

Department of Electrical Engineering

Sessional Assignment

Date: 01/06/2020

Course Details

Course Title: _____ Digital Signal Processing _____

Module: _____ 6th _____

Instructor: _____ Sir Pir Mehr _____

Total Marks: _____ 20 _____

Student Details

Name: _____ Bakht Zaman Gohar _____

Student ID: _____ 13678 _____

Q1.	(a)	Determine the response $y(n)$, $n \geq 0$, of the system described by the second order difference equation $y(n) - 3y(n - 1) - 4y(n - 2) = x(n) + 2x(n - 1)$ To the input $x(n) = 4^n u(n)$.	Marks 6
	(b)	Determine the impulse response and unit step response of the systems described by the difference equation. $y(n) = 0.6y(n - 1) - 0.8y(n - 2) + x(n)$	
Q2.	(a)	Determine the causal signal $x(n)$ having the z-transform $x(z) = \frac{1}{(1 - 2z^{-1})(1 - z^{-1})^2}$ (Hint: Take inverse z-transform using partial fraction method)	Marks 6
	(b)	Determine the partial fraction expansion of the following proper function $X(z) = \frac{1}{1 - 1.5z^{-1} + 0.5z^{-2}}$	
Q.3	(a)	A two- pole low pass filter has the system response $H(z) = \frac{b_o}{(1 - pz^{-1})^2}$ Determine the values of b_o and p such that the frequency response $H(\omega)$ satisfies the condition $H(0) = 1$ and $\left H\left(\frac{\pi}{4}\right)\right ^2 = \frac{1}{2}$.	Marks 4

	(b)	Design a two-pole bandpass filter that has the center of its passband at $\omega = \pi/2$, zero in its frequency response characteristics at $\omega = 0$ and $\omega = \pi$ and its magnitude response in $\frac{1}{\sqrt{2}}$ at $\omega = 4\pi/9$.	
Q 4	(c)	A finite duration sequence of Length L is given as $x(n) = \begin{cases} 1, & 0 \leq n \leq L - 1 \\ 0, & \text{otherwise} \end{cases}$ Determine the N- point DFT of this sequence for $N \geq L$	Marks 4
	(d)	Compute the DFT of the four-point sequence $x(n) = (0 \ 1 \ 2 \ 3)$	

IQRA NATIONAL UNIVERSITY PESHAWAR



Sessional Assignment

Digital Signal Processing

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Submitted to

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Q 1:- (Part - a)

* Solution:-

Consider the difference equation

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

The homogenous equation of the system

$$\text{is; } y(n) - 3y(n-1) - 4y(n-2) = 0$$

The characteristic equation of the system is;

$$\lambda - 3\lambda^{-1} - 4\lambda^{-2} = 0$$

$$\lambda^2 - 3\lambda - 4 = 0$$

Determine the root of the characteristic equation

$$\lambda^2 - 4\lambda + \lambda - 4 = 0$$

$$\lambda(\lambda - 4) + 1(\lambda - 4) = 0$$

$$(\lambda - 4)(\lambda + 1) = 0$$

$$\lambda_1 = 4, \quad \lambda_2 = -1$$

The homogenous solution is;

$$y_h(n) = C_1 (-1)^n u(n) + C_2 (4)^n u(n)$$

Since 4 is a characteristic root & the excitation is;

$$x(n) = 4^n u(n)$$

We assume a particular solution of the form

$$y_p(n) = kn 4^n u(n)$$

Then, ~~the~~

$$k_n 4^n u(n) - 3k(n-1)4^{n-1}u(n-1) - 4k(n-2)4^{n-2}u(n-2) \\ = 4^n u(n) + 2(4)^{n-1}u(n-1)$$

For $n=2$

$$k(32-12) = 4^2 + 8 = 24$$

$$\Rightarrow k = \frac{6}{5}$$

The total solution is;

$$y(n) = y_p(n) + y_h(n)$$

$$= \left[\frac{6}{5}n4^n + c_1 4 + c_2 (-1)^n \right] u(n)$$

To solve for c_1 & c_2 we assume that

$$y(-1) = y(-2) = 0 \text{ then,}$$

$$y(0) = 1 \text{ \&}$$

$$y(1) = 3y(0) + 4 + 2 = 9$$

$$\text{Hence, } c_1 + c_2 = 1 \text{ \&}$$

$$\frac{24}{5} + 4c_1 - c_2 = 9$$

$$4c_1 - c_2 = \frac{21}{5}$$

Therefore,

$$c_1 = \frac{26}{25} \text{ \& } c_2 = -\frac{1}{25}$$

The total solution is;

$$y(n) = \left[\frac{6}{5}n4^n + \frac{26}{25}4^n - \frac{1}{25}(-1)^n \right] u(n)$$

Q 1:- (Part-b):-

* Solution:-

$$y(n) = 0.6y(n-1) - 0.8y(n-2) + x(n)$$

Consider the different equation.

$$y(n) = 0.6y(n-1) - 0.8y(n-2) + x(n)$$

$$y(n) = 0$$

$$y(n) - 0.6y(n-1) + 0.8y(n-2) = x(n)$$

To obtain the homogeneous equation

set input $x(n) = 0$ then,

$$y(n) - 0.6y(n-1) + 0.8y(n-2) = 0$$

Determine the solution to the homogeneous equation

$$y_n(n) = \lambda^n$$

$$\lambda^n - 0.6\lambda^{n-1} + 0.08\lambda^{n-2} = 0$$

$$\lambda^{n-2}(\lambda^2 - 0.6\lambda + 0.08) = 0$$

$$\lambda^2 - 0.6\lambda + 0.08 = 0$$

$$(\lambda - 0.2)(\lambda - 0.4) = 0$$

So, the roots are

$$\lambda_1 = 0.2 \quad \& \quad \lambda_2 = 0.4$$

Thus, the general form of solution to the homogeneous equation is

$$y_n(n) = C_1(\lambda_1)^n + C_2(\lambda_2)^n$$

$$y(n) = C_1(0.2)^n + C_2(0.4)^n$$

$$\lambda = 0.2, \quad \lambda = 0.4$$

$$\text{Hence, } y_n(n) = C_1\left(\frac{1}{5}\right)^n + C_2\left(\frac{2}{5}\right)^n$$

With $x(n) = \delta(n)$, the initial conditions are;

$$y(0) = 1$$

$$y(1) - 0.6y(0) = 0$$

$$y(1) = 0.6$$

$$\text{Hence } c_1 + c_2 = 1 \quad \{$$

$$\frac{1}{5}c_1 + \frac{2}{5}c_2 = 0.6$$

$$\text{So, } c_1 = -1, \quad c_2 = 2$$

$$\text{Therefore, } h(n) = \left[-\left(\frac{1}{5}\right)^n + 2\left(\frac{2}{5}\right)^n \right] u(n)$$

The step response is;

$$s(n) = \sum_{k=0}^n h(n-k), \quad n \geq 0$$

$$= \sum_{k=0}^n \left[2\left(\frac{2}{5}\right)^{n-k} - \left(\frac{1}{5}\right)^{n-k} \right]$$

$$= \left\{ \frac{1}{0.12} \left[\frac{2}{5}^{n+1} - 1 \right] - \frac{1}{0.16} \left[\frac{1}{5}^{n+1} - 1 \right] \right\} u(n)$$



~~Q18~~

Q 2:- (Part-a)

* Solution:-

The z-transform is;

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

The expression can be written as;

$$\begin{aligned} & \frac{1}{\left(1-\frac{z}{z}\right)\left(1-\frac{1}{z}\right)^2} \\ &= \frac{1}{\left(\frac{z-2}{z}\right)\left(\frac{z-1}{z}\right)^2} \\ &= \frac{1}{\frac{(z-2)(z-1)^2}{z^3}} \\ &= \frac{z^3}{(z-2)(z-1)^2} \rightarrow \textcircled{1} \end{aligned}$$

$X(z)$ has a simple pole at $P_1=2$ & a double $P_2=P_3=1$. In such a case the appropriate partial-fraction expression is

$$X(z) = \frac{z^3}{(z-2)(z-1)^2} = \frac{A_1}{(z-2)} + \frac{A_2}{(z-1)} + \frac{A_3}{(z-1)^2}$$

Now determine the coefficient A_1, A_2, A_3 we proved that as in the case of distinct pole to determine A_1 , we multiply both sides of above eq by $(z-2)$

$$(z-2)X(z) = A_1 + \frac{z-2}{z-1} A_2 + \frac{z-2}{(z-1)^2} A_3$$

which evaluated at $z=2$

$$A_1 = \frac{(z-2)X(z)}{z} \Big|_{z=2} \quad \therefore z=2$$

$$A_1 = 4$$

$$A_2 = A_1 + \frac{z-2}{z-1}$$

$$A_2 = -3$$

$$A_3 = A_2 + \frac{z-2}{z-1} A_2$$

$$A_3 = -1$$

$$\text{Hence } x(n) = [4(2)^n - 3 - n]u(n)$$



Q2:- (Part-b)

Solution:-

First we eliminate the negative power by multiplying both numerator & denominator by z^2 then,

$$X(z) = \frac{z^2}{z^2 - 1.5z + 0.5}$$

The poles $X(z)$ are $P_1 = 1$ & $P_2 = 0.5$

Consequently, the expansion of the form

$$\frac{X(z)}{z} = \frac{z}{(z-1)(z-0.5)} = \frac{A_1}{z-1} + \frac{A_2}{z-0.5}$$

To determine A_1 & A_2 then, multiply the equation by the denominator term

$(z-1)(z-0.5)$ thus, we obtain

$$z = (z-0.5)A_1 + (z-1)A_2 \longrightarrow \textcircled{1}$$

Now if we set $z=p_1=1$ in eq $\textcircled{1}$
& we eliminate the term involving A_2
hence, $1 = (1-0.5)A_1$

Thus, we obtain the result $A_1=2$

& $z=p_2=0.5$ then, eliminating the term involving A_1 so we have

$$0.5 = (0.5-1)A_2$$

& hence $A_2 = -1$

Therefore, the result of the partial fraction expansion is;

$$\frac{X(z)}{(z)} = \frac{z}{z-1} - \frac{1}{z-0.5}$$

Q 3:- (Part -a)

★ Solution:-

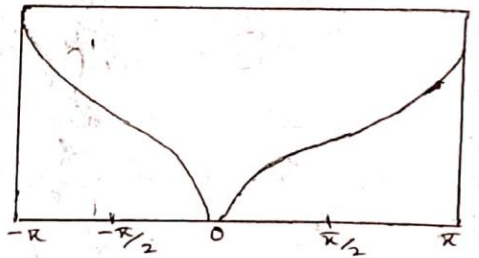
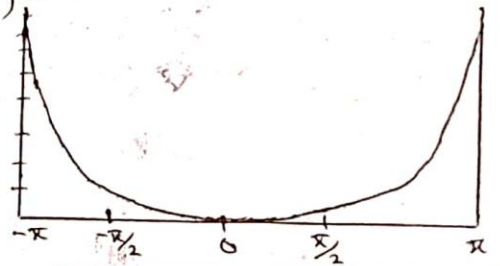
At $\omega = 0$ we have

$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

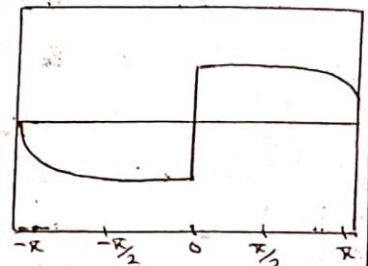
Hence $b_0 = (1-p)^2$

At $\omega = \frac{\pi}{4}$

$$\begin{aligned} H\left(\frac{\pi}{4}\right) &= \frac{(1-p)^2}{(1-pe^{-j\pi/4})^2} \\ &= \frac{(1-p)^2}{(1-p\cos(\pi/4) + jp\sin(\pi/4))^2} \\ &= \frac{(1-p)^2}{(1-p/\sqrt{2} + jp/\sqrt{2})^2} \end{aligned}$$



Hence $\frac{(1-p)^4}{[(1-p/\sqrt{2})^2 + p^2/2]} = \frac{1}{2}$



or equivalently

$$\sqrt{2}(1-p)^2 = 1 + p^2 - \sqrt{2}p$$

The value of $p = 0.32$ satisfies this equation. Consequently, the system function for the desired filter is

$$H(z) = \frac{0.46}{(1 - 0.32z^{-1})^2}$$

The same principles can be applied for the design of bandpass filters.

Q 3:- (Part-b):-

* Solution:-

Clearly, the filter must have poles at

$$P_{1,2} = re^{\pm j\pi/2}$$

& zeros at $z=1$ & $z=-1$

Consequently the system function is

$$\begin{aligned} H(z) &= G_1 \frac{(z-1)(z+1)}{(z-jr)(z+jr)} \\ &= G_1 \frac{z^2-1}{z^2+r^2} \end{aligned}$$

The gain factor is determined by evaluating the frequency response $H(\omega)$ of the filter at $\omega = \pi/2$

Thus we have

$$H(\pi/2) = G_1 \frac{2}{1-r^2} = 1$$

$$G_1 = \frac{1-r^2}{2}$$

~~The~~

The value of r is determined by evaluating $H(\omega)$ at $\omega = 4\pi/9$. Thus we have

$$\left| H\left(\frac{4\pi}{9}\right) \right|^2 = \frac{(1-r^2)^2}{4} \frac{2-2\cos(8\pi/9)}{2+r^4+2r^2\cos(8\pi/9)} = \frac{1}{2}$$

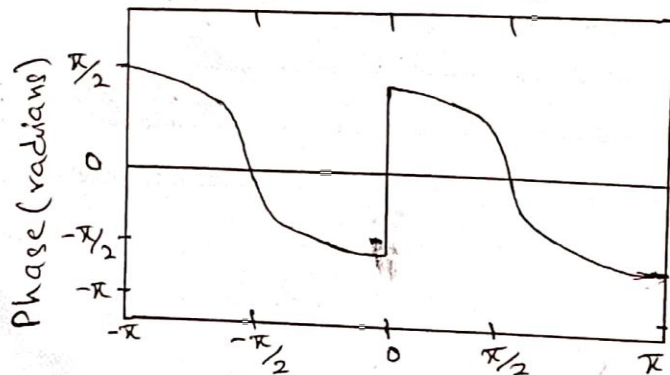
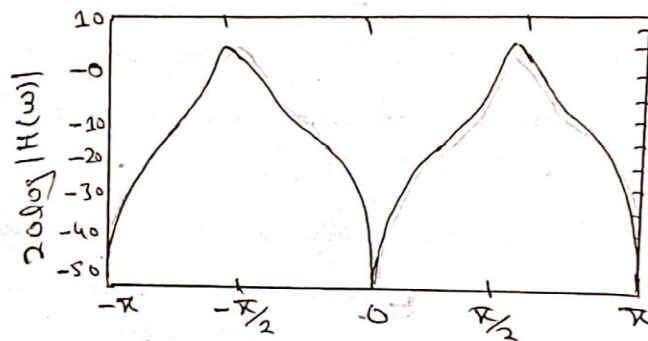
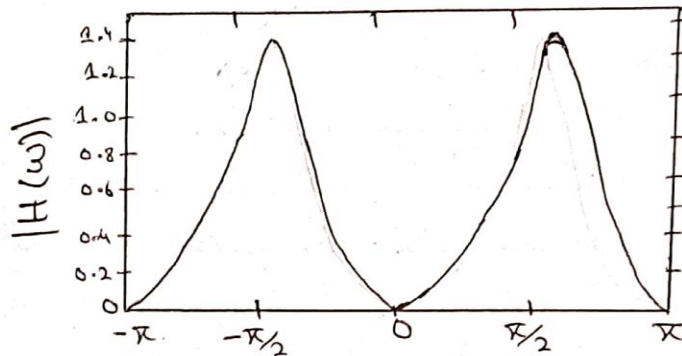
or equivalently,

$$1.94(1-r^2)^2 = 1 - 1.88r^2 + r^4$$

The value of $r^2 = 0.7$ satisfies this equation. Therefore, the system function for the desired filter is

$$H(z) = 0.15 \frac{1 - z^{-2}}{1 + 0.7z^{-2}}$$

Its frequency response is illustrated in below figure.



Q4:- (Part-a)

★ Solution:-

The Fourier transform of this sequence ~~for~~ is;

$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}}$$

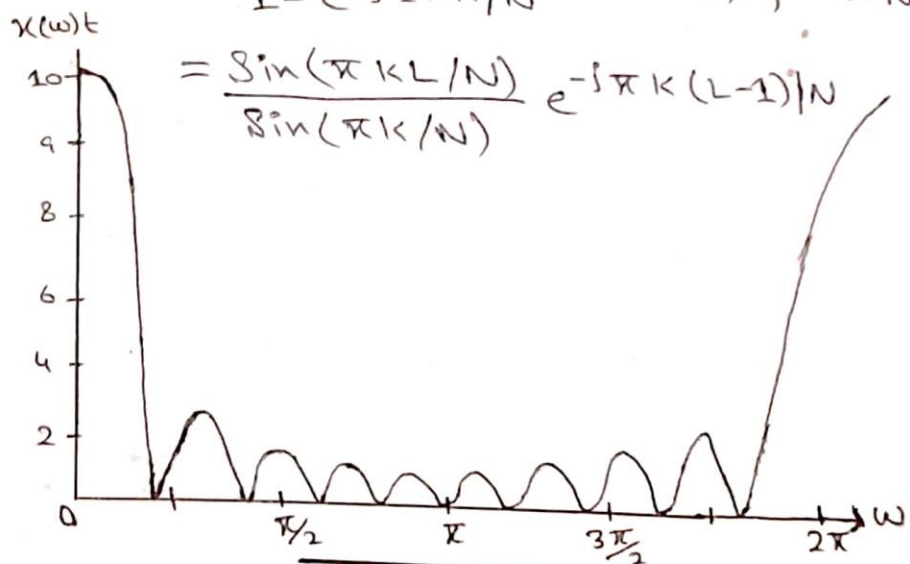
$$= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}$$

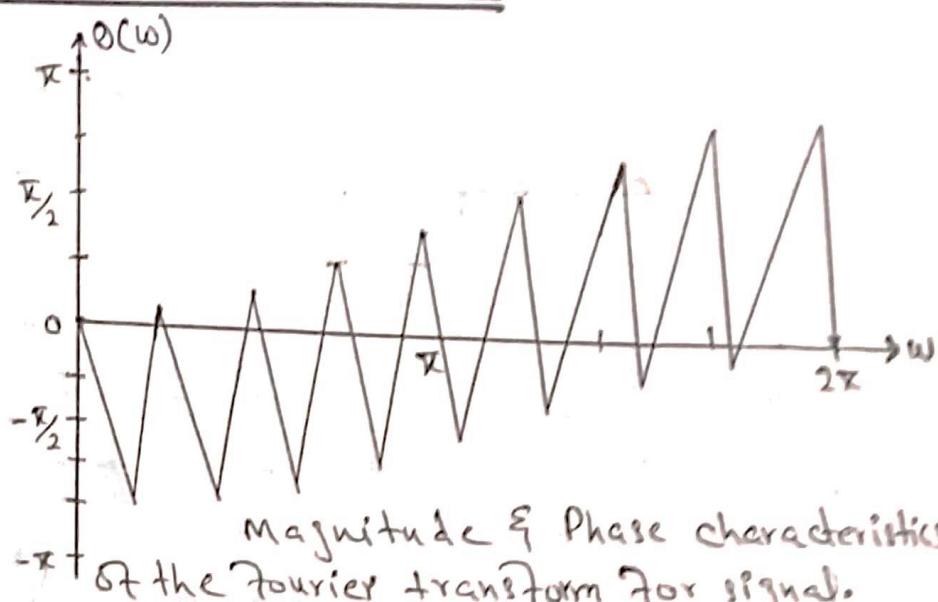
The magnitude & phase of $X(\omega)$ are illustrated in below figure for $L=10$. The N -Point DFT of $x(n)$ is simply $X(\omega)$ evaluated at the set of N equally spaced frequencies

$\omega_k = 2\pi k/N$, $k=0, 1, \dots, N-1$ Hence

$$X(k) = \frac{1 - e^{-j2\pi kL/N}}{1 - e^{-j2\pi k/N}} \quad k=0, 1, \dots, N-1$$

$$= \frac{\sin(\pi kL/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$





If N is selected such that $N=L$ then the DFT becomes

$$X(k) \begin{cases} L, & k=0 \\ 0, & k=1, 2, \dots, L-1 \end{cases}$$

Thus there is only non-zero value in DFT. This is apparent from observation of $x(\omega)$, since $x(\omega) = 0$ at the frequencies $\omega_k = 2\pi k/L$, $k \neq 0$. The reader should verify that $x(n)$ can be recovered from $X(k)$ by performing an L -point IDFT.

Q 4:- (Part - b)

* Solution:-

The first step is to determine the matrix W_4 . By exploiting the periodicity property of W_N & the symmetry property

$$W_N^{k+n/2} = -W_N^k$$

the matrix W_4 may be expressed as;

$$W_4 = \begin{bmatrix} W_4^0 & W_4^0 & W_4^0 & W_4^0 \\ W_4^0 & W_4^2 & W_4^2 & W_4^3 \\ W_4^0 & W_4^2 & W_4^4 & W_4^4 \\ W_4^0 & W_4^3 & W_4^6 & W_4^3 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^2 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^0 & W_4^2 \\ 1 & W_4^3 & W_4^2 & W_4^3 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$\text{Then, } y_4 = W_4 x_4 \begin{bmatrix} 6 \\ -2+2j \\ -2 \\ -2-2j \end{bmatrix}$$

The DFT of y_4 may be determined by conjugating the element in W_4 to obtain W_4^* & then applying the formula.