

Assgn

Assignment

Name

Majid Mahmood.

ID

13876.

Subject

Dsp.

— x — x — x — x — x —

Question no 1(a)

Determine the response $y(n)$ $n \geq 0$. . .

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

to put $x(n) = 4^n u(n)$.

Solution:

$$y_h(n) = C_1(-1)^n + C_2(4)^n$$

$$y_p(n) = K(4)^n u(n).$$

upon substitution we obtain . . .

$$K(4)^n u(n) - 3K(n-1)(4)^{n-1} u(n-1) - 4K(n-2)(4)^{n-2} u(n-2) \\ = (4)^n u(n) + 2(4)^{n-1} u(n-1)$$

To determine K we evaluate this equation for any $n \geq 2$. we select $n=2$ for we obtain

$$K = 6/5 \text{ therefore.}$$

$$y_p(n) = \frac{6}{5} n(4)^n u(n).$$

The total solution to the difference equation.

$$y(n) = C_1(-1)^n + C_2(4)^n + \frac{6}{5} n(4)^n \quad n \geq 0$$

where the constants C_1 and C_2 are determined such that the initial condition are satisfied.

$$y(0) = 3y(-1) + 4y(-2) + 1$$

$$y(1) = 3y(0) + 4y(-1) + 6$$

$$\Rightarrow 13y(-1) + 12y(-2) + 9$$

We evaluate at $n=0$ and $n=1$

$$y(0) = 4 + c_2$$

$$y(1) = -c_1 + 4c_2 + \frac{24}{5}$$

We can equate these two sets of relations to obtain c_1 and c_2 .

We can simplify the computations above by setting $y(-1) = y(-2) = 0$.

then we have.

$$c_1 + c_2 = 4$$

$$-c_1 + 4c_2 + \frac{24}{5} = 9$$

$$\text{hence } c_1 = -\frac{1}{25} \text{ and } c_2 = \frac{26}{25}$$

Finally we have the zero mode state response to the forcing f/n $x(n) = (4)^n u(n)$

$$y_{25}(n) = -\frac{1}{25}(-1)^n + \frac{26}{25}(4)^n + \frac{6}{5}n(4)^n \geq 0$$

————— x ————— x —————

Q no 1 (B).

Solution:

$$y(n) = 0.6y(n-1) - 0.08y(n-2) + x(n)$$

$$y(n) - 0.6y(n-1) + 0.08y(n-2) = x(n)$$

To obtain the homogeneous equation.

Set input $x(n) = 0$.

$$y(n) - 0.6y(n-1) + 0.08y(n-2) = 0.$$

$$y_h(n) = \lambda^n$$

Substitute the solution to the homogeneous equation.

$$\lambda^n - 0.6\lambda^{n-1} + 0.08\lambda^{n-2} = 0.$$

$$\lambda^{n-2} [\lambda^2 - 0.6\lambda + 0.08] = 0.$$

$$\lambda^2 - 0.6\lambda + 0.08 = 0.$$

$$(\lambda - 0.2)(\lambda - 0.4) = 0$$

$$\lambda_1 = 0.2 \quad \lambda_2 = 0.4.$$

$$y_h(n) = C_1(\lambda_1)^n + C_2(\lambda_2)^n$$

$$y(n) = C_1(0.2)^n + C_2(0.4)^n \rightarrow \textcircled{1}$$

The particular solution is

$$y_p(n) = K(-1)^n u(n)$$

$$K(-1)^n u(n) - 4K(-1)^{n-1} u(n-1) + 4K(-1)^{n-2} u(n-1) = 0$$
$$= (-1)^n u(n) - (-1)^{n-1} u(n-1)$$

$$\text{For } n \geq 2, K(1+4+4) = 2.$$

$$K = \frac{2}{9}$$

The total solution is.

$$y(n) = (C_1 \cdot 2^n + C_2 n \cdot 2^n + \frac{2}{9} (-1)^n) u(n).$$

From initial condition we obtained.

$$y(-1) = y(-2) = 0.$$

$$C_1 + \frac{2}{9} = 0.$$

$$\boxed{C_1 = -\frac{2}{9}}$$

$$2C_1 + 2C_2 - \frac{2}{9} = 0.$$

$$\boxed{C_2 = \frac{1}{3}}$$

Question no 2 (a).

Determine the casual signal $x(n]$

$$X(z) = \frac{1}{(z-2)(z-1)^2}$$

Sol

$$\frac{X(z)}{z} = \frac{z^2}{(z-2)(z-1)^2}$$

$$\frac{X(z)}{z} = \frac{A}{z-2} + \frac{B}{z-1} + \frac{C}{(z-1)^2}$$

Find A, B and C.

$$A = 4$$

$$B = 3$$

$$C = -1$$

hence $x(n) = [4(2)^n - 3 - n]u(n)$.

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Q no 2 (b)
 Determine the partial fraction expansion
 of the prop's function.

$$x(z) = \frac{1}{1 - 1.5z^{-1} + 0.5z^{-2}}$$

Solution

First we eliminate negativity powers
 by multiply both numerator and denominator
 by z^2 thus

$$x(z) = \frac{z^2}{z^2 - 1.5z + 0.5}$$

The poles are $P_1 = 1$ and $P_2 = 0.5$

$$\frac{x(z)}{(z)} = \frac{z}{(z-1)(z-0.5)} = \frac{A_1}{z-1} + \frac{A_2}{z-0.5}$$

A very simple method to determine
 A_1 and A_2 is multiply the equation
 by denominator term $(z-1)(z-0.5)$.

$$\text{Thus } z = (z-0.5)A_1 + (z-1)A_2 \rightarrow \textcircled{1}$$

Set $z = P_1 = 1$ in eq $\textcircled{1}$.

$$1 = 1 - 0.5A_1$$

$$A_1 = 2$$

Set $z = P_2 = 0.5$ in eq $\textcircled{1}$.

$$0.5 = (0.5 - 1)A_2$$

$$A_2 = -1$$

We can determine the coefficient A_1, A_2, \dots, A_N
 by multiplying both sides $\textcircled{1}$
 by each term $(z - P_k)$ $k = 1, 2, 3, 4, \dots, N$.

Question No 3 (a).

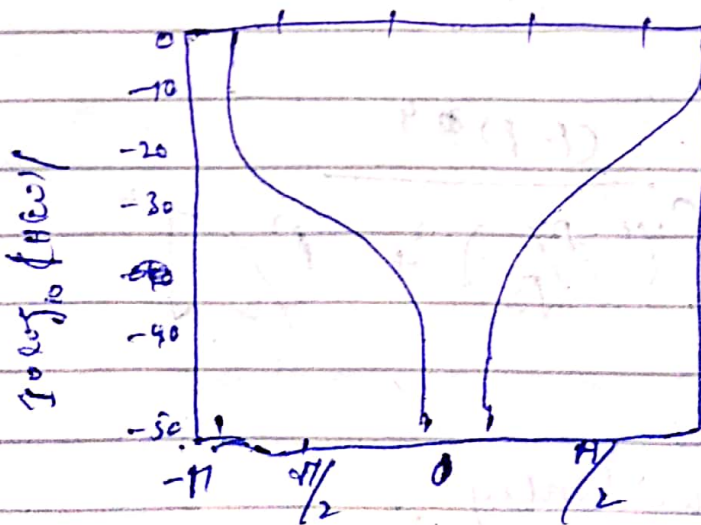
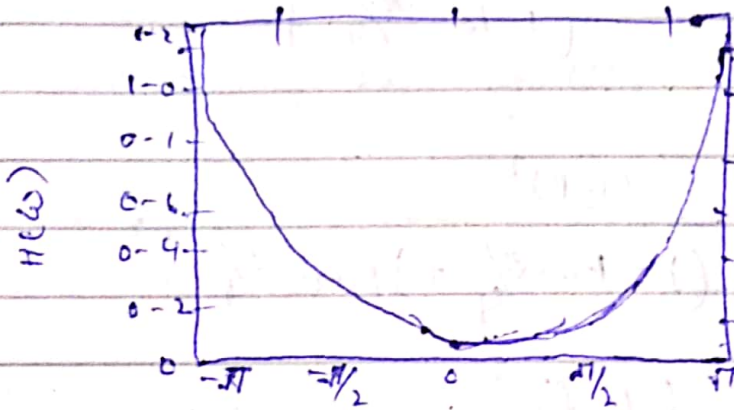
A two-pole low pass has the system response --- $H(z) = \frac{b_0}{(1-pz^{-1})^2}$ ---

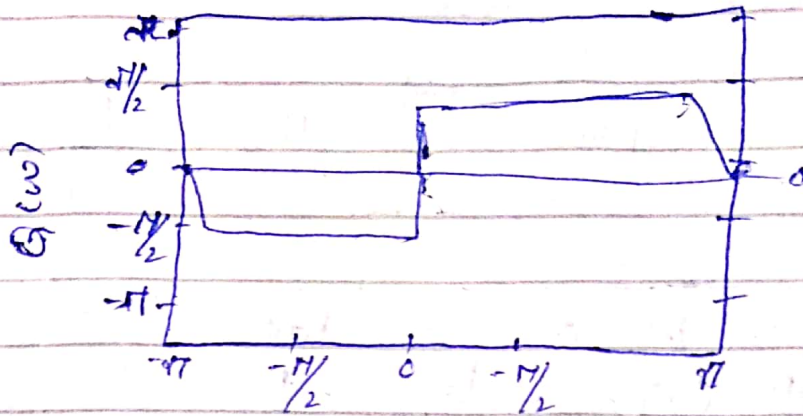
Sol:

At $\omega = 0$ we have

$$H_2(0) = \frac{b_0}{(1-p)^2} = 1$$

$$b_0 = (1-p)^2$$





At $\omega = \frac{\pi}{4}$

$$H\left(\frac{\pi}{4}\right) = \frac{(1-P)^2}{\left(1 - Pe^{-j\pi/4}\right)^2}$$

$$= \frac{(1-P)^2}{\left(1 - P\cos\frac{\pi}{4} + jP\sin\frac{\pi}{4}\right)^2}$$

$$= \frac{(1-P)^2}{\left(1 - P/\sqrt{2} + jP/\sqrt{2}\right)^2}$$

$$\text{Hence} = \frac{(1-P)^4}{\left[\left(\frac{1-P}{\sqrt{2}}\right)^2 + \frac{P^2}{2}\right]^2} = \frac{1}{2}$$

or equivalently

$$\sqrt{2} (1-P)^2 = 1 + P^2 - \sqrt{2}P$$

The value of $P = 0.32$ satisfy this equation the system f/m is

the desired filter is

$$H(z) = \frac{0.46}{(1 - 0.32z^{-1})^2}$$

the same principle can be applied for design of band pass filters.

x ————— x

Question 3(B):

Design a Two-Pole band pass filter.

Solution:

clearly, The filter must have poles at $p_{1,2} = \pm j\omega_c/2$ and

zeros at $z = 1$ and $z = -1$
consequently the system f/m is:

$$H(z) = \frac{g(z-1)(z+1)}{(z-jr)(z+jr)}$$
$$\Rightarrow G = \frac{(z^2-1)}{(z^2-r^2)}$$

The Gain factor is determined by evaluating the frequency response $|H(\omega)|$ of the filter at $\omega = \omega_c/2$. Thus we have,

$$H(\omega_c/2) = g \frac{2}{1-r^2} = 1$$

$$g = \frac{1-r^2}{2}$$

The value of r is determined by evaluating $H(\omega)$ at $\omega = \omega_c/2$. Thus we have.

$$\left[H\left(\frac{4\pi}{9}\right) \right]^2 = \frac{(1-r^2)^2}{4} \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)}$$

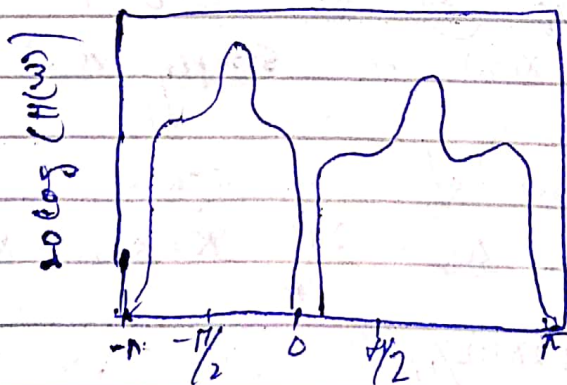
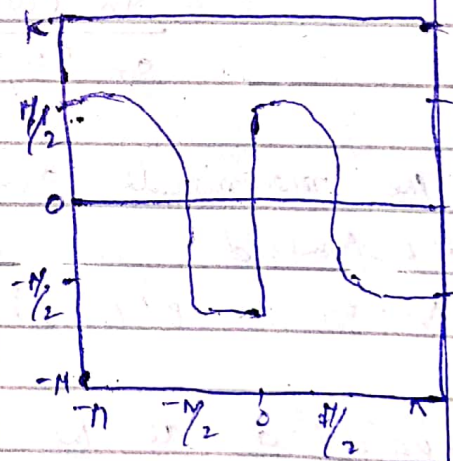
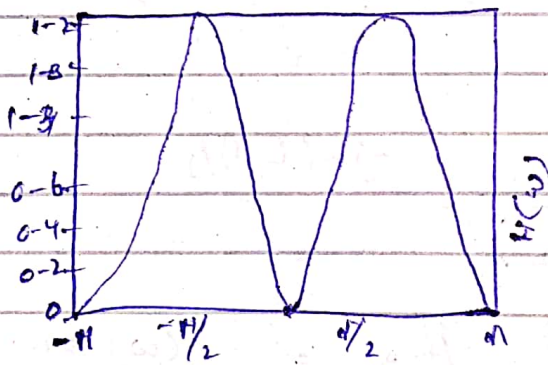
$$= \frac{1}{2}$$

$$\text{or } 1.94(1-r^2)^2 = 1 - 1.08r^2 + r^4$$

The value of $r^2 = 0.7$ satisfy this equation.

The system f/m for the design filter is -

$$H(z) = \frac{0.15(1-z^2)}{(1+0.7z^2)}$$



Question no 4 (a)

A finite duration sequence of length L is given as.

$$x(n) = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{otherwise} \end{cases}$$

Determine N point DFT for this sequence for $N \geq L$.

Sol:

The Fourier Transform of this sequence is.

$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} \Rightarrow$$

$$= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}$$

The magnitude and phase of $X(\omega)$ are illustrated in figure (1) for $L=8$. The N -point DFT of $x(n)$ is simply $X(\omega)$.

evaluated at the set of N equally spaced frequencies - $\omega_k = \frac{2\pi k}{N}$ $k = 0, 1, 2, \dots, N-1$

$$X(k) = \frac{1 - e^{-j2\pi k L/N}}{1 - e^{-j2\pi k/N}} \quad k = 0, 1, 2, \dots, N-1$$

$$= \frac{\sin(\pi k L/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$

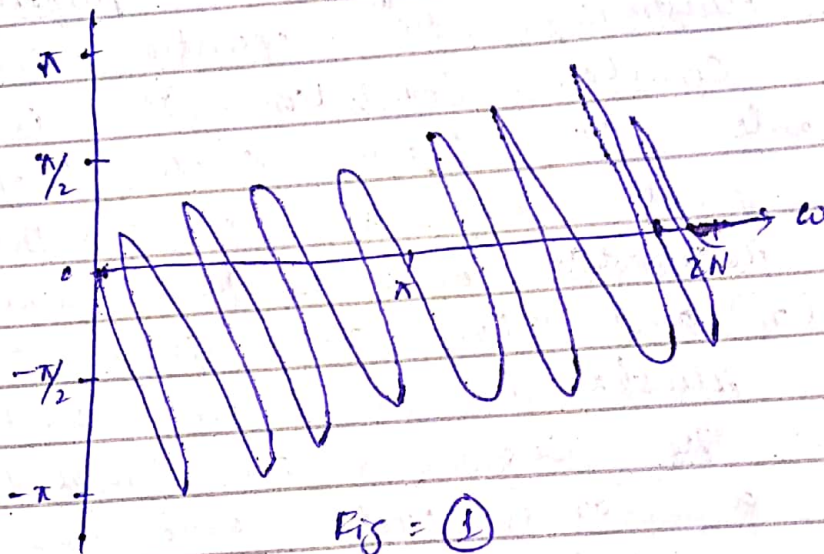
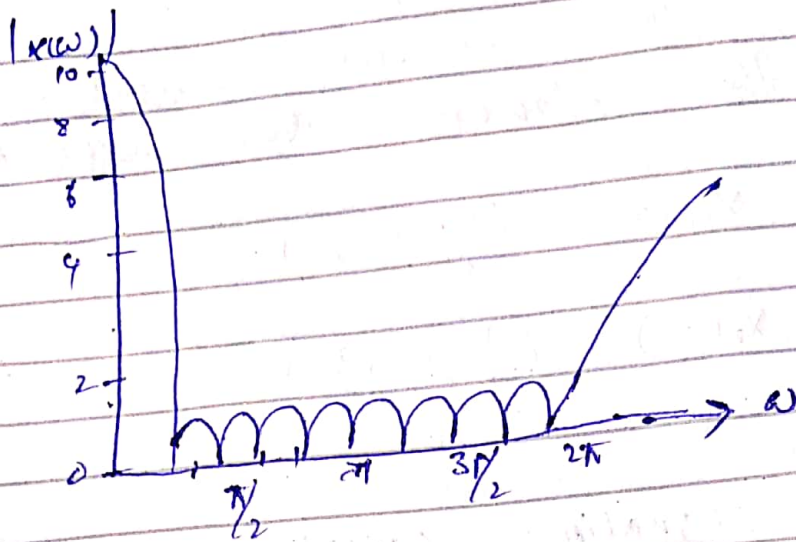


Fig = ③

If N is selected such that $N=L$.
then DFT become.

$$x^*(k) = \begin{cases} L & k=0 \\ 0 & k=1, 2, 3, \dots, L-1 \end{cases}$$

thus there is only one non zero value in the DFT. This is apparent from observation of $x(\omega) = 0$ at frequencies $\omega_k = \frac{2\pi k}{L}$ $k \neq 0$ then reader should verify that $x(n)$ can be cat recovered from $x(k)$ by performing an L -point IDFT.

Question no 4. (d).

Compute the DFT of the four point sequence.

$$x(n) = (0, 1, 2, 3)$$

Solution.

The first step to determine the matrix W_4 . By exploiting the periodicity property of W_4 and the symmetry property.

$$W_4^{k+n/2} = -W_4^k$$

the matrix W_4 may be expressed as.

$$W_4 = \begin{bmatrix} W_4^0 & W_4^0 & W_4^0 & W_4^0 \\ W_4^0 & W_4^1 & W_4^2 & W_4^3 \\ W_4^0 & W_4^2 & W_4^4 & W_4^6 \\ W_4^0 & W_4^3 & W_4^6 & W_4^9 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^1 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^4 & W_4^6 \\ 1 & W_4^3 & W_4^6 & W_4^9 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & j \end{bmatrix}$$

then

$$X_4 = W_4 x_1 = \begin{bmatrix} 6 \\ -2+2j \\ -2 \\ -2-2j \end{bmatrix}$$

the IDFT of X_4 may be determined by conjugating the elements in W_4 to obtain W_4^* and then apply formula