

Q1.	(a)	Determine the response $y(n)$, $n \geq 0$, of the system described by the second order difference equation $y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$ To the input $x(n) = (-1)^n u(n)$. And the initial conditions are $y(-1) = y(-2) = 0$.	Marks 7
			CLO 2
	(b)	Determine the impulse response and unit step response of the systems described by the difference equation. $y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$	Marks 7
			CLO 2
Q2.	(a)	Determine the causal signal $x(n)$ having the z -transform $X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$ (Hint: Take inverse z -transform using partial fraction method)	Marks 6
			CLO 2
	(b)	Evaluate the inverse z - transform using the complex inversion integral $X(z) = \frac{1}{1-az^{-1}} \quad z > a $	Marks 6
			CLO 2
Q3	(a)	A two- pole low pass filter has the system response $H(z) = \frac{b_0}{(1-pz^{-1})^2}$ Determine the values of b_0 and p such that the frequency response $H(\omega)$ satisfies the condition $H(0) = 1$ and $ H(\frac{\pi}{4}) ^2 = \frac{1}{2}$.	Marks 6
			CLO 3

	(b)	Design a two-pole bandpass filter that has the center of its passband at $\omega = \pi/2$, zero in its frequency response characteristics at $\omega = 0$ and $\omega = \pi$ and its magnitude response in $\frac{1}{\sqrt{2}}$ at $\omega = 4\pi/9$.	Marks 6
			CLO 3
Q4	(a)	A finite duration sequence of Length L is given as $x(n) = \begin{cases} 1, & 0 \leq n \leq L-1 \\ 0, & \text{otherwise} \end{cases}$ Determine the N - point DFT of this sequence for $N \geq L$.	Marks 6
			CLO 2
	(b)	Perform the circular convolution of the following two sequences. Solve the problem step by step $x_1(n) = \left\{ \frac{2}{7}, 1, 2, 1 \right\}$ $x_2(n) = \left\{ \frac{1}{7}, 2, 3, 4 \right\}$	Marks 6
			CLO 2

(1)

Q 1

Part a

$$y[n] - 4y[n-1] + 4y[n-2] = 2^n - 2^{n-1}$$

Soll:

$$y(n) - 4y(n-1) + 4y(n-2) = 2^n - 2^{n-1}$$

The characteristic equation is

$$\lambda^2 - 4\lambda + 4 = 0$$

$$\lambda = 2, 2 \text{ Hence}$$

$$y_n = C_1 2^n + C_2 n 2^n$$

The particular solution is

$$y_p(n) = k(-1)^n u(n)$$

Substituting this solution in to
different equation we obtain

$$\begin{aligned} k(-1)^n u(n) - 4k(-1)^{n-1} u(n-1) + 4k(-1)^{n-2} \\ u(n-2) \\ = (-1)^n - (-1)^{n-1} u(n-1) \end{aligned}$$

for $n=2$

$$k(1+4+4) = 2 \Rightarrow k = \frac{2}{9}$$

the total solution is

$$y(n) = \left[C_1 2^n + C_2 n 2^n + \frac{2(-1)^n}{9} \right] u(n)$$

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From the Initial condition
we obtain $y(0) = 1$, $y(1) = 2$ then

$$c_1 + \frac{2}{9} = 1$$

$$\Rightarrow c_1 = \frac{7}{9}$$

$$2c_1 + 2c_2 - \frac{2}{9} = 2$$

$$\Rightarrow c_2 = \frac{1}{3}$$

Q.1

Part B

$$y(n) - 0.7y(n-1) + 0.1y(n-2)$$

$$= z^2(n) - z(n-2)$$

$$= \lambda^2 - 0.7\lambda + 0.1 = 0$$

Note,

$$\lambda = \frac{1}{2}, \frac{1}{5} \text{ Here}$$

$$y_n(n) = c_1 \frac{1}{2}^n + c_2 \frac{1}{5}^n$$

with $z(n) = f(n)$ we have

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0 \Rightarrow y(1) = 1.4$$

$$\text{Hence } c_1 + c_2 = 2 \text{ and } \frac{1}{2}(c_1 + \frac{1}{5})$$

$$= 1.4 = \frac{7}{5} \quad p=0$$

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$$\Rightarrow c_1 + \frac{2}{5} c_2 = \frac{14}{5}$$

These equation yield

$$c_1 = \frac{10}{3}, c_2 = -\frac{4}{3}$$

$$h(n) = \left[\frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

The step response is

$$\Rightarrow S(n) = \sum_{k=0}^n h(n-k)$$

$$\frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{4}{3} \left(\frac{1}{5}\right)^n (5^{n+1} - 1) u(n)$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{4}{3} \left(\frac{1}{5}\right)^n (5^{n+1} - 1) u(n)$$

Part "b" (b)

Soln.

$$X(z) = \frac{1}{1-az^{-1}} \quad |z| > |a|$$

Using the complex inversion integral we have

$$\begin{aligned} x(n) &= \frac{1}{2\pi j} \oint_C \frac{z^{n-1}}{1-az^{-1}} dz \\ &= \frac{1}{2\pi j} \oint_C \frac{z^n dz}{z-a} \end{aligned}$$

Where 'C' is a circle at radius greater than |a| we shall evaluate this integral with $f(z) = z^n$ we distinguish two cases

~~The following are the applications.~~

1) If $n \geq 0$, $f(z)$ has only zeros and hence no pole inside C. The only pole inside C is $z = a$ hence

$$x(n) = f(z_0) = a^n \quad n \geq 0$$

2) If $n < 0$, $f(z) = z^n$ has an n th order pole at $z = 0$ which is also inside. There are contributions from both poles. For $n = -1$ we have

$$x(-1) = \frac{1}{2\pi j} \oint_C \frac{1}{z^2(z-a)} dz$$

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$$= \frac{1}{z-a} \Big|_{z=0} + \frac{1}{z} \Big|_{z=a} = 0$$

If $n = -2$ we have

$$x(-2) = \frac{1}{2\pi j} \oint_C \frac{1}{z^2(z-a)} dz$$

$$= \frac{d}{dz} \left(\frac{1}{z-a} \right) \Big|_{z=0} + \frac{1}{z^2} \Big|_{z=a} = 0$$

By continuing in the same way we can show that $x(n) = 0$ for $n < 0$ thus

$$x(n) = a^n u(n)$$

x x x x x x x x x = x = x

Q2

Part (a)

Sol:

$$X(z) = \frac{1}{(1+2z^{-1})(1-z^{-1})^2}$$

$$X(z) = \frac{1}{4(1+2z^{-1})} + \frac{3}{4} \frac{1}{1-2z^{-1}}$$

$$+ \frac{1}{2} \frac{z^{-1}}{(1-2z^{-1})^2}$$

By applying inverse transform from

$$x(n) = \frac{1}{8} (-1)^n u(n) - \frac{3}{8} u(n) + \frac{1}{2} n u(n)$$

$$= \left[\frac{1}{8} (-1)^n + \frac{3}{8} + \frac{n}{2} \right] u(n)$$

Q3

(b)

(6)

Soll. clearly, the filter must
move pole

$$P_1 = re^{j\pi/2}$$

and zeros at $z=1$ and $z=-1$

$$H(z) = G \frac{(z-1)(z+1)}{(z-jr)(z+jr)}$$

$$(z-jr)(z+jr)$$

$$= G \frac{z^2-1}{z^2+r^2}$$

the gain factor is determined

by evaluating the frequency response
 $H(\omega)$ of the filter at

$\omega = \pi/2$ Thus we have

$$H(\pi/2) = G \frac{2}{1-r^2} = 1$$

$$G = \frac{1-r^2}{2}$$

The value of r is determined
by evaluating $H(\omega)$ at $\omega = 4\pi/9$

thus we have

$$\left| H\left(\frac{4\pi}{9}\right) \right|^2 = \frac{(1-r^2)^2}{4} \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)} = \frac{1}{2}$$

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$$= H\left(\frac{4\pi}{9}\right) = \frac{(1-r^4)^2}{1+r^4+2r^2\cos\left(\frac{8\pi}{9}\right)}$$

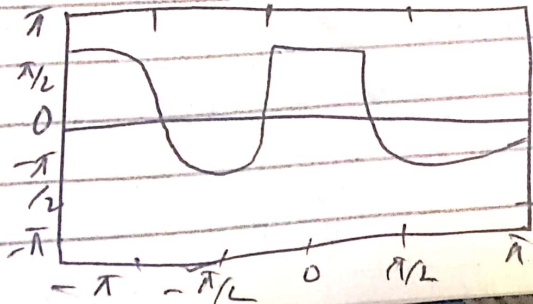
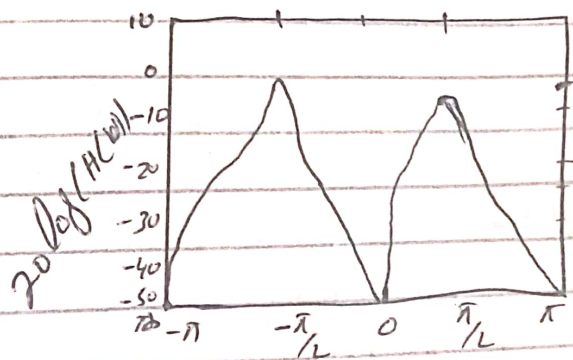
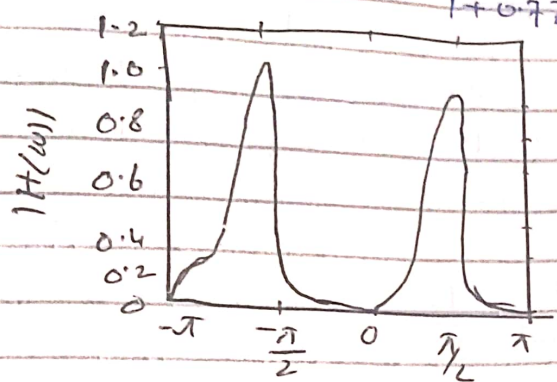
$$= \frac{1}{2}$$

or

$$1.94(1-r^4)^2 = 1 - 1.82r^2 + r^4$$

The value of $r^2 = 0.7$ satisfies this equation therefore the system function for the closed filter is

$$H(z) = \frac{0.15(1-z^{-2})}{1+0.7z^{-2}}$$



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Figure magnitude and phase response
of a sample filter.

$$H(z) = 0.15(1-z^{-2}) / (1+0.7z^{-2})$$

It should be emphasized that the main purpose of the foregoing methodology for designing simple digital filter by pole-zero placement to provide insight into the effect that poles and zeroes have on the frequency response characteristic of system.

The methodology is not intended as a good method for designing digital filter with well-specified pass band and stop band characteristic.

Systematic method for the design of sophisticated digital filter for practical application are discussed.

Q3

Part a

Solution

At $w=0$ we have

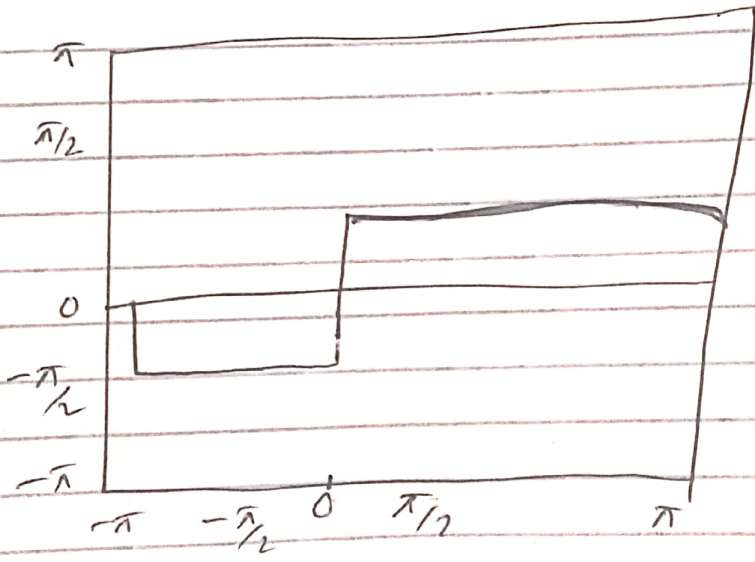
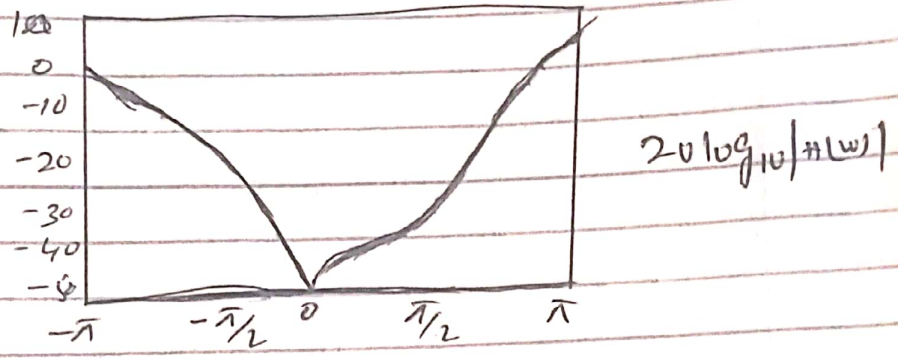
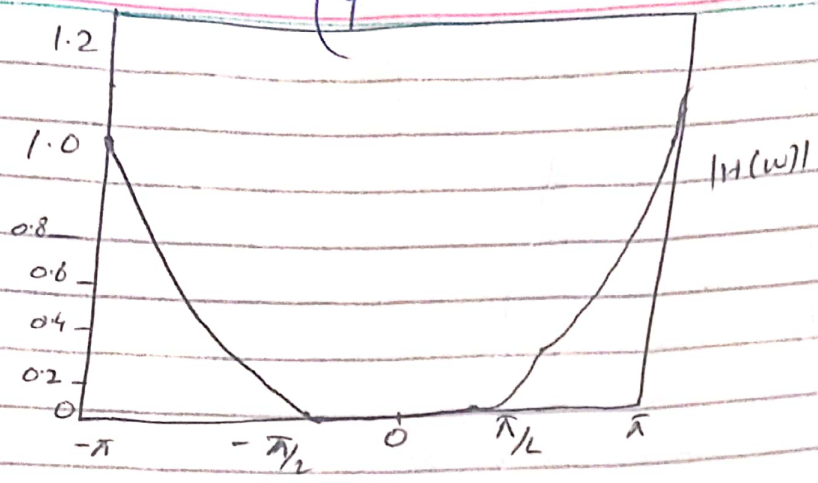
$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

Hence

$$b_0 = (1-p)^2$$

$p \neq 0$

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$R =$

At $\omega = \pi/4$

$$(1 - P)^2$$

$$1 - P \cos(\pi/4) + \sqrt{P} \sin(\pi/4)$$

$$= \frac{(1 - P)^2}{1 - P/\sqrt{2} + \sqrt{P}/\sqrt{2}}$$

$$= \frac{(1 - P)^2}{(1 - P/\sqrt{2} + \sqrt{P}/\sqrt{2})^2} = \frac{1}{2}$$

$$= \frac{(1 - P)^2}{(1 - P/\sqrt{2} + \sqrt{P}/\sqrt{2})^2} = \frac{1}{2}$$

x x x x

~~Q4~~

Q4:

Part (a)

$$X(n) = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{otherwise} \end{cases}$$

Determine the N-point DFT of this sequence for $n \geq L$

Soll:

$$X(\omega) = \sum_{n=0}^{L-1} X(n) e^{-j\omega n}$$

$$= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}}$$

$$= \frac{\sin(\omega L/2) e^{-j\omega(L-1)/2}}{\sin(\omega/2)}$$

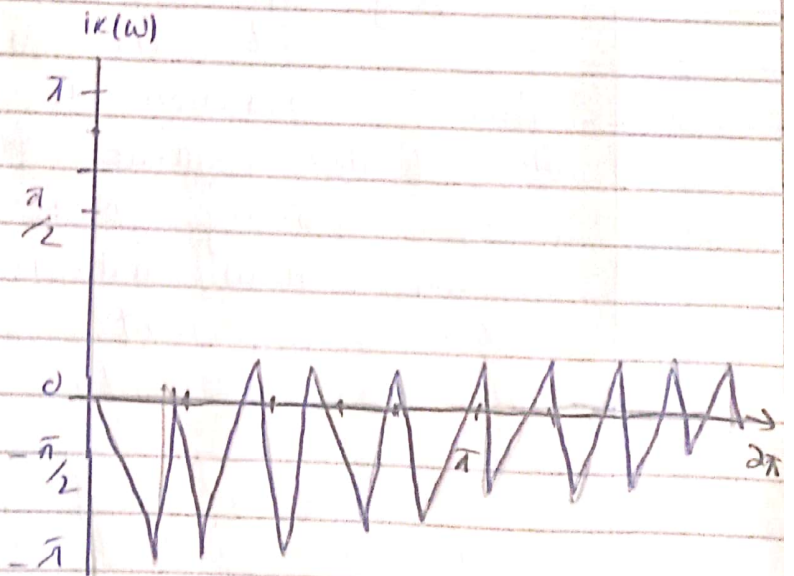
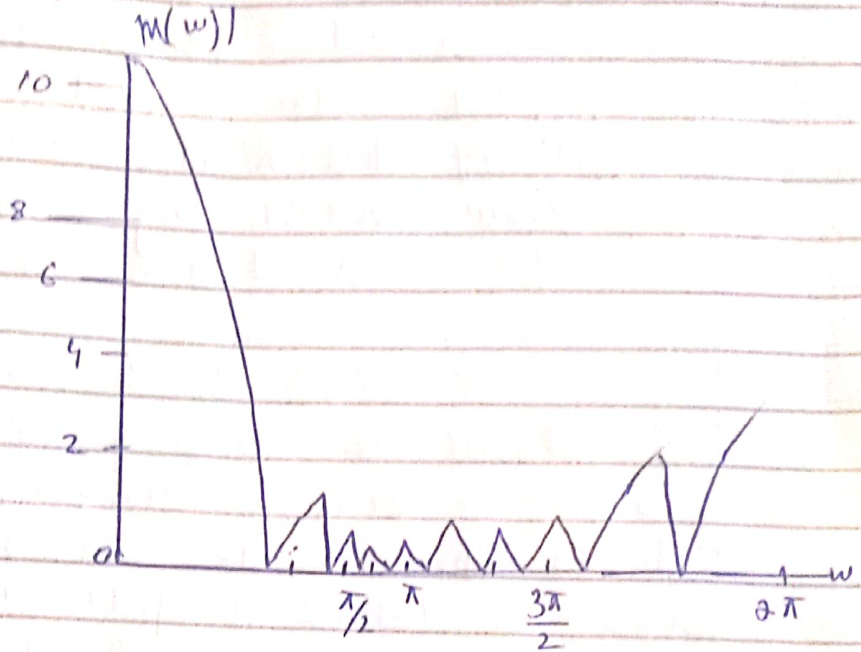
The magnitude and phase of $x(\omega)$ are illustrated for $L=10$. The N-point DFT of $x(n)$ is simply $x(\omega)$ evaluated at the set of N equality spaced frequencies

$$\omega_k = \frac{2\pi k}{N}, \quad k = 0, 1, \dots, N-1$$

$$X(k) = \frac{1 - e^{-j2\pi k L/N}}{1 - e^{-j2\pi k/N}}$$

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$$\frac{\sin(\pi kL/N) - j\pi k(L-1)/N}{\sin(\pi k/N)}$$



If N is selected such that $N=L$
then DFT become

$$X(k) = \begin{cases} L & k=0 \\ 0 & k=1, 2, \dots, L-1 \end{cases}$$

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Thus there is only one non zero value in the DFT this is apparent from observation of $x(w) \sin \pi k/L = 0$ at frequencies $w_k = 2\pi k/L$. The reader should verify that $x(n)$ can be recovered from $X(k)$ by performing an L -point IDFT.

Although the L -point DFT is sufficient to uniquely represent the sequence $x(n)$ in the frequency domain it is apparent that it does not provide sufficient detail to yield a good picture. We must evaluate (interpolate) $x(w)$ at more closely spaced frequencies say $w_k = 2\pi k/N$ where $N > L$, in effect we can view this computation as expanding the size of the sequence from ' L ' points to ' N ' by appending $N-L$ zeroes to the sequence $x(n)$ that is zero size padding. Then the N -point DFT provides finer interpolation than the L -point

$L=10$, $N=50$, and $N=100$
Now spectral characteristics of the sequence are more clearly evident as one will conclude by comparing these spectra with the continuous spectrum $x(w)$.

Q4:

Circular Convolution

(a)

$$x_1(n) = \{1, 2, 3, 4\}$$

(b)

$$\text{and } x_2(n) = \{1, 2, 1, 2\}$$

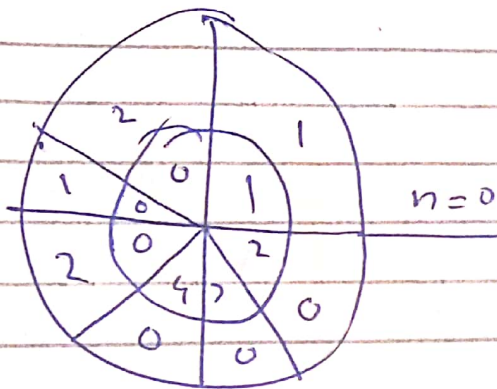
$$L = 4, m = 4$$

$$\text{length of } y[n] = L + m - 1 = 4 + 4 - 1 = 7$$

$$x_1[n] = \{1, 2, 3, 4, 0, 0, 0\}$$

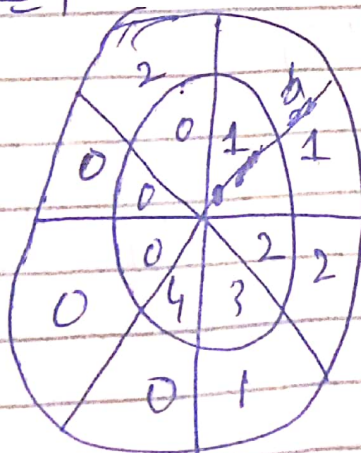
$$x_2[n] = \{1, 2, 1, 2, 0, 0, 0\}$$

for $y(0)$



$$y(0) = 1 \times 1 = 1$$

for $y(1)$

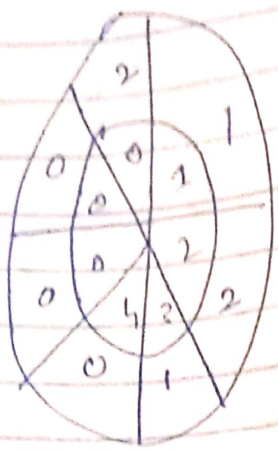


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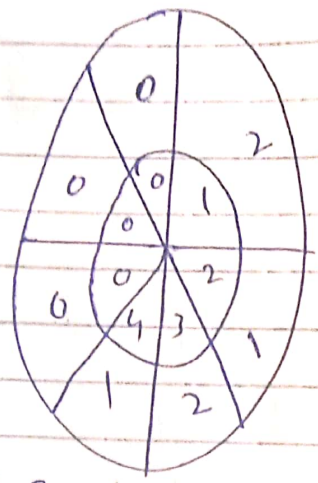
$$y(1) = 1 \times 1 + 2 \times 2 + 3 = 4$$

For $y(2)$,



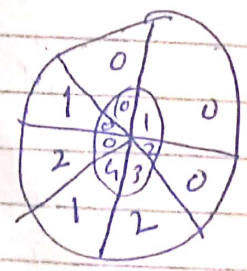
$$y(2) = 1 \times 1 + 2 \times 2 + 3 \times 1 = 8$$

For $y(3)$



$$\therefore y(4) = 4 \times 2 + 3 \times 1 + 2 \times 2 = 15$$

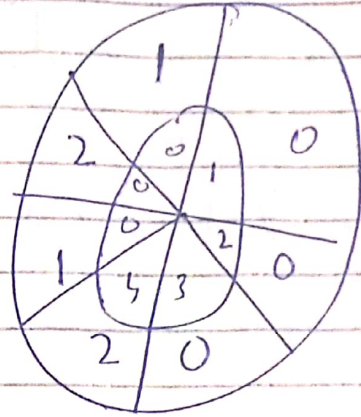
For $y(5)$



0, 1, 0

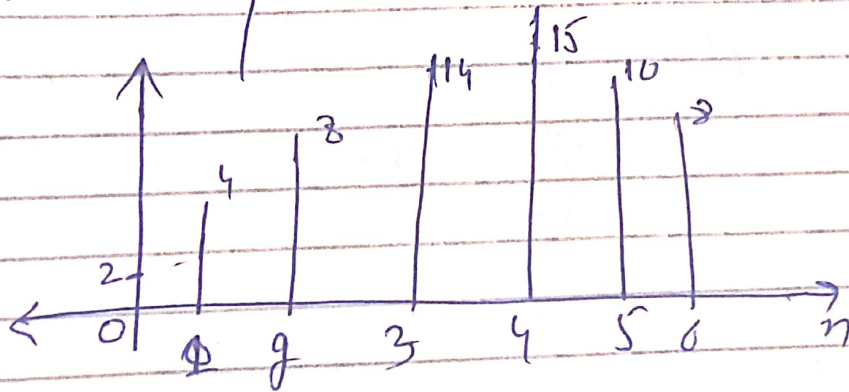
$$y(5) = 4 \times 1 + 3 \times 2 = 10$$

$$y(6)$$



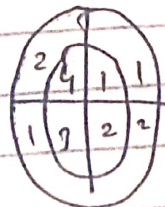
$$: y(6) = 4 \times 2 = 8$$

$$y[n] = \{ 1, 4, 8, 14, 15, 10, 8 \}$$



Linner using circular convolution

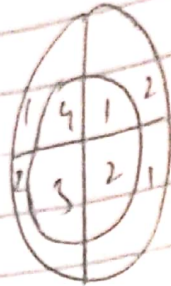
For $y=0$



0.10

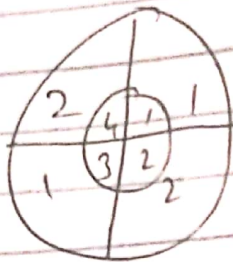
$$y(0) = 1 + 4 + 3 + 8 = 16$$

for $y(1)$:



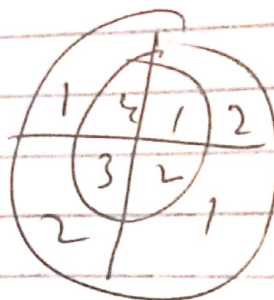
$$: y(1) = 2 + 2 + 6 + 4 = 14$$

for $y(2)$



$$y(2) = 1 + 4 + 3 + 8 = 16$$

for $y(3)$



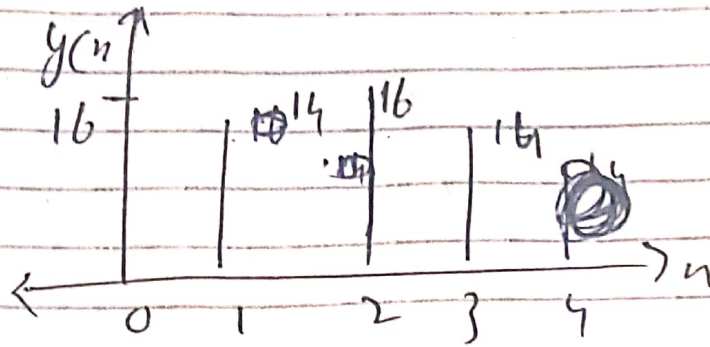
$$y(3) = 2 + 2 + 6 + 4 = 14$$

$$y(n) = 16, 14, 16, 14$$

p. to

12

$$y(n) = 16, 14, 16, 14$$



End.