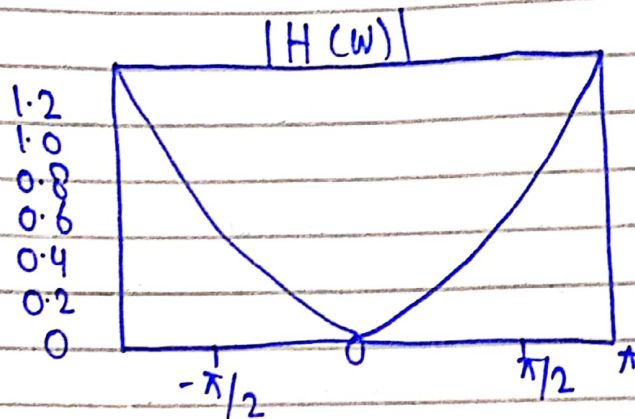


Solution:

At  $\omega=0$  we have

$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

$$\text{Hence } b_0 = (1-p)^2$$



At  $\omega = \pi/4$

$$H(\pi/4) = \frac{(1-p)^2}{(1-p e^{-j\pi/4})^2}$$

$$= \frac{(1-p)^2}{[1-p \cos(\pi/4) + jp \sin(\pi/4)]^2}$$

$$= \frac{(1-p)^2}{[1-p(\frac{1}{\sqrt{2}} + j\frac{1}{\sqrt{2}})]^2}$$

Hence

$$= \frac{(1-p)^4}{[1-p(\frac{1}{\sqrt{2}})^2 + p^2/2]^2}$$

$$= 1/2$$

$\Rightarrow$  Equivalently

$$\sqrt{2} (1-p)^2 = 1+p^2 - \sqrt{2}p$$

The system function for desired filter

$$H(z) = \frac{0.46}{(1-0.32z^{-1})^2} \quad \text{Ans}$$

- b) Design a two-pole bandpass filter that has the center of its passband at  $\omega = \pi/2$ , zero in its frequency response characteristics at  $\omega = 0$  and  $\omega = \pi$  and its magnitude response is  $\frac{1}{\sqrt{2}}$  at  $\omega = 4\pi/9$

Solutions:

The filter must have poles at

$$P_{1,2} = \pm \gamma e$$

And zero at  $z=1$  and  $z=-1$  consequently the same system function is

$$\Rightarrow H(z) = C_1 \frac{(z-1)(z+1)}{(z-j\gamma e)(z+j\gamma e)}$$

$$\Rightarrow H(z) = \frac{C_1 z^2 - 1}{z^2 + \gamma^2 e^2}$$

The gain factor is determined evaluating the frequency response  $H(\omega)$  of the filter at  $\omega = \pi/2$

$$H = H(\pi/2) = C_1 \frac{2}{1 - \gamma^2 e^2} = 1$$

$$C_1 = \frac{1 - \gamma^2 e^2}{2}$$

The value of  $\gamma$  is determined by evaluating  $H(\omega)$  at  $\omega = 4\pi/9$  we have

$$\begin{aligned} |H(4\pi/9)|^2 &= \left( \frac{1 - \gamma^2 e^2}{2} \right)^2 \frac{2 - 2 \cos(8\pi/9)}{1 + \gamma^4 e^4 + 2\gamma^2 e^2 \cos(8\pi/9)} \\ &= 1/2 \end{aligned}$$

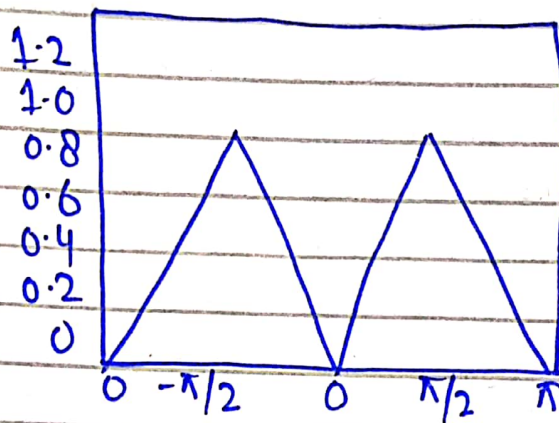
$\Rightarrow$  Equivalently

1.94  $(1-z^{-2})^2 = 1 - 1.88z^{-2} + z^{-4}$  the value of  $z^{-2} = 0.7$  satisfies this equation therefore the system function for the desired filter is

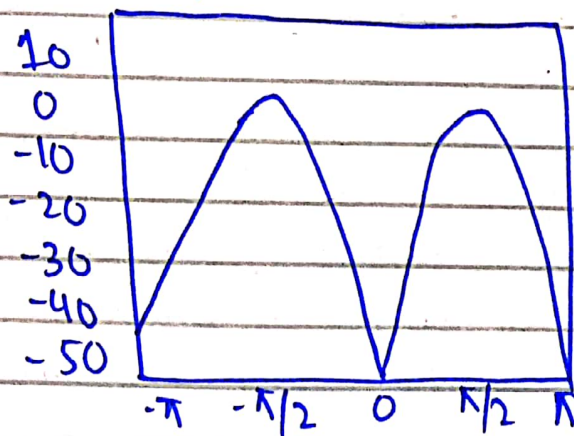
$$\Rightarrow H(z) = 0.15 \frac{1-z^{-2}}{1+0.7z^{-2}}$$

its frequency response is illustrated

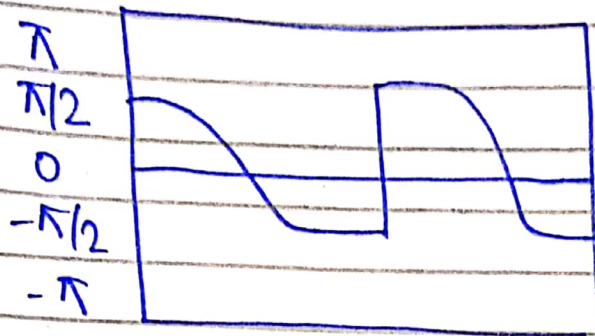
$|H(\omega)|$



$20 \log(|H(\omega)|)$



## Phase radiation



Magnitude and Phase response of a simple bandpass filter is  $H(z)$   
 $= 0.15 [1 - z^{-2}] / (1 + 0.7z^{-2})$

Ans.