

**Department of Electrical Engineering**

**Final Exam Assignment**

**Date: 27/06/2020**

**Course Details**

<b>Course Title:</b>	Digital Signal Processing	<b>Module:</b>	6th
<b>Instructor:</b>		<b>Total Marks:</b>	50

**Student Details**

<b>Name:</b>	<b>Owais Afridi</b>	<b>Student ID:</b>	<b>13686</b>
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<b>Q1.</b>	(a)	<p>Determine the response <math>y(n)</math>, <math>n \geq 0</math>, of the system described by the second order difference equation</p> $y(n) - 4y(n - 1) + 4y(n - 2) = x(n) - x(n - 1)$ <p>To the input <math>x(n) = (-1)^n u(n)</math>. And the initial conditions are <math>y(-1) = y(-2) = 0</math>.</p>	<b>Marks</b> <b>7</b>
			<b>CLO</b> <b>2</b>
	(b)	<p>Determine the impulse response and unit step response of the systems described by the difference equation.</p> $y(n) - 0.7y(n - 1) + 0.1y(n - 2) = 2x(n) - x(n - 2)$	<b>Marks</b> <b>7</b>
			<b>CLO</b> <b>2</b>
<b>Q2.</b>	(a)	<p>Determine the causal signal <math>x(n)</math> having the z-transform</p> $X(z) = \frac{1}{(1 - 2z^{-1})(1 - z^{-1})^2}$ <p>(Hint: Take inverse z-transform using partial fraction method)</p>	<b>Marks</b> <b>6</b>
			<b>CLO</b> <b>2</b>
	(b)	<p>Evaluate the inverse z- transform using the complex inversion integral</p> $X(z) = \frac{1}{1 - az^{-1}} \quad  z  >  a $	<b>Marks</b> <b>6</b>
			<b>CLO</b> <b>2</b>
<b>Q3</b>	(a)	<p>A two- pole low pass filter has the system response</p> $H(z) = \frac{b_o}{(1 - pz^{-1})^2}$ <p>Determine the values of <math>b_o</math> and <math>p</math> such that the frequency response <math>H(\omega)</math> satisfies the condition <math>H(0) = 1</math> and <math> H(\frac{\pi}{4}) ^2 = \frac{1}{2}</math>.</p>	<b>Marks</b> <b>6</b>
			<b>CLO</b> <b>3</b>

	(b)	Design a two-pole bandpass filter that has the center of its passband at $\omega = \pi/2$ , zero in its frequency response characteristics at $\omega = 0$ and $\omega = \pi$ and its magnitude response in $\frac{1}{\sqrt{2}}$ at $\omega = 4\pi/9$ .	Marks 6
			CLO 3
	(a)	A finite duration sequence of Length L is given as $x(n) = \begin{cases} 1, & 0 \leq n \leq L - 1 \\ 0, & \text{otherwise} \end{cases}$ Determine the N- point DFT of this sequence for $N \geq L$	Marks 6
			CLO 2
Q 4	(b)	Perform the circular convolution of the following two sequences. Solve the problem step by step $x_1(n) = \begin{cases} 2 \\ \uparrow \end{cases}, 1, 2, 1$ $x_2(n) = \begin{cases} 1 \\ \uparrow \end{cases}, 2, 3, 4$	Marks 6
			CLO 2

①

Owais Afridi

ID #13686

Question # 01

$$\text{Part (a)} \quad y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$$

To the input  $x(n) = (-1)^n u(n)$ . Initial Conditions are  $y(-1) = y(-2) = 0$

Solution:

The characteristics equation is

$$\lambda^2 - 4\lambda + 4 = 0$$

$$\lambda = 2, 2 \text{ hence}$$

$$y_h(n) = c_1 2^n + c_2 n 2^n$$

The particular solution is

Substituting this solution into the difference eq, we obtain

$$K(-1)^n u(n) - 4K(-1)^{n-1} u(n-1) + 4K(-1)^{n-2} u(n-2) = (-1)^n u(n) - (-1)^{n-1} u(n-1)$$

$$\text{For } n=2, K(1+4+4) = 2$$

$$K(a) = 2$$

$$K = \frac{2}{9}$$

The total solution is

$$y[n] = \left[ c_1 2^n + c_2 n 2^n + \frac{2}{9} (-1)^n \right] u(n)$$

②

From the initial condition we obtain

$$y(0) = 1, \quad y(1) = 2 \text{ then}$$

$$c_1 = \frac{2}{9} = 1$$

$$c_1 = \frac{2}{9} - 1$$

$$= \frac{9-2}{9}$$

$$\boxed{c_1 = \frac{7}{9}}$$

$$2c_1 + 2c_2 - \frac{2}{9} = 2$$

$$2\left(\frac{7}{9}\right) + 2c_2 = 2 + \frac{2}{9}$$

$$+2c_2 = \frac{20}{9} - \frac{14}{9}$$

$$2c_2 = \frac{6}{9}$$

$$c_2 = \frac{2}{3} \times 2$$

$$\boxed{c_2 = \frac{4}{3}}$$

③

Question # 1

Part (B): Determine the impulse response and unit step response

$$y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$$

Solution:

The characteristic is

$$\lambda^n - 0.7\lambda^{n-1} + 0.1\lambda^{n-2} = 0.$$

$$\lambda^{2n} [\lambda^2 - 0.7\lambda + 0.1] = 0.$$

$$\lambda^2 - 0.7\lambda + 0.1 = 0$$

$$\lambda^2 - 0.5\lambda - 0.2\lambda + 0.1 = 0.$$

$$\lambda(\lambda - 0.5) - 0.1(\lambda - 0.5) = 0.$$

$$(\lambda - 0.5)(\lambda - 0.1) = 0$$

$$\lambda = 0.5, \lambda = 0.1.$$

General form of the solution to be Homogenous equation

$$y_n(n) = C_1 \lambda_1^n + C_2 \lambda_2^n$$

$$y_n(n) = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{1}{5}\right)^n$$

$$y(0) = 2$$

$$y_2(1) = 0.7y(0) = 0$$

$$y(1) = 1.4$$

④

Hence  $C_1 + C_2 = 2$  and impulse response:

$$\begin{aligned}\frac{1}{2}C_1 + \frac{1}{5}C_2 &= 1.4 \\ &= \frac{7}{5}\end{aligned}$$

$$C_1 + \frac{2}{5}C_2 = \frac{14}{5}$$

These equations yield

$$C_1 = \frac{10}{3}$$

$$C_2 = -\frac{4}{3}$$

$$h(n) = \left[ \frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

Step Response is:

$$S(n) = \sum_{k=0}^n h(n-k) \Rightarrow \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$= \frac{10}{3} \left[ \left(\frac{1}{2}\right)^n (2^{n+1} - 1) \right] u(n) - \frac{4}{3} \left[ \left(\frac{1}{5}\right)^n (5^{n+1} - 1) \right] u(n)$$

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Question # 02

Part (a) Determine the Causal Signal  $x[n]$  having  $z$ -transform.

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

Solution:  $X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$

= By partial fraction method:

$$\frac{1}{(1-2z^{-1})(1-z^{-1})^2} = \frac{A}{(1-2z^{-1})} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$\frac{1}{(1-2z^{-1})(1-z^{-1})^2} = \frac{A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1})}{(1-2z^{-1})(1-z^{-1})^2}$$

$$1 = A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1}) \quad \text{--- (1)}$$

Put  $z = 1$

$$1 = A(1-0)^2 + B(1-2)(1-0) + C(0)(1-2)$$

$$1 = 0 + 0 - C$$

$$1 = -C$$

$$\boxed{C = -1}$$

⑥

Put  $z=2$  in eq (1)

$$1 = A\left(1-\frac{1}{2}\right)^2 + B\left(1-\frac{2}{2}\right)\left(1-\frac{1}{2}\right) + C\left(\frac{1}{2}\right)\left(1-\frac{2}{2}\right)$$

$$1 = A\left(\frac{1}{2}\right)^2 + B(1-1)\left(\frac{1}{2}\right) + C\left(\frac{1}{2}\right)(1-1)$$

$$1 = \frac{A}{4} + B(0)\left(\frac{1}{2}\right) + C\left(\frac{1}{2}\right)(0)$$

$$1 = \frac{A}{4} + 0 + 0$$

$$\boxed{A=4}$$

Put  $z=3$  in eq (1)

$$1 = A\left(1-\frac{1}{3}\right)^2 + B\left(1-\frac{2}{3}\right)\left(1-\frac{1}{3}\right) + C\left(\frac{1}{3}\right)\left(1-\frac{2}{3}\right)$$

$$1 = A\left(\frac{4}{9}\right) + B\left(\frac{1}{3}\right)\left(\frac{2}{3}\right) + C\left(\frac{1}{3}\right)\left(\frac{1}{3}\right)$$

$$1 = \frac{4}{9}A + \frac{2}{9}B + \frac{1}{9}C$$

$$1 = \frac{4}{9}(4) + \frac{2}{9}B - \frac{1}{9}$$

$$1 + \frac{1}{9} - \frac{16}{9} = \frac{2}{9}B$$

$$-\frac{6}{9} \times \frac{9}{2} = B$$

$$\boxed{-3=B}$$

Hence  $x(n) = [4(2)^n - 3 - n]u(n)$ .



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## Question # 02

Part(b): Evaluate the inverse z-transform

$$X(z) = \frac{1}{1 - az^{-1}} \quad |z| > |a|$$

Solution:

We have

$$\begin{aligned} x(n) &= \frac{1}{2\pi j} \oint_C \frac{z^{n-1}}{1 - az^{-1}} dz \\ &= \frac{1}{2\pi j} \oint_C \frac{z^n dz}{z - a} \end{aligned}$$

where  $C$  is a circle at radius greater than  $|a|$ . We shall evaluate this integral using

$$\frac{1}{2\pi j} \oint_C \frac{f(z)}{z - z_0} dz = \begin{cases} f(z_0) & \text{if } z_0 \text{ is inside } C \\ 0 & \text{if } z_0 \text{ is outside } C \end{cases}$$

with  $f(z) = z^n$ , We distinguish two cases.

1. If  $n \geq 0$ ,  $f(z)$  has only zeros and hence no poles inside  $C$ . The only pole inside  $C$  is  $z = a$ , Hence

$$x(n) = f(z_0) = a^n \quad n \geq 0$$

(8)

a- If  $n < 0$ ,  $f(z) = z^n$  has an  $n$ th-order pole at  $z=0$ , which is also inside  $C$ . Thus there are contributions from both poles.

For  $n = -1$ , we have

$$\begin{aligned}x(-1) &= \frac{1}{2\pi j} \oint_C \frac{1}{z(z-a)} dz \\&= \frac{1}{z-a} \Big|_{z=0} + \frac{1}{z} \Big|_{z=a} \\&= \frac{1}{-a} + \frac{1}{a} \\&= 0\end{aligned}$$

If  $n = -2$ , we have

$$\begin{aligned}x(-2) &= \frac{1}{2\pi j} \oint_C \frac{1}{z^2(z-a)} dz \\&= \frac{d}{dz} \left( \frac{1}{(z-a)} \right) \Big|_{z=0} + \frac{1}{z^2} \Big|_{z=a} \\&= \frac{1}{-a^2} + \frac{1}{a^2} \\&= 0\end{aligned}$$

By continuing in the same way we can show that  $x(n) = 0$  for  $n < 0$ . Thus

$$x(n) = a^n u(n).$$

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### Question # 03

Part (a) A two-pole low pass filter has the System response

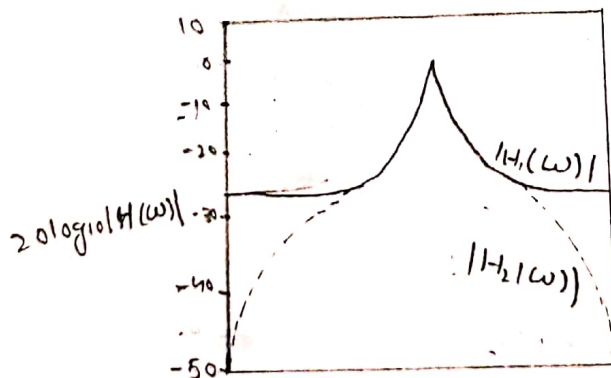
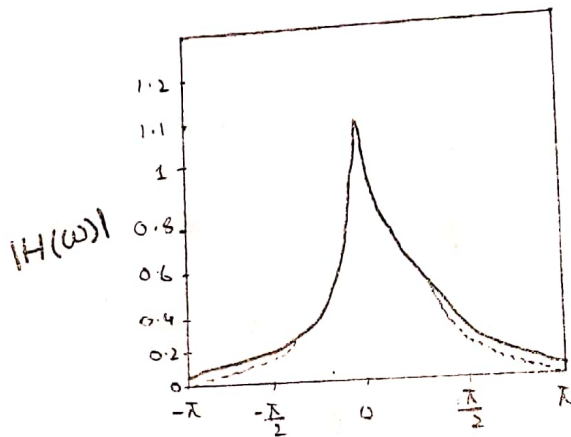
$$H(z) = \frac{b_0}{(1-pz^{-1})^2}$$

Determine the value of  $b_0$  &  $p$ , Condition

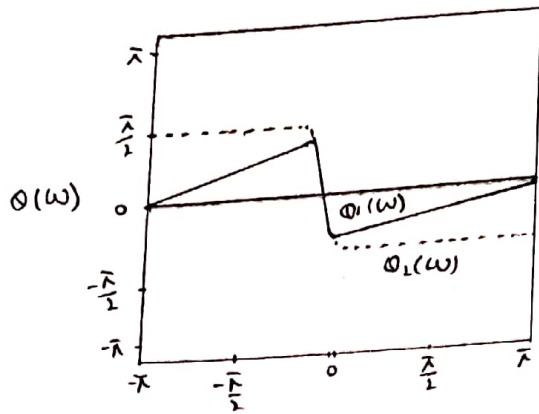
$$H(0) = 1 \text{ \& } |H(\frac{\pi}{4})|^2 = \frac{1}{2}$$

Solution:

Linear Time-Invariant Systems as Frequency-Selective filters



(10)



Magnitude of and phase response of (1) a single-pole filter and (2) a one-pole, one-zero filter:

$$H_1(z) = (1-a)/(1-az^{-1})$$

$$H_2(z) = \{(1-a)/2\} [1+z^{-1}/(1-az^{-1})]$$

$$a = 0.9$$

Now we have to determine the value of  $b_0$  of P. Such that the frequency response  $H(\omega)$  satisfies the conditions.

$$H(0) = 1 \quad \text{and} \quad |H(\frac{\pi}{4})|^2 = \frac{1}{2}$$

Sol At  $\omega = 0$

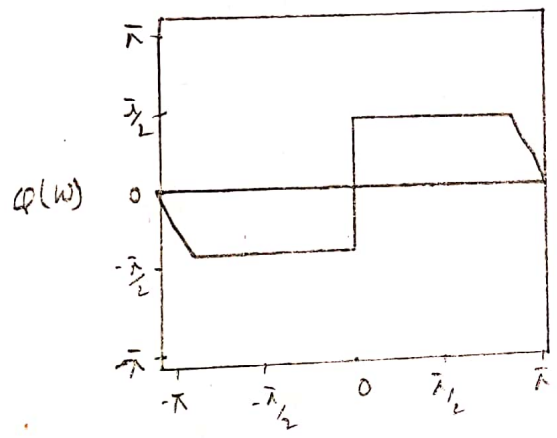
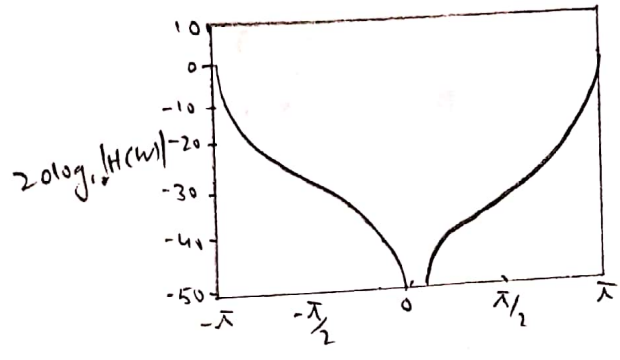
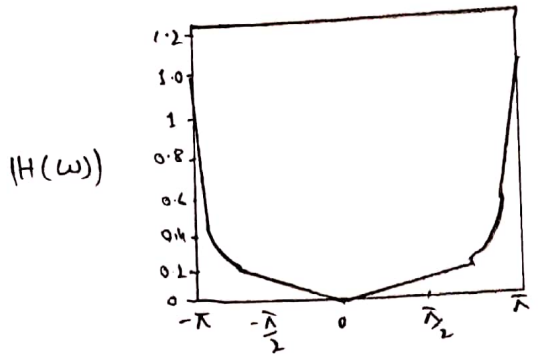
$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

Both sides multiply by  $(1-p)^2$

$$= \frac{b_0}{(1-p)^2} \times (1-p)^2 = 1 \times (1-p)^2$$

$$= b_0 = (1-p)^2$$

(11)



Magnitude & phase response of a simple highpass filter;

$$H(z) = \frac{(1-a)}{2} \left[ \frac{1-z^{-1}}{1+az^{-1}} \right]$$

$$a = 0.9$$

$$\text{at } \omega = \frac{\pi}{4}$$

$$H\left(\frac{\pi}{4}\right) = \frac{(1-p)^2}{(1-pe^{-j\pi/4})^2}$$

(12)

$$= \frac{(1-p)^2}{(1-p \cos(\frac{\pi}{4}) + jp \sin(\frac{\pi}{4}))^2}$$

$$= \frac{(1-p)^2}{\left(\frac{1-p}{\sqrt{2}} + \frac{jp}{\sqrt{2}}\right)^2}$$

= take square on above equation

$$= \frac{([1-p]^2)^2}{\left(\left[\frac{1-p}{\sqrt{2}} + \frac{jp}{\sqrt{2}}\right]^2\right)^2}$$

$$H(z) = \frac{[1-p]^4}{[(1-p/\sqrt{2})^2 + p^2/2]^2} = \frac{1}{2}$$

$$= \sqrt{2}(1-p)^2 = 1 + p^2 - \sqrt{2}p$$

the value of  $p=0.32$  satisfies this equation.  
desid. filter is:

$$H(z) = \frac{0.46}{(1-0.32z^{-1})^2}$$

(13)

Question #03

Part (B):

The filter must have poles at

$$p_{1,2} = re^{\pm j\frac{\pi}{2}}$$

and zeros at  $z=1$  &  $z=-1$

Consequently, the system function is

$$H(z) = G_1 \frac{(z-1)(z+1)}{(z-jr)(z+jr)}$$

The gain factor is determined by evaluating the frequency response  $H(\omega)$  of the filter at  $\omega = \frac{\pi}{2}$ . Thus we have

$$H\left(\frac{\pi}{2}\right) = G_1 \frac{2}{1-r^2} = 1$$

$$G_1 = \frac{1-r^2}{2}$$

The value is determined by evaluating  $H(\omega)$  at  $\omega = 4\pi/9$ . Thus

$$\left|H\left(\frac{4\pi}{9}\right)\right|^2 = \frac{(1-r^2)^2}{4} \cdot \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)} = \frac{1}{2}$$

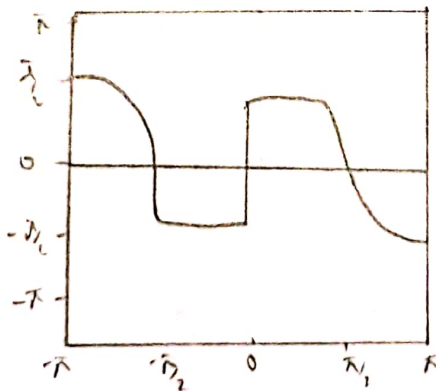
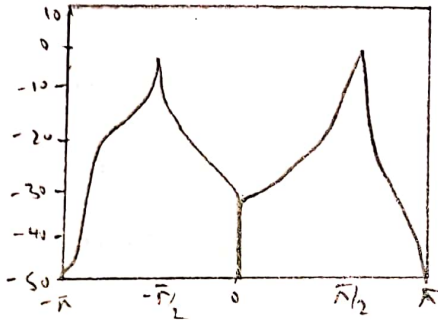
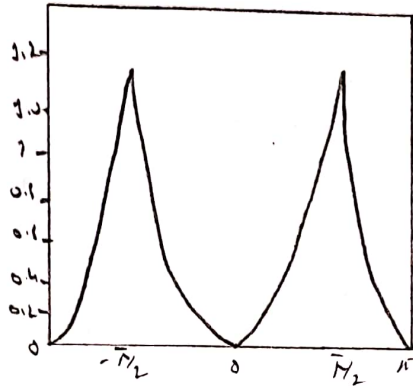
(14)

Evaluating

$$1.94(1-r^2)^2 = 1 - 1.88r^2 + r^2$$

The value of  $r^2 = 0.7$  satisfies this equation.

$$H(z) = 0.15 \frac{1-z^{-2}}{1+0.7z^{-2}}$$





(15)

Question # 04  $x(n) = \begin{cases} 1, & 0 \leq n \leq L-1 \\ 0, & \text{otherwise} \end{cases}$   
Part (a)

Solution: The Fourier transform of this sequence

$$\begin{aligned} X(\omega) &= \sum_{n=0}^{L-1} x(n) e^{-j\omega n} \\ &= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} \\ &= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2} \end{aligned}$$

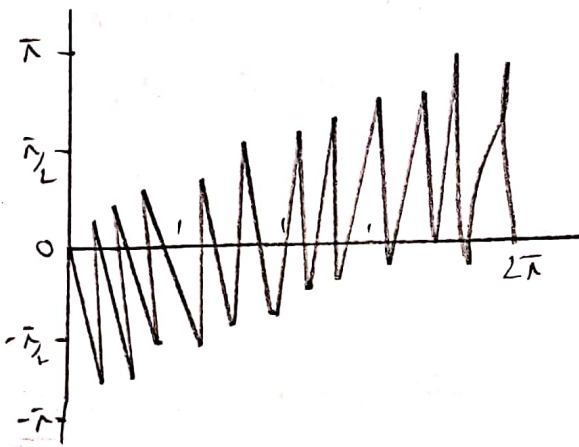
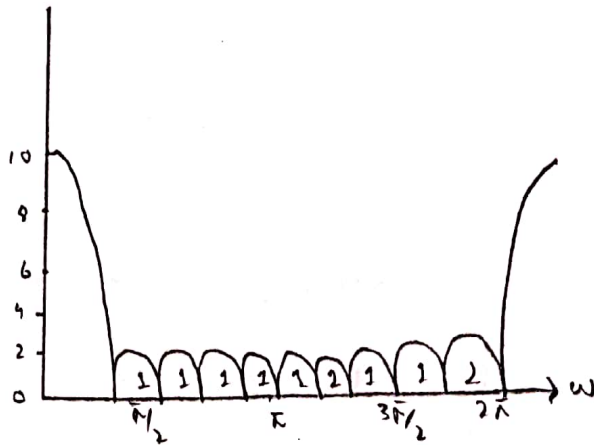
The magnitude & phase of  $X(\omega)$  are illustrated for  $L=10$ . The  $N$  point DFT of  $x(n)$  is simply  $X(\omega)$  evaluated at the set of  $N$  equally spaced frequencies  $\omega_k = \frac{2\pi k}{N}$

$$k = 0, 1, \dots, N-1$$

Hence

$$\begin{aligned} X(k) &= \frac{1 - e^{-j2\pi k L/N}}{1 - e^{-j2\pi k/N}} \quad k = 0, 1, N-1 \\ &= \frac{\sin(\pi k L/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N} \end{aligned}$$

(16)



If  $N$  is selected such that  $N=L$  then the DFT becomes

$$X(k) = \begin{cases} L & k=0 \\ 0 & k=1, 2, 3, \dots, L-1 \end{cases}$$

Thus there is one non-zero value in DFT this is apparent from observation of  $X(\omega)$ . Since  $X(\omega) = 0$  at the frequency.

$X(k) = \frac{2\pi k}{L}$   $k \neq 0$ . The reader should verify that

$x(n)$  can be recovered from

$X(k)$  by performing  $L$ -point IDFT.

(17)

Question # 04

Part (b):  $x_1(n) = \{2, 1, 2, 1\}$   $x_2(n) = \{1, 2, 3, 4\}$

Solution:

Each Sequence consist of four non-zero points for the purpose of illustrating the operation involved in circular convolution.

Now  $x_3(m)$  obtained by circularly convolving  $x_1(n)$  with  $x_2(n)$ .

With  $m=0$ , we have

$$x_3(0) = \sum_{n=0}^3 x_1(n) x_2(1-n) N$$

$x_2(1-n)$  is simply the sequence  $x_2(n)$  the folded sequence is simply  $x_2(n)$ .

The product sequence is obtained by multiplying  $x_1(n)$  with  $x_2(-n)_4$  point by point finally we sum the values

$$x_3(0) = 14$$

For  $m=1$  we have

$$x_3(1) = \sum_{n=0}^3 x_1(n) x_2[(1-n)]_4$$

It is easily verified  $x_2[(1-n)]_4$  is the simple sequence  $x_2[1-n]_4$  rotated counter clock wise by one unit in the time illustrated.

(18)

Finally, we sum the values:

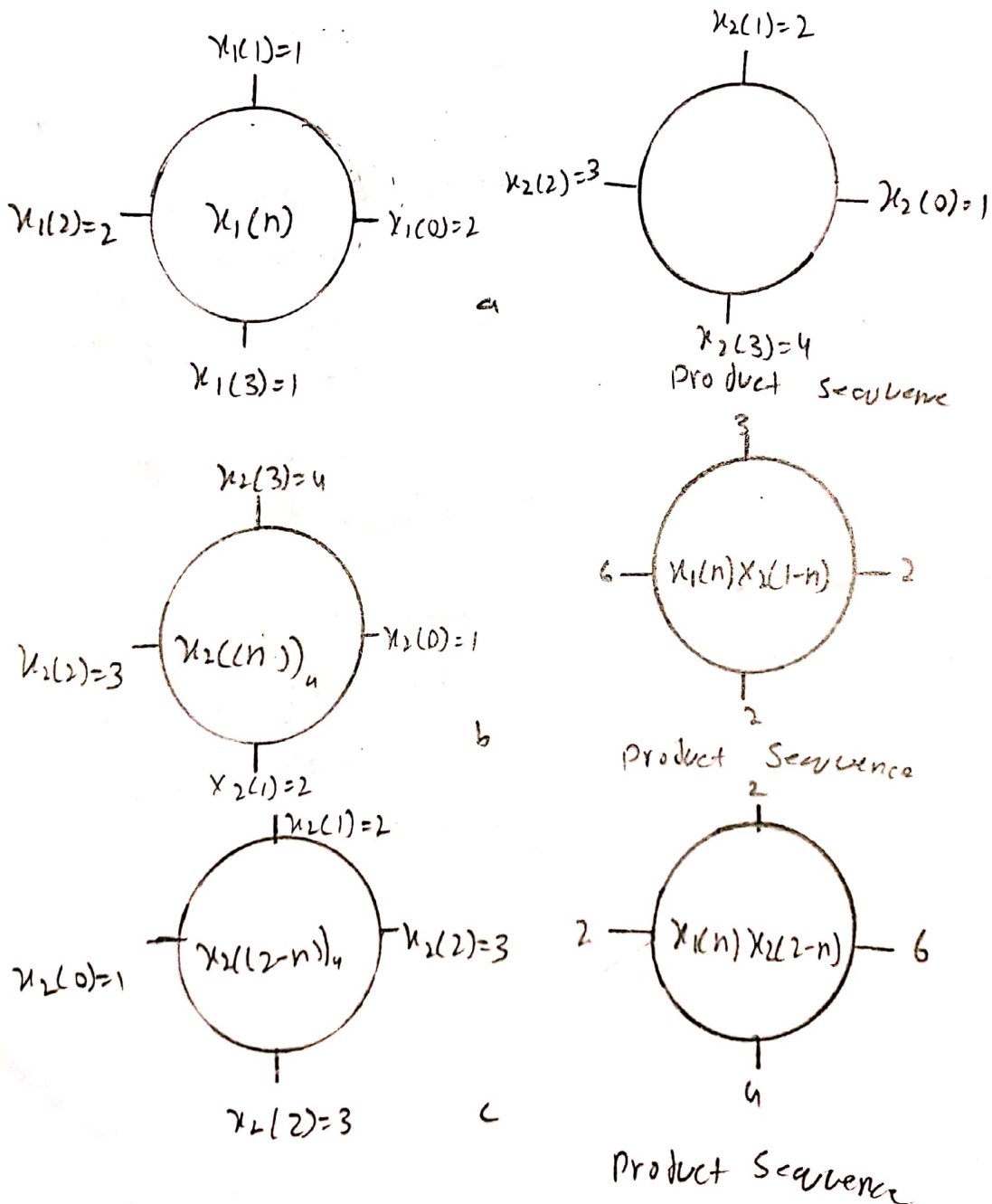
$$X_3(1) = 16$$

$m=2$  we have:

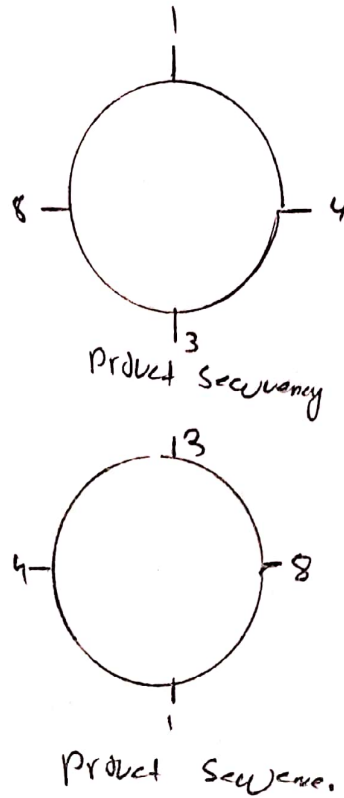
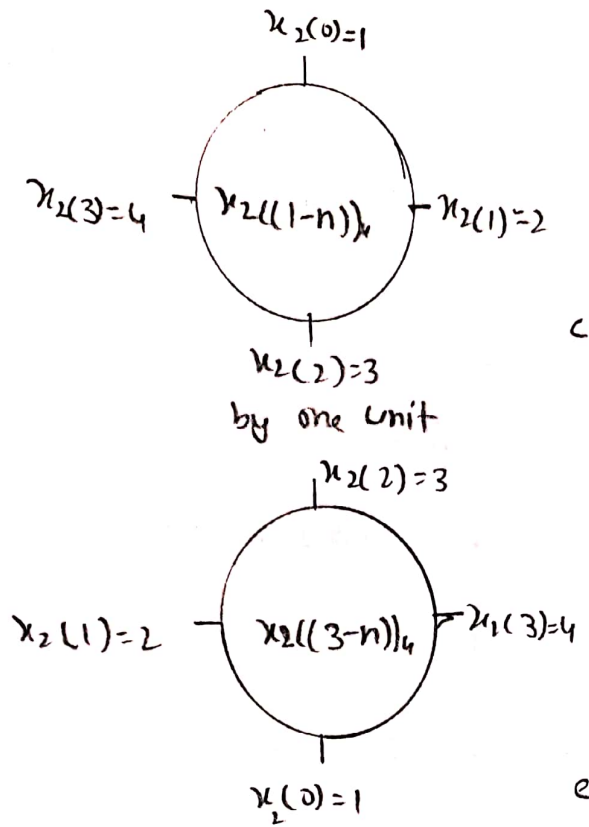
$$X_3(2) = \sum_{n=0}^3 X_1(n) X_2((2-n))_4$$

Now  $X_2((2-n))_4$  is folded sequence.

rotated two units of time in the counter clock wise direction.



(19)



Folde Sequence rotated by three units.  
 along with product Sequence  $x_1(n)x_2(2-n)_4$  By  
 Summing the four terms in the product Sequence.  
 $x_3(2) = 14$

$m=3$ , We have,

$$x_3(3) = \sum_{n=0}^3 x_1(n)x_2((3-n))_4$$

The folded Sequence  $x_2((n))_4$  is now three units in  
 the time to yield  $x_2((3-n))_4$  is multiplied by  $x_1(n)$  to  
 yield the product.

Sequence  $x_3(3) = 16$

(20)

We observe that if the computation above is continued beyond  $m=3$  we simply repeat the sequence of 4 values.

Therefore circular convolution of two sequences  $x_1(n)$  and  $x_2(n)$  yield sequence

$$x_3(n) = \{14, 16, 14, 16\}.$$