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Assignment

01

Digital

Signal

Processing

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(1)

Q (a):

Determine the response $y(n]$
 $n \geq 0$ of the system described
by the second-order difference
equation

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

to the input $x(n) = 4^n u(n)$

Solution:

Consider the difference equation

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

The homogeneous equation of the system
is $y(n) - 3y(n-1) - 4y(n-2) = 0$

The characteristic equation of the
system

$$\lambda^2 - 3\lambda - 4 = 0$$

$$\lambda^2 - 4\lambda + \lambda - 4 = 0$$

Determine the roots of the
characteristic equation

$$\lambda^2 - 4\lambda + \lambda - 4 = 0$$

$$\lambda(\lambda - 4) + 1(\lambda - 4) = 0$$

$$\lambda(-4)(\lambda + 1) = 0$$

$$\lambda = -1, 4$$

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The homogeneous solution is
 $y_h(n) = c_1 (-1)^n u(n) + c_2 (4)^n u(n)$

Since 4 is a characteristic root
and the equation is

$$y(n) = 4^n u(n)$$

We assume a particular solution
of the form

$$y_p(n) = kn 4^n u(n)$$

Then

$$\begin{aligned} kn 4^n u(n) - 3k(n-1) 4^{n-1} u(n-1) - 4k(n-2) \\ 4^{n-2} u(n-2) \\ = 4^n u(n) + 2(4)^{n-1} u(n-1) \end{aligned}$$

For $n=2$

$$k(32-12) = 4^2 + 8 = 24 \rightarrow k = \frac{6}{5}$$

The total solution is

$$\begin{aligned} y(n) &= y_p(n) + y_h(n) \\ &= \left[\frac{6}{5} n 4^n + c_1 4^n + c_2 (-1)^n \right] u(n) \end{aligned}$$

The solve for c_1 and c_2 we assume
that

$$y(-1) = y(-2) = 0 \text{ then}$$

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$$y(0) = 1 \quad \text{and}$$

$$y(1) = 3y(0) + 4 + 2 = 9$$

Hence

$$C_1 + C_2 = 1 \quad \text{and}$$

$$\frac{24}{5} + 4C_1 - C_2 = 9$$

$$4C_1 - C_2 = \frac{21}{5}$$

$$\text{Therefore } C_1 = \frac{26}{25} \quad \text{and } C_2 = -\frac{1}{25}$$

The total solution is,

$$y(n) = \left[\frac{6}{5} n 4^n + \frac{26}{25} 4^n - \frac{1}{25} (-1)^n \right] u(n)$$

Q (1) Part (B)

Determine the impulse response and unit step response of the system describe by the difference equation

$$y(n] = 0.6y[n-1] - 0.08y[n-2] + x[n]$$

Sol:.

Consider the difference eq

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$$y(n) = 0.6y(n-1) - 0.08y(n-2) + x(n)$$

$$y(n) = 0.6y(n-1) + 0.08y(n-2) = x(n)$$

To obtain the homogeneous equation
set input

$$x(n) = 0$$

$$y(n) = 0.6y(n-1) + 0.08y(n-2) = 0$$

Determine the solution of the
homogeneous equation

$$y_h(n) = \lambda^n$$

Substitute the solution obtained in
the homogeneous equation

$$\lambda^n - 0.6\lambda^{n-1} + 0.08\lambda^{n-2} = 0$$

$$\lambda^{n-2} (\lambda^2 - 0.6\lambda + 0.08) = 0$$

$$\lambda^2 - 0.6\lambda + 0.08 = 0$$

$$(\lambda - 0.2)(\lambda - 0.4) = 0$$

there fore the root are

$$\lambda_1 = 0.2 \quad \lambda_2 = 0.4$$

Thus the general form of the
solution to the homogeneous equation is

$$y_h(n) = C_1 (\lambda_1)^n + C_2 (\lambda_2)^n$$

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$$y(n) = C_1 (0.2)^n + C_2 (0.4)^n \quad \dots (i)$$

$\lambda = 0.2$, $\lambda = 0.4$ hence

$$y(n) = C_1 \frac{1}{5}^n + C_2 \frac{2}{5}^n$$

With $x(n) = \delta(n)$ the initial conditions are

$$y(0) = 1$$

$$y(1) - 0.6y(0) = 0$$

$$y(1) = 0.6$$

Hence $C_1 + C_2 = 1$ and

$$\frac{1}{5} C_1 + \frac{2}{5} C_2 = 0.6$$

$$C_1 = -1, \quad C_2 = 2$$

Therefore $h(n) = \left[-\left(\frac{1}{5}\right)^n + 2\left(\frac{2}{5}\right)^n \right] u(n)$

The step response is

$$y(n) = \sum_{k=0}^n h(n-k), \quad n > 0$$

$$= \sum_{k=0}^n \left[2 \left(\frac{2}{5}\right)^{n-k} - \left(\frac{1}{5}\right)^{n-k} \right]$$

$$= \left[\frac{1}{0.12} \left[\frac{2^{n+1}}{5} - 1 \right] - \frac{1}{0.16} \left[\frac{1^{n+1}}{5} - 1 \right] \right] u(n)$$

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Q 2 Part (a)

Determine the causal signal $x(n]$ having z-transform

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

Hint: Take inverse z-transform using Partial fraction method

Solution:

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

By Partial fraction method

$$\frac{1}{(1-2z^{-1})(1-z^{-1})^2} = \frac{A}{(1-2z^{-1})} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$\frac{1}{(1-2z^{-1})(1-z^{-1})^2} = \frac{A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1})}{(1-2z^{-1})(1-z^{-1})^2}$$

$$1 = A(1-z^{-1})^2 + B(1-2z^{-1})(1-z^{-1}) + Cz^{-1}(1-2z^{-1})$$

Put $z = 1$

$$1 = (1-0)^2 + B(-2)(1-1) + C(1)(1-2)$$

$$1 = 0 + 0 - C$$

$$1 = -C$$

$$C = -1$$

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Put $z = 2$ in eqn (i)

$$1 = A\left(1 - \frac{1}{2}\right)^2 + B\left(1 - \frac{2}{2}\right)\left(1 - \frac{1}{2}\right) + C\left(\frac{1}{2}\right)\left(1 - \frac{2}{2}\right)$$

$$1 = A\left(\frac{1}{2}\right)^2 + B(1-1)\left(\frac{1}{2}\right) + C\left(\frac{1}{2}\right)(1-1)$$

$$1 = \frac{A}{4} + B(0)\left(\frac{1}{2}\right) + C\frac{1}{2}(0)$$

$$1 = \frac{A}{4} + 0 + 0$$

$$A = 4$$

Put $z = 3$ in eqn (i)

$$1 = A\left(1 - \frac{1}{3}\right)^2 + B\left(1 - \frac{2}{3}\right)\left(1 - \frac{1}{3}\right) + C\left(\frac{1}{3}\right)\left(1 - \frac{2}{3}\right)$$

$$1 = A\left(\frac{4}{9}\right) + B\left(\frac{1}{3}\right)\left(\frac{2}{3}\right) + C\left(\frac{1}{3}\right)\left(\frac{1}{3}\right)$$

$$1 = A \frac{4}{9} + \frac{2}{9} B + \frac{1}{9} C$$

$$1 = \frac{4}{9}(4) + \frac{2}{9} B - \frac{1}{9}$$

$$= 1 + \frac{1}{9} - \frac{1}{9} = \frac{2}{9} B$$

$$= \frac{6}{9} \times \frac{9}{2} = B$$

$$B = -3$$

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Hence $x(n) = [4(2)^n - 3 - n] u_n$

Q 2 Part (B)

Determine the partial fraction expansion of the following

$$X(z) = \frac{1}{1 - 1.5z^{-1} + 0.5z^{-2}}$$

Sol:

Eliminate the $-ve$ Power by multiplying both numerator and denominator by z^2

$$X(z) = \frac{z^2}{z^2 - 1.5z + 0.5}$$

$$\frac{X(z)}{z} = \frac{z}{(z-1)(z-0.5)} = \frac{A}{z-1} + \frac{B}{z-0.5}$$

$$z = (z-0.5)A + (z-1)B \quad \text{--- L.C.M}$$

Now Set $z = p_1 = 1$ in eq (1)

We eliminate the term involving

$$1 = (1-0.5)A$$

$$A = 2$$

Return to eq (1) $z = p_2 = 0.5$ those

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eliminating the term involving

A So we have

$$0.5 = (0.5 - 1)B$$

$$B = -1$$

Q3 Part (1)

A two pole low pass filter has the system response

$$H(z) = \frac{b_0}{(1 - pz^{-1})^2}$$

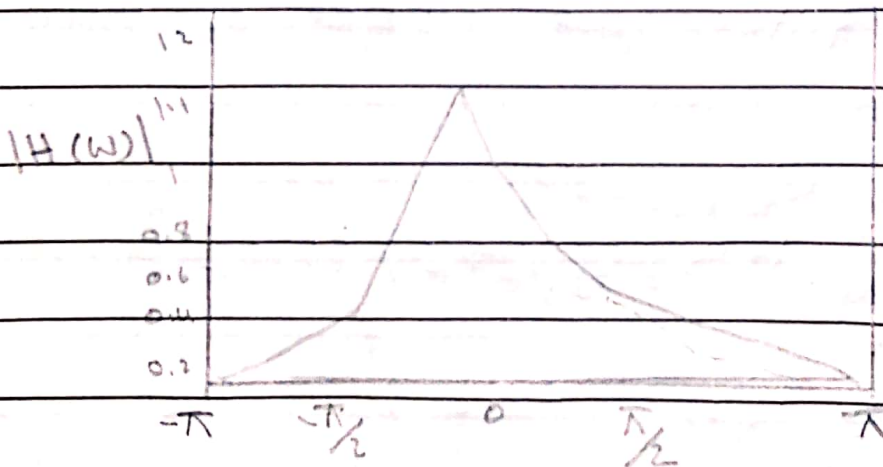
Determine the value of b_0 and

p condition

$$H(0) = 1 \text{ and } |H\left(\frac{T}{4}\right)|^2 = \frac{1}{2}$$

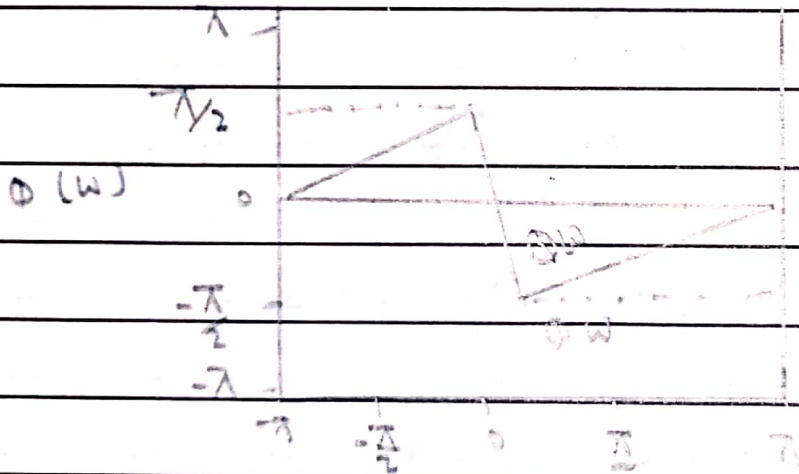
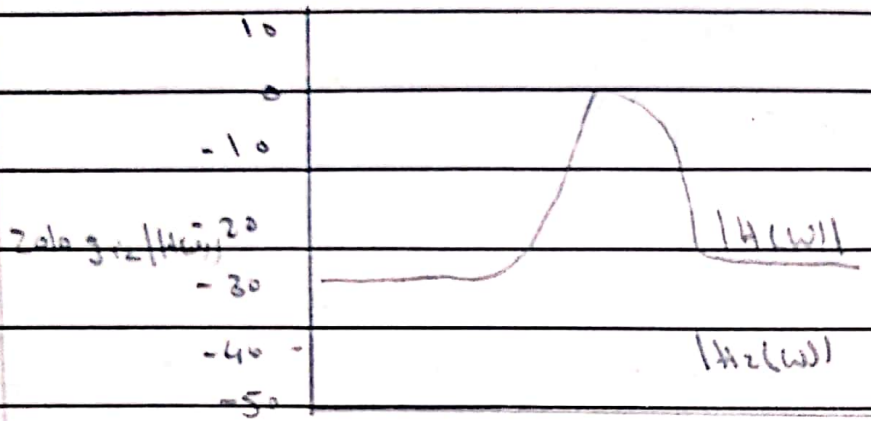
Sol:

Linear time-invariant system as frequency



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Magnitude of and Phase Response of (1) Single Pole filter and (2) a one-pole one-zero filter

$$H_1(z) = (1-a)/(1-az^{-1})$$

$$H_2(z) = \frac{(1-a)(1+z^{-1})}{2(1-az^{-1})}$$

$$a = 0.9$$

Now we have determine the value of h_0 and P

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Such that the frequency response $H(\omega)$ satisfies the condition

$$H(0) = 1 \quad \text{and} \quad |H(\frac{\pi}{4})|^2 = \frac{1}{2}$$

Sol: At

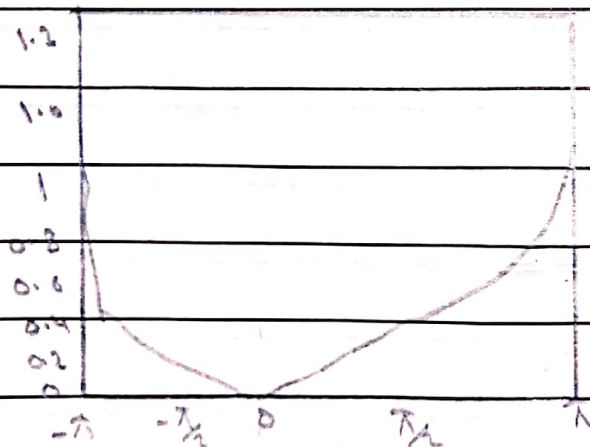
$$\omega = 0$$

$$H(0) = \frac{b_0}{(1-p^2)} = 1$$

both side multiply by $(1-p)^2$

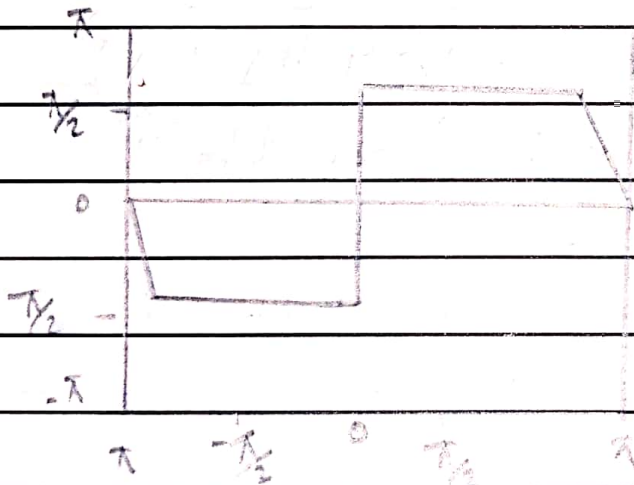
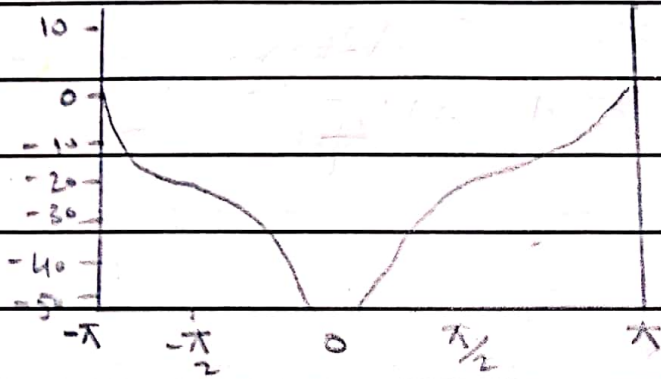
$$\Rightarrow \frac{b_0}{(1-p)^2} \times (1-p)^2 = 1 \times (1-p)^2$$

$$b_0 = 2(1-p)^2$$



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Magnitude and Phase response of a simple high-pass filter

$$H(z) = \frac{(1-a)}{2} \frac{(1-\bar{z})}{(1+az)}$$

$$a = 0.9$$

$$\text{at } \omega = \frac{\pi}{4}$$

$$H\left(\frac{\pi}{4}\right) = \frac{(1-P)^2}{(1-Pe^{j\pi/4})^2}$$

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$$= \frac{(1-P)^2}{(1-P \cos(\pi/4) + JP \sin(\pi/4))^2}$$

$$= \frac{1-P^2}{\left(\frac{1-P}{\sqrt{2}} + \frac{JP}{\sqrt{2}}\right)^2}$$

take square on above eq

$$\frac{(1-P^2)}{\left(\frac{1-P}{\sqrt{2}} + \frac{JP}{\sqrt{2}}\right)^2}$$

$$H(z) = \frac{(1-P)^2}{\left[\left(1-P/\sqrt{2}\right)^2 + P^2/2\right]^2} = \frac{1}{2}$$

$$= \sqrt{2} (1-P)^2 = 1 + P^2 - \sqrt{2} P$$

the value of $P = 0.32$ satisfies this equation desired filter is

$$H(z) = \frac{0.46}{(1-0.32z^{-1})^2}$$

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Q 4 Part (a)
$$\begin{cases} 1, & 0 \leq n \leq L-1 \\ 0, & \text{otherwise} \end{cases}$$

Sol:

The Fourier transform of this sequence is

$$\begin{aligned} X(\omega) &= \sum_{n=0}^{L-1} x(n) e^{-j\omega n} \\ &= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} = \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2} \end{aligned}$$

The magnitude and phase of $X(\omega)$ are illustrated in C-10. The N -Point DFT of $x(n)$ is simply $X(\omega)$ evaluated at the set of N equally spaced frequencies $\omega_k = \frac{2\pi k}{N}$

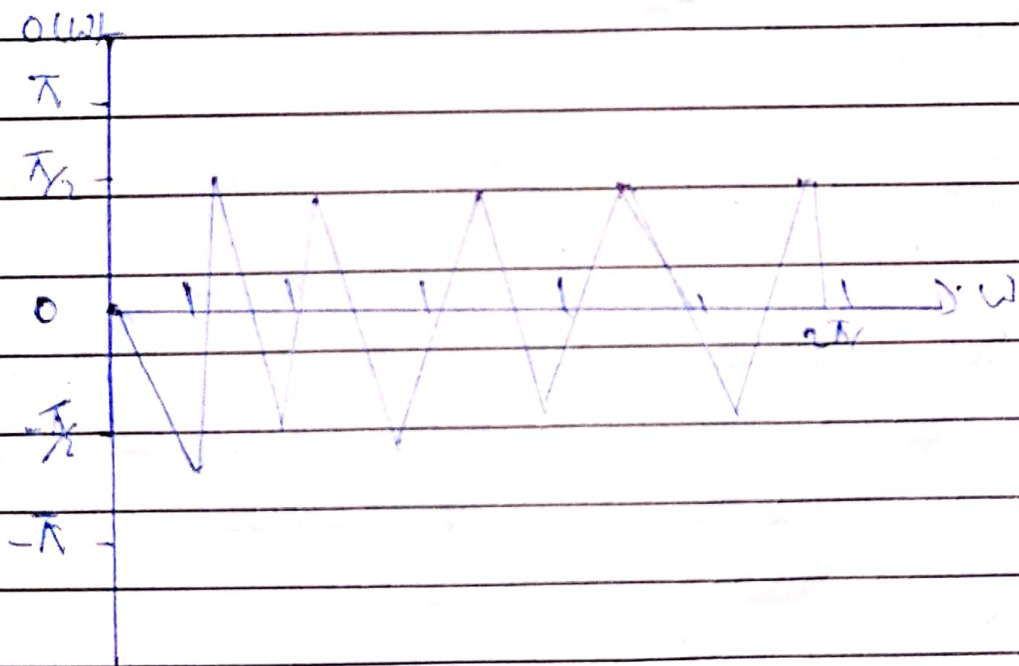
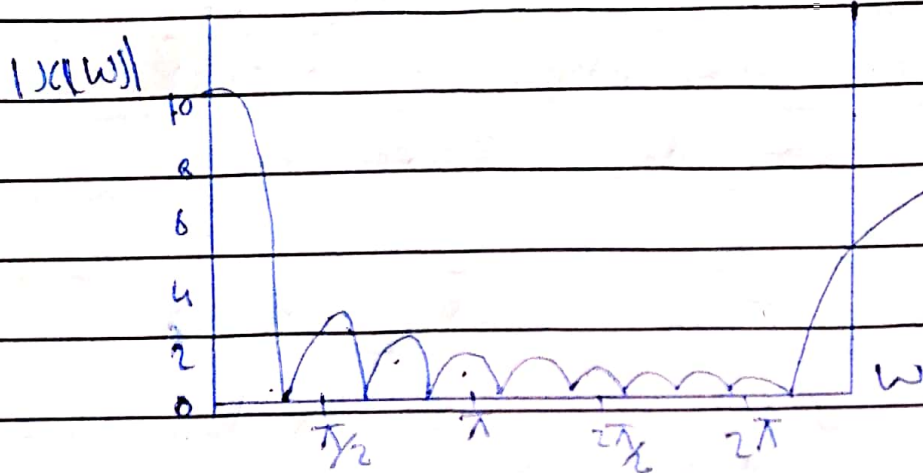
$$k = 0, 1, \dots, N-1$$

$$X(k) = \frac{1 - e^{-j2\pi k L/N}}{1 - e^{-j2\pi k/N}} \quad k = 0, 1, \dots, N-1$$

$$= \left(\frac{\sin(\pi k L/N)}{\sin(\pi k/N)} \right) e^{-j\pi k(L-1)/N}$$

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45,



If N is selected such that $N=L$ then the DFT become

$$X(k) = \begin{cases} L, & k=0 \\ 0, & k=1, 2, \dots, L-1 \end{cases}$$

Thus there is only one non-zero

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non zero value in DFT. This is apparent from observation of $X(\omega)$

Since $X(\omega) = 0$ at the frequencies $\omega_k = 2\pi k/N$

$k \neq 0$ The reader should verify that $x(n)$ can be recovered from $X(k)$ by performing an L -Point DFT

Q No 4 Part (B)

Sol:

The first step is to determine the matrix W_N by exploiting the periodicity property of W_N and the symmetry property

$$W_N^{k+N/2} = W_N^k$$

The matrix W_N may be expressed as

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$$W_4 = \begin{bmatrix} W_4^0 & W_4^0 & W_4^0 & W_4^0 \\ W_4^0 & W_4^1 & W_4^2 & W_4^3 \\ W_4^0 & W_4^2 & W_4^4 & W_4^4 \\ W_4^0 & W_4^3 & W_4^6 & W_4^2 \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & W_4^1 & W_4^2 & W_4^3 \\ 1 & W_4^2 & W_4^0 & W_4^2 \\ 1 & W_4^3 & W_4^4 & W_4^1 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & 1 \\ 1 & j & -1 & -j \end{bmatrix}$$

$$\text{Then } Y_4 = W_4 X_4 = \begin{bmatrix} -2 + 2j \\ -2 - 2j \end{bmatrix}$$

The DFT of X_4 may be determined by conjugating the element in W_4 to obtain W_4^* and then applying the formula