

DSP  $\Rightarrow$  (1)

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Q No (1) (Part a)

Solution:

$$y(n) - 4y(n-1) + 4y(n-2) = z(n) - z(n-1)$$

The Characteristic eq is

$$\lambda^2 - 4\lambda + 4 = 0$$

$\lambda = 2, 2$  Hence

$$y_h(n) = C_1 2^n + C_2 n 2^n$$

The Particular Solution is

$$y_p(n) = K(-1)^n u(n)$$

Substituting this eq in difference eq we get

$$\begin{aligned} K(-1)^n u(n) - 4K(-1)^{n-1} u(n-1) + 4K(-1)^{n-2} u(n-2) \\ = (-1)^n u(n) - (-1)^{n-1} u(n-1) \end{aligned}$$

$$\text{For } n=2, K(1+4+4) = 2 \Rightarrow K = \frac{2}{9}$$

Total solution is

$$y(n) = \left[ C_1 2^n + C_2 n 2^n + \frac{2}{9} (-1)^n \right] u(n)$$

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From the initial condition.

$$C_1 + \frac{2}{9} = 1$$

$$\Rightarrow C_1 = \frac{7}{9}$$

$$2C_1 + 2C_2 - \frac{2}{9} = 2$$

$$\Rightarrow C_2 = \frac{1}{3}$$

Q No 1 (b)

Solution:

$$h^n - 0.7h^{n-1} + 0.1h^{n-2} = 0$$

$$h^{n-2}(h^2 - 0.7h + 0.1) = 0$$

$$h^2 - 0.5 - 0.2h + 0.1 = 0$$

$$h(h - 0.5) - 0.1(h - 0.5) = 0$$

$$h = 0.5 \quad h = 0.1$$

General form of solution to be Homogeneous equation is

$$y_h(n) = C_1 h_1^n + C_2 h_2^n$$

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$$y_h(n) = c_1 \left(\frac{1}{2}\right)^n + c_2 \left(\frac{1}{5}\right)^n$$

with  $z_2(n) = y(n)$  we have

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0 \Rightarrow$$

$$y(1) = 1.4$$

Hence  $c_1 + c_2 = 2$  and

$$\frac{1}{2} c_1 + \frac{1}{5} c_2 = 1.4 = \frac{7}{5}$$

$$c_1 + \frac{2}{5} c_2 = \frac{14}{5}$$

These equations yield

$$c_1 = \frac{10}{3} \quad c_2 = -\frac{4}{3}$$

$$h(n) = \left[ \frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

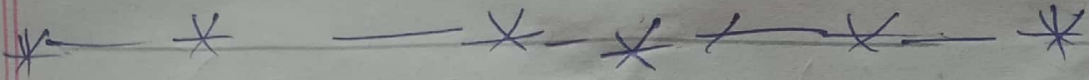
Step response is

$$S(n) = \sum_{k=0}^n h(n-k) \Rightarrow \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

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$$= \frac{10}{3} \left( \frac{1}{2}^n (2^{n+1} - 1) U(n) - \frac{1}{3} \right) \\ \left( \frac{1}{5}^n (5^{n+1} - 1) U(n) \right)$$



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Q No 2 (a)

Sol<sup>o</sup>

$$\frac{X(z)}{z} = \frac{z^2}{(2z-1)(z-1)^2}$$

$$\frac{X(z)}{z} = \frac{A}{2z-1} + \frac{B}{z-1} + \frac{C}{(z-1)^2}$$

Find A, B, C

$$A = 4$$

$$B = 3$$

$$C = -1$$

Hence

$$X(z) = [4(z)^n - 3 - n]u(n)$$

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Q No 2 (b)

Solution:

We have

$$\begin{aligned}x(n) &= \frac{1}{2\pi j} \oint_C \frac{z^{n-1}}{1-az^{-1}} dz \\ &= \frac{1}{2\pi j} \oint_C \frac{z^n dz}{z-a}\end{aligned}$$

Where  $C$  is a circle radius greater than  $|a|$ . we evaluate

$f(z) = z^n$ . We distinguish two cases.

Case (1)  $\Rightarrow$

if  $n \geq 0$ ,  $f(z)$  has only zeros and hence no poles inside  $C$ . The only pole inside  $C$  is

$$z = a$$

Hence

$$x(n) = f(z_0) = a^n \quad n \geq 0$$

Case (2)  $\Rightarrow$

if  $n < 0$ ,  $f(z) = z^n$  has an  $n$ th-order pole at  $z = 0$

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Having Contributors from both pole

For  $n = -1$  we have

$$x(-1) = \frac{1}{2\pi j} \oint_C \frac{1}{z(z-a)} dz$$
$$= \left. \frac{1}{z-a} \right|_{z=0} + \left. \frac{1}{z} \right|_{z=a} = 0$$

if  $n = -2$  we

have

$$x(-2) = \frac{1}{2\pi j} \oint_C \frac{1}{z^2(z-a)} dz$$
$$= \left. \frac{d}{dz} \left( \frac{1}{z-a} \right) \right|_{z=0} + \left. \frac{1}{z^2} \right|_{z=a}$$

By continuing same way we can show that

$x(n) = 0$  for  $n < 0$  thus

$$x(n) = a^n u(n)$$

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### Question No. 3

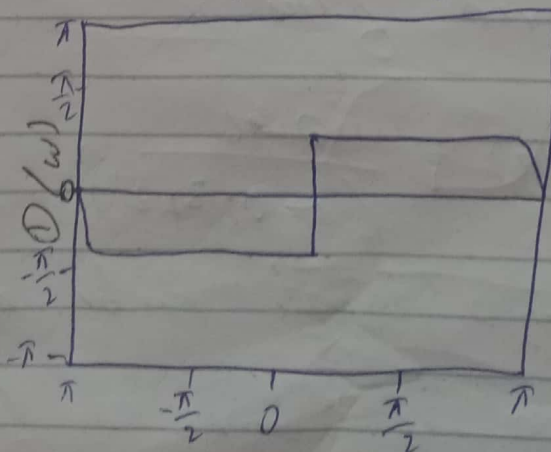
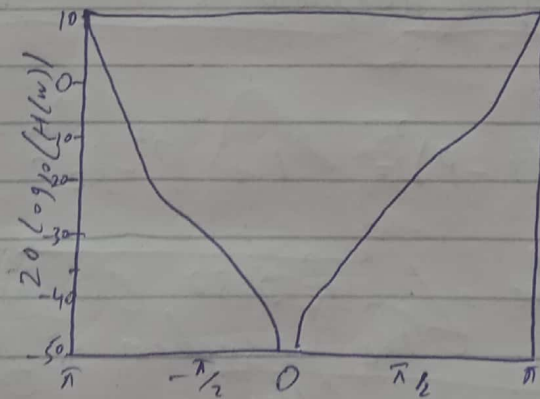
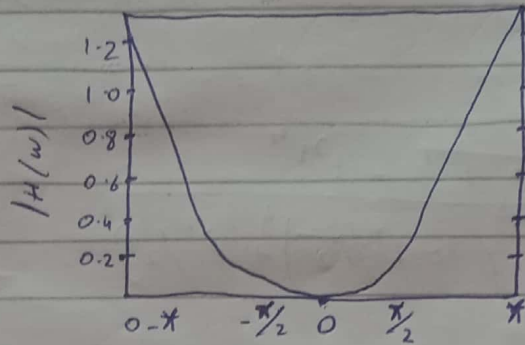
#### Part (a)

Solution:

At  $\omega = 0$  we have

$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

Hence  $b_0 = (1-p)^2$





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At  $\omega = \frac{\pi}{4}$

$$\begin{aligned} H\left(\frac{\pi}{4}\right) &= \frac{(1-P)^2}{(1 - P e^{-j\pi/4})^2} \\ &= \frac{(1-P)^2}{\left((1 - P \cos(\pi/4)) + jP \sin(\pi/4)\right)^2} \\ &= \frac{(1-P)^2}{\left(1 - P/\sqrt{2} + jP/\sqrt{2}\right)^2} \end{aligned}$$

Hence

$$\frac{(1-P)^4}{\left[1 - (1-P)/\sqrt{2} + P^2/2\right]^2}$$

or equivalently

$$\sqrt{2}(1-P)^2 = 1 + P^2 - \sqrt{2}P$$

The value of  $P = 0.32$  satisfy this equation. The system for desired filter is  $\frac{1}{1+z^{-1}}$

$$H(z) = \frac{0.46}{(1 - 0.32z^{-1})^2}$$

The same Principle can be applied for Design of band Pass Filter

## Part (b)

Clearly, Filter must have poles at  $P_1, z = \delta e^{j\pi/2}$  and zeros at  $z = 1$  and  $z = -1$ .

The system  $H(z)$  is

$$H(z) = G \frac{(z-1)(z+1)}{(z-j\delta)(z+j\delta)}$$

$$= G \frac{z^2 - 1}{z^2 - \delta^2}$$

The frequency response  $H(\omega)$  of filter at  $\omega = \pi/2$ .

Thus we have

$$H(\pi/2) = G \frac{2}{1 - \delta^2} = 1$$

$$G = \frac{1 - \delta^2}{2}$$

The value of  $\delta$  is determined by evaluating  $H(\omega)$  at  $\omega = 4\pi/9$ .

Thus we have

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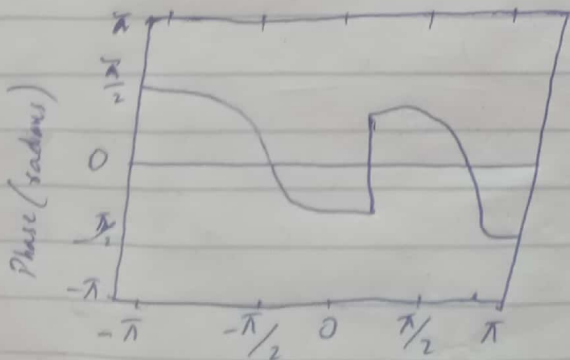
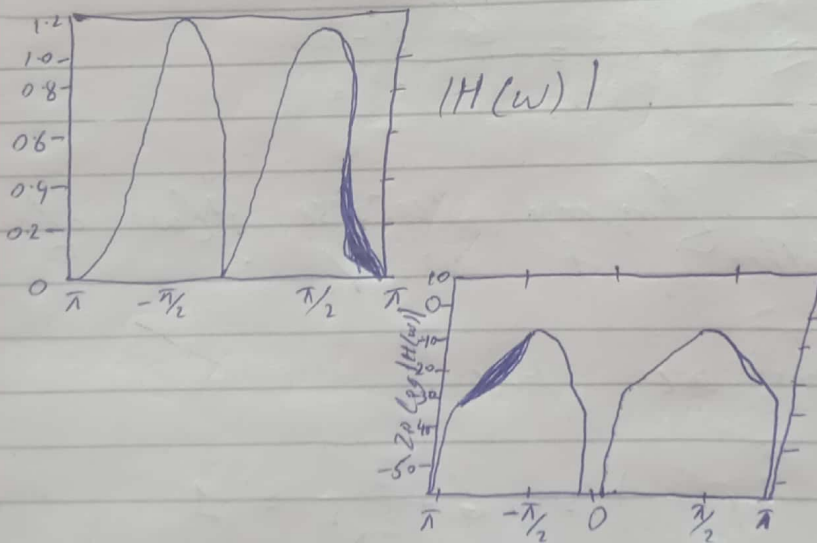
$$\left| H\left(\frac{4\pi}{9}\right) \right|^2 = \frac{(1-\delta^2)^2}{4} = \frac{2-2\cos(8\pi/9)}{1+1^4+2r^2\cos(8\pi/9)}$$
$$= \frac{1}{2}$$

$$0.8 \quad 1.94 (1-\delta^2)^2 = 1 - 1.818\delta^2 + \delta^4$$

Value of  $r^2 = 0$  satisfy equation.

Filter for for desired

$$\text{Filter is } H(z) = 0.15 \frac{1-z^{-2}}{1+0.7z^{-2}}$$



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Q No # 4 (a)

Sol<sup>n</sup>:-

The Fourier transform of the sequence is

$$\begin{aligned}
 X(\omega) &= \sum_{n=0}^{L-1} x(n) e^{-j\omega n} \\
 &= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} \\
 &= \frac{(\sin \omega L / 2)}{\sin(\omega / 2)} e^{-j\omega (L-1)/2}
 \end{aligned}$$

The magnitude and phase of  $X(\omega)$  are illustrated in for  $L=10$ . The  $N$ -point DFT of  $x(n)$  is simply  $X(\omega)$ .

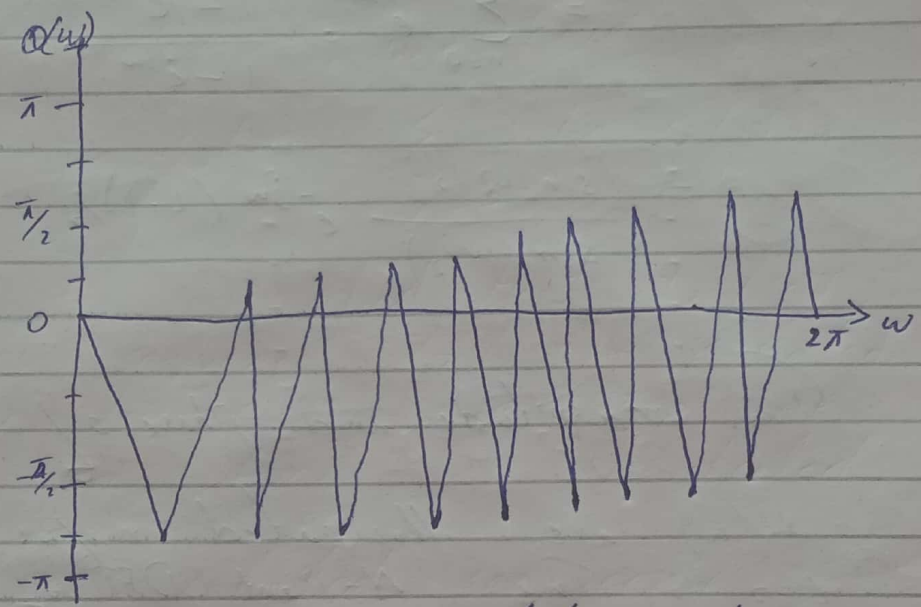
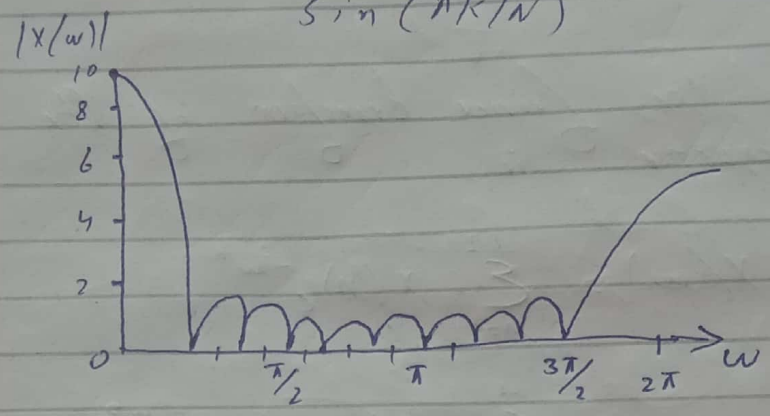
evaluated at the set of  $N$  equally spaced frequencies

$$\omega_k = 2\pi k / N \quad k = 0, 1, 2, \dots, N-1$$

Hence

$$X(k) = \frac{1 - e^{-j2\pi k L / N}}{1 - e^{-j2\pi k / N}} \quad k = 0, 1, \dots, N-1$$

$$z = \frac{\sin(\pi K L / N)}{\sin(\pi K / N)} e^{-j\pi K(L-1)/N}$$



if  $N$  is selected such that  $N=L$   
 then IDFT becomes

$$X(K) = \begin{cases} L & K=0 \\ 0 & K=1, 2, 3, \dots, L-1 \end{cases}$$

Thus there is only one <sup>non</sup> zero value in DFT.  
 This is apparent from  $X(w)$ . Since  $X(w)=0$   
 at frequencies  $w_k = 2\pi k/L$   $k \neq 0$ . The  
 reader should verify that

$x(n)$  can be recovered from  
 $X(K)$  by performing an 2-point  
 IDFT.

### Q No # 4 (b)

Solution:

Each Sequence consists of four non-zero points. For the purposes of illustrating the operation involved in circular convolution, it is desirable to graph each sequence as points on a circle.

Thus sequence  $x_1(n)$  and  $x_2(n)$  are graphed as illustrated in figure. We note that, the sequences are graphed in counter clock direction on circle. Now  $x_3(m)$  is obtained by circularly convolving  $x_1(n)$  and  $x_2(n)$  as specified by beginning with  $m=0$ .

We have

$$x_3(0) = \sum_{n=0}^3 v_1(n) x_2((L-n))_N$$

$x_2((L-n))_4$  is simply the sequence  $x_2(n)$  followed and graphed on circle.

The Product Sequence is obtained by multiplying  $x_1(n)$  with  $x_2(1-n)$ . Pointed in figure. Finally we add values to obtain.

$$x_3(0) = 14$$

for  $m=1$  we have

$$x_3(1) = \sum_{n=0}^3 x_1(n) x_2(1-n)$$

it is verified that  $x_2(1-n)$  is sequence  $x_2(1-n)$  rotated counter clock wise by one unit time as illustrated.

The rotated sequence multiplies  $x_1(n)$  to yield product sequence

$$x_3(1) = 16$$

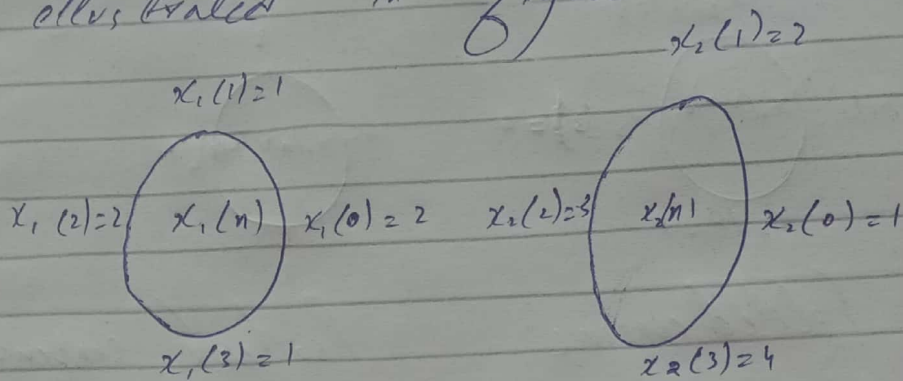
$m=2$  we have

$$x_3(2) = \sum_{n=0}^3 x_1(n) x_2(2-n)$$

Now  $x_2(2-n)$  is the folded sequence in fig. Rotated two units of time in counter clock wise direction

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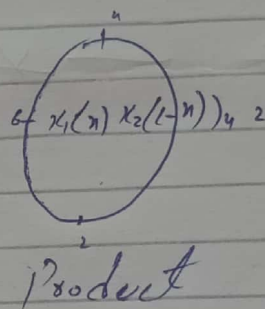
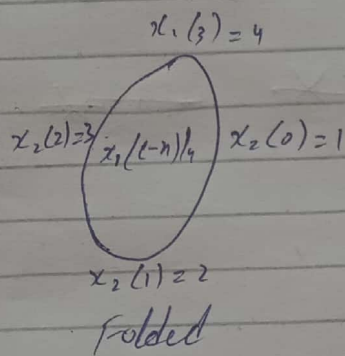
The Resultant sequence is illustrated in figure



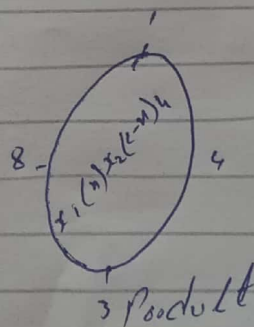
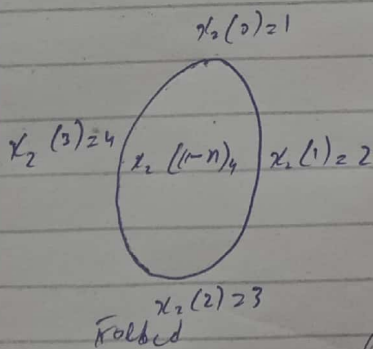
Folded sequence

Product seq

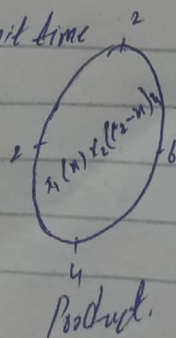
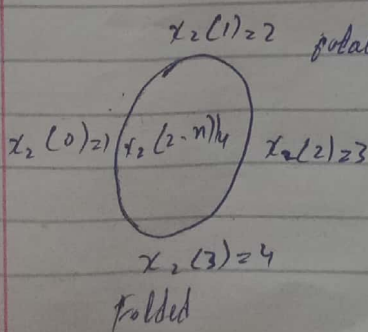
(a)



(b)



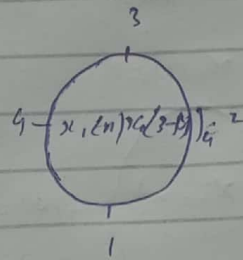
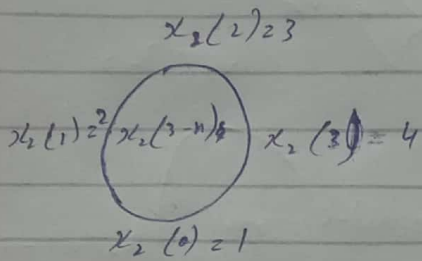
(c)





(17)

(d)



Folded sequence

Product sequence

Rotated by 3 units in time

