

M UMAIR

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Digital Signal Processing

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## Question 1(a)

$$y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$$

The characteristic equation is

$$x^2 - 4x + 4 = 0$$

$$x = 2, 2 \text{ Hence}$$

$$y_h(n) = c_1 2^n + c_2 n 2^n$$

The particular solution is

$$y_p(n) = k(-1)^n u(n)$$

Substituting the solution into the difference eqn, we obtain

$$k(-1)^n u(n) - 4k(-1)^{n-1} u(n-1) + 4k(-1)^{n-2} u(n-2)$$

$$= (-1)^n u(n) - (-1)^{n-1} u(n-1)$$

$$\text{For } n=2, k(1+4+4) = 2 \Rightarrow k = 2/9$$

$$y(n) = [c_1 2^n + c_2 n 2^n + 2/9 (-1)^n] u[n]$$

~~y = (a)~~

$$y(0) = 1 \quad y(1) = 2 \quad \text{then}$$

$$c_1 + 2/9 = 1$$

$$\Rightarrow c_1 = 7/9$$

$$2c_1 + 2c_2 - 2/9 = 2$$

$$c_2 = 1/3$$

## Part (B)

$$y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$$

The characteristic solution -

$$\lambda^2 - 0.7\lambda + 0.1 = 0$$

$$\lambda = 1/2, 1/5 \text{ Hence}$$

$$y_n(n) = c_1 \left(\frac{1}{2}\right)^n + c_2 \left(\frac{1}{5}\right)^n$$

with

$$x(n) = \delta(n), \text{ we have}$$

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0 \Rightarrow y(1) = 1.4$$

Hence  $c_1 + c_2 = 2$  and

$$\frac{1}{2}c_1 + \frac{1}{5}c_2 = 1.4 = \frac{7}{5}$$

$$c_1 + \frac{2}{5}c_2 = \frac{14}{5}$$

The eqn yield

$$c_1 = \frac{10}{3} \quad c_2 = \frac{4}{3}$$

$$h(n) = \left[ \frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

The step response

$$s(n) = \sum_{k=0}^n h(n-k),$$

$$= \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{4}{3} \left(\frac{1}{5}\right)^n (5^{n+1} - 1) u(n)$$

## Question 2 (a)

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

$$= \frac{A}{(1-2z^{-1})} + \frac{B}{(1-z^{-1})} + \frac{Cz^{-1}}{(1-z^{-1})^2}$$

$$A = 4 \quad B = -3, \quad C = 1$$

$$\text{Hence } x(n) = [4(2)^n - 3 - n]u(n).$$

## Question (2B)

Solution :-

we have .

$$x(n) = \frac{1}{2\pi j} \oint_C \frac{z^n - 1}{1 - az^{-1}} dz = \frac{1}{2\pi j} \oint_C \frac{z^{n+1} dz}{z - a}$$

where 'c' is a circle at radius greater than |a| we shall evaluate this integral using  $f(z) = z^n$  we distinguish two cases.

→ If  $n > 0$ ,  $f(z)$  has only zero and hence no pole inside 'c'. The only pole 'c' is  $z = a$  Hence

$$x(n) = f(z) = a^n \quad n > 0$$

→ If  $n < 0$   $f(z) = z^n$  has  $n+1$  order pole at  $z = 0$  which is also inside 'c'.

Thus there are contributions from both poles for  $n = -1$

we have

$$x(-1) = \frac{1}{2\pi j} \oint_C \frac{1}{z(z-a)} dz = \frac{1}{z-a} \Big|_{z=0} + \frac{1}{z^2} \Big|_{z=a} = 0$$

By continuity in the same way we can show that

$$x(n) = 0 \quad \text{for } n < 0$$

$$x(n) = a^n \quad (n).$$

## Question No 3(A)

Sol :-

At  $\omega = 0$  we have

$$H(0) = \frac{b_0}{(1-p)^2} = 1$$

Hence

$$b_0 = (1-p)^2$$

At  $\omega = \pi/4$ 

$$H\left(\frac{\pi}{4}\right) = \frac{(1-p)^3}{(1-pe^{-j\pi/4})^2}$$

$$= \frac{(1-p)^2}{(1-p\cos(\pi/4) + jP\sin(\pi/4))^2}$$

$$= \frac{(1-p)^2}{(1-p/\sqrt{2} + jP/\sqrt{2})^2}$$

Hence

$$\frac{(1-p)^2}{[(1-p/\sqrt{2})^2 + P^2/2]} = 1/2.$$

as equivalently

$$2(1-p)^2 = 1 + p^2 - \sqrt{2p}$$

The value of  $p = 0.32$  satisfied this equation - consequently the system function for the desired filter is

$$H(z) = \frac{0.46}{(1 - 0.32z^{-1})^2}$$

The SAMP principle can be applied for the design of bandpass filter.



## Question No 3(b)

Sol:

Clearly the filter must have pass at.

$$P_{b_1} = r e^{j\omega T}$$

and zeros at  $z=1$  and  $z=-1$

Consequently the System function

$$H(z) = G \frac{(z-1)(z+1)}{(z-jr)(z+jr)}$$

$$= G \frac{z^2 - 1}{z^2 + r^2}$$

The gain factor is determined by evaluating the frequency response  $H(e^{j\omega})$  at the filter at

$$\omega = \pi/2 \text{ thus.}$$

$$H(\pi/2) = G \frac{2}{1-r^2} = 1$$

$$G = \frac{1-r^2}{2}$$

The value of  $r$  is determined by evaluating  $H(\omega)$  at  $\omega = 4\pi/q$

Thus

$$\left| H\left(\frac{4\pi}{9}\right) \right|^2 = \frac{(1-r^2)^2}{4} \quad \frac{2-2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)} = \frac{1}{2}$$

or equivalently

$$1.94(1-r^2) = 1 - 1.88r^2 + r^4$$

The value of  $r^2 = 0.7$  satisfied this equation. Therefore the system function desired filter is

$$H(z) = \frac{0.15(1-z^{-2})}{1-0.7z^{-2}}$$

## Question No 4 (A)

A filter duration sequence of length 1 is given as.

$$x(n) = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{otherwise} \end{cases}$$

Solution :-  $\rightarrow$

The Fourier transform of this sequence

$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}}$$

$$= \frac{\sin(\omega L/2) e^{-j\omega(L-1)/2}}{\sin(\omega/2)}$$

The magnitude and phase of  $x(\omega)$  are illustrated  $L=10$ . The  $N$ -Point DFT of  $x(n)$  is simply  $x(\omega)$  evaluated at the set of  $N$  equally spaced frequencies  $k=0, 1, \dots, N-1$  Hence.

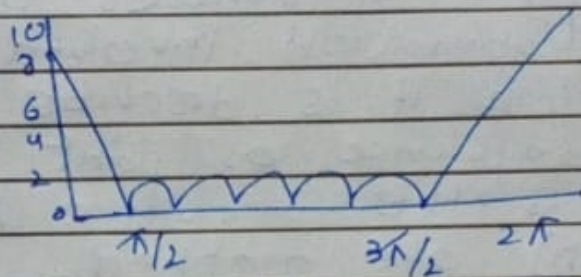
$$X(k) = \frac{1 - e^{-j2\pi k L/N}}{1 - e^{-j2\pi k/N}} \quad k=0, 1, \dots, N-1$$

$$= \frac{\sin(\pi k L / N)}{\sin(\pi k / N)} e^{-j\pi k (L-1) / N}$$

If  $N$  is selected such

that  $N=L$  then DFT

$$X(k) = \begin{cases} C & k=0 \\ 0 & k=1, 2, \dots, L-1 \end{cases}$$



Thus there is only one non zero value in DFT this apparent from observation of  $X(\omega)$  since  $X(\omega) = 0$  at the frequency  $\omega_k = 2\pi k / L$

$$k=0.$$

## QUESTION No 4(b)

$$x_1(n) = \{2, 1, 2, 1\}$$

$$x_2(n) = \{1, 2, 3, 4\}$$

Sol: →

Each sequence consists of 4 points for the purposes of illustrating the operation involved in circular convolution it is describe to graph each sequence as point on a circle. This is the sequence  $x_1(n)$  and  $x_2(n)$  are graphed.

Now  $x_3(m)$  is obtained by circulating value  $x_1(n)$  and  $x_2(n)$

$$x_3(0) = \sum_n x_1(n) x_2((1-n))_3$$

$x_2((1-n))_3$  is simply sequence  $x_2(n)$  folded and graph in a circle

$$x_3(0) = 14$$

For  $m=1$  we have.

$$x_3(1) = \sum_{n=0}^3 x_1(n) x_2((1-n))_4$$

It is easily verified that  $x_2((1-n))_4$  is simply the sequence.

$X_2((1-n))_4$  rotated by one unit in time as illustrated

for  $m=2$

we have

$$X_3(2) = \sum_{n=0}^3 X_1(n) v_2((2-n))_4$$

Now  $X_2((2-n))_4$  is the folded sequence.