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Subject

= DSP

Question no 1 (a).

Determine the response  $y(n)$ ,  $n \geq 0$  of system.

Sol:

The characteristic equation is.

$$\lambda^2 - 4\lambda + 4 = 0.$$

$$\lambda = 2, 2 \text{ hence.}$$

$$y_h(n) = C_1 2^n + C_2 n 2^n$$

The particular solution is.

$$y_p(n) = K(-1)^n u(n).$$

Substituting the solution into difference equation we obtain.

$$K(-1)^n u(n) - 4K(-1)^{n-1} u(n-1) + 4K(-1)^{n-2} u(n-2) =$$

$$(-1)^n u(n) - (-1)^{n-1} u(n-1).$$

$$\text{For } n=2, K(1+4+4) = 2 \Rightarrow K = \frac{2}{9}$$

The total solution is

$$y(n) = \left( C_1 2^n + C_2 n 2^n + \frac{2}{9} (-1)^n \right) u(n).$$

From the initial condition we obtain.

$$y(-1) = y(-2) = 0.$$

$$c_1 + \frac{2}{9} = 0$$

$$c_1 = -\frac{2}{9}$$

$$2c_1 + 2c_2 = \frac{2}{9} = 0.$$

$$2c_2 = \frac{2}{9} - 2c_1.$$

$$c_1 = -\frac{2}{9} \quad \text{and} \quad \text{cancel}$$

$$2c_2 = \frac{2}{9} - 2\left(-\frac{2}{9}\right)$$

$$= \frac{2}{9} + \frac{4}{9} = \frac{2+4}{9}$$

$$2c_2 = \frac{6}{9}$$

$$c_2 = \frac{6}{9 \times 2} = \frac{3}{9}.$$

$$c_2 = \frac{1}{3}$$

Question NO 1 (b)  
 Determine the Impulse response and unit step response.

$$y(n) - 0.7y(n-1) + 0.1y(n-2) = 2x(n) - x(n-2)$$

Sol:

$$\lambda^n - 0.7\lambda^{n-1} + 0.1\lambda^{n-2} = 0$$

$$\lambda^{2-2} [\lambda^2 - 0.7\lambda + 0.1] = 0$$

$$\lambda^2 - 0.5\lambda - 0.2\lambda + 0.1 = 0$$

$$\lambda(\lambda - 0.5) - 0.1(\lambda - 0.5) = 0$$

$$(\lambda - 0.5)(\lambda - 0.1) = 0$$

$$\lambda = 0.5 \quad \lambda = 0.1$$

General form of the solution to be homogeneous equation is.

$$y_h(n) = C_1 \lambda_1^n + C_2 \lambda_2^n$$

$$y_h(n) = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{1}{2}\right)^n$$

with  $x(n) = \delta(n)$  we have

$$y(0) = 2$$

$$y(1) - 0.7y(0) = 0 \Rightarrow y(1) = 1.4$$

Hence  $C_1 + C_2 = 2$  and

Impulse response.

$$\frac{1}{2} C_1 + \frac{1}{2} C_2 = 1.4 = \frac{7}{5}$$

$$c_1 + \frac{2}{5} c_2 = \frac{14}{5}$$

These equation yield

$$c_1 = \frac{10}{3}$$

$$c_2 = -\frac{4}{3}$$

$$h(n) = \left[ \frac{10}{3} \left(\frac{1}{2}\right)^n - \frac{4}{3} \left(\frac{1}{5}\right)^n \right] u(n)$$

Step response is:

$$f(n) = \sum_{k=0}^n h(n-k) \Rightarrow \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2}\right)^{n-k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5}\right)^{n-k}$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5}\right)^n \sum_{k=0}^n 5^k$$

$$= \frac{10}{3} \left(\frac{1}{2}\right)^n (2^{n+1} - 1) u(n) - \frac{4}{3} \left(\frac{1}{5}\right)^n (5^{n+1} - 1) u(n)$$

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Question no 2 (a).

Determine the casual signal  $x(n]$

$$X(z) = \frac{1}{(1-2z^{-1})(1-z^{-1})^2}$$

Sol

$$\frac{X(z)}{z} = \frac{z^2}{(z-1)(z-1)^2}$$

$$\frac{X(z)}{z} = \frac{A}{z-1} + \frac{B}{z-1} + \frac{C}{(z-1)^2}$$

Find A, B and C.

$$A = 4$$

$$B = 3$$

$$C = -1$$

hence  $x(n] = [4(2)^n - 3 - n]u(n]$ .

## Question no 2 (b).

Solution:

we have

$$x(n) = \frac{1}{2\pi j} \oint \frac{z^{n-1}}{1-a_3^{-1}} dz$$

$$= \frac{1}{2\pi j} \oint \frac{z^n dz}{z-a}$$

where  $C$  is a circle of radius greater than  $|a|$  we shall evaluate this integral.

1) if  $n \geq 0$   $f(z)$  has only zeros

and have no poles inside  $C$ . The only pole inside  $C$  is  $z=a$  hence

$$x(n) = f(z_0) = a^n \quad n \geq 0$$

If  $n < 0$   $f(z) = z^n$  has an  $n$ th-order pole at zero which is also inside  $C$ . Thus there are contribution from both poles. For  $n = -1$  we have.

$$x(-1) = \frac{1}{2\pi j} \oint \frac{1}{z^2(z-a)} dz$$

$$\frac{d}{dz} \left( \frac{1}{z-a} \right) \Big|_{z=0} + \frac{1}{z^2} \Big|_{z=a} = 0$$

By continuing in the  
same way - we can  
show that.

$$x(n) = 0.$$

for  $n < 0$  thus

$$x(n) = a^n u(n).$$

\_\_\_\_\_ x \_\_\_\_\_

Question No 3 (a).

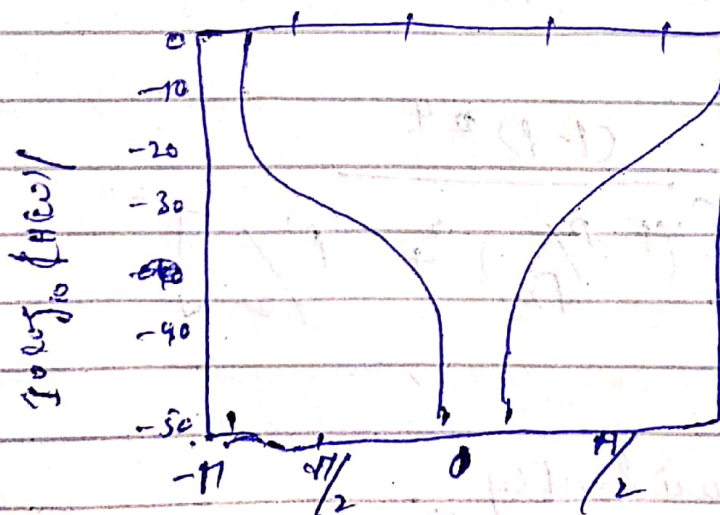
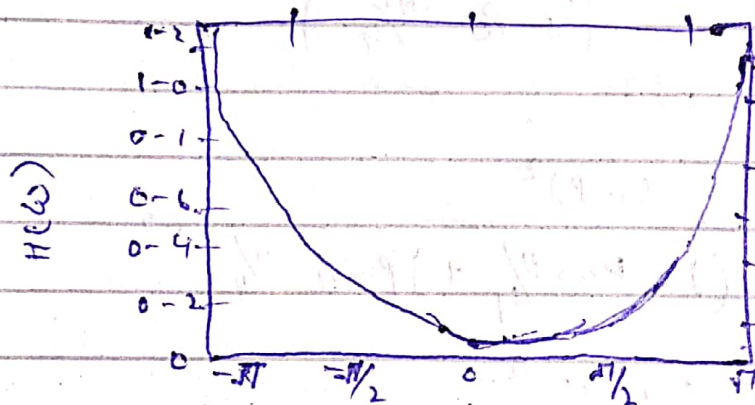
A two-pole low pass has the system response ---  $H(z) = \frac{b_0}{(1-pz^{-1})^2}$  ---

Sol.

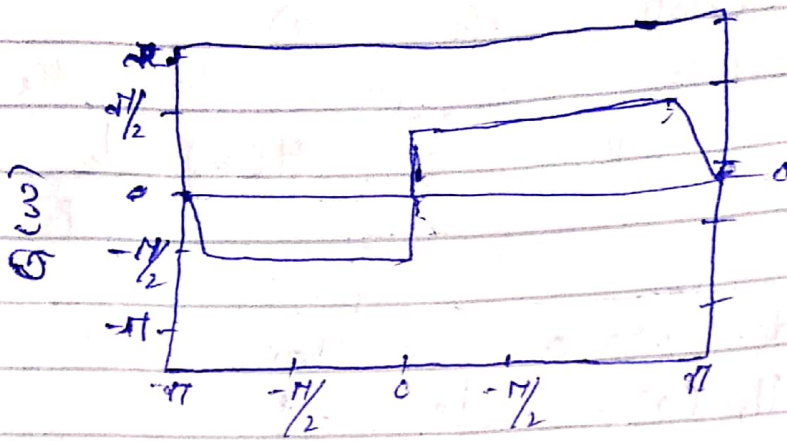
At  $\omega = 0$  we have

$$H_2(0) = \frac{b_0}{(1-p)^2} = 1$$

$$b_0 = (1-p)^2$$







At  $\omega = \pi/4$

$$H(\pi/4) = \frac{(1-P)^2}{(1 - Pe^{-j\pi/4})^2}$$

$$= \frac{(1-P)^2}{(1 - P \cos \pi/4 + jP \sin \pi/4)^2}$$

$$= \frac{(1-P)^2}{(1 - P/\sqrt{2} + jP/\sqrt{2})^2}$$

Hence =  $\frac{(1-P)^2}{\left[ \left( \frac{1-P}{\sqrt{2}} \right)^2 + \frac{P^2}{2} \right]^{1/2}} = \frac{1}{2}$

or equivalently

$$\sqrt{2} (1-P)^2 = 1 + P^2 - \sqrt{2}P$$

The value of  $P = 0.32$  satisfy this equation. The system  $f/m$  is

the desired filter is

$$H(z) = \frac{0.46}{(1 - 0.32z^{-1})^2}$$

The same principle can be applied for design of band pass filters.

x ————— x

### Question 3(b):

Design a Two-Pole band pass filter.

Solution:

clearly, The filter must have poles at  $p_{1,2} = \pm j\omega_c/2$  and

Zeros at  $z=1$  and  $z=-1$

consequently the system f/m is:

$$H(z) = \frac{g(z-1)(z+1)}{(z-j\omega_c/2)(z+j\omega_c/2)}$$
$$\Rightarrow G = \frac{(z^2-1)}{(z^2-\omega_c^2/4)}$$

The Gain factor is determined by evaluating the frequency response  $|H(\omega)|$  of the filter at  $\omega = \omega_c/2$ . Thus we have,

$$H(\omega_c/2) = g \frac{2}{1-\omega_c^2/4} = 1$$
$$g = \frac{1-\omega_c^2/4}{2}$$

The value of  $\omega_c$  is determined by evaluating  $H(\omega)$  at  $\omega = \omega_c/2$ . Thus we have.

(2)

$$\left[ H\left(\frac{4\pi}{9}\right) \right]^2 = \frac{(1-r^2)^2}{4} \frac{2 - 2\cos(8\pi/9)}{1+r^4+2r^2\cos(8\pi/9)}$$

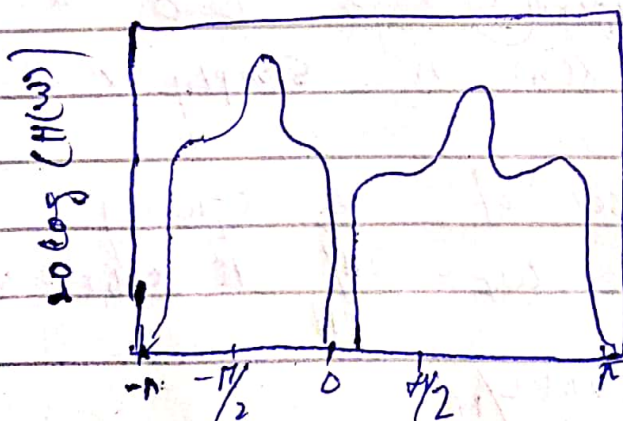
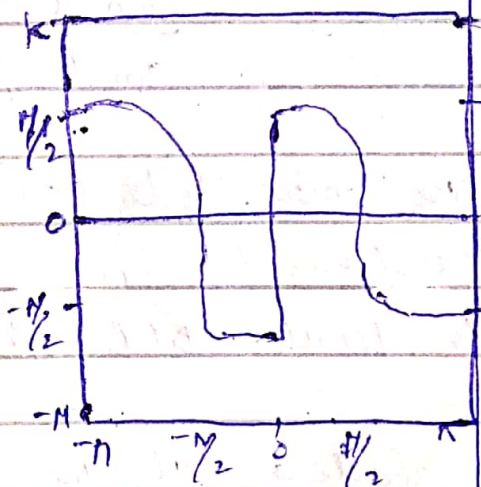
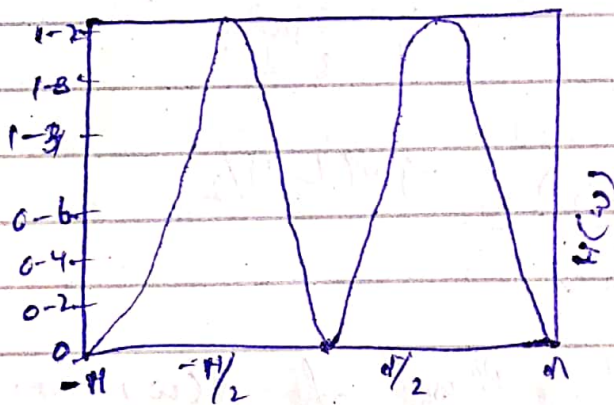
$$= \frac{1}{2}$$

$$\text{or } 1.94(1-r^2)^2 = 1 - 1.08r^2 + r^4$$

The value of  $r^2 = 0.7$  satisfy this equation.

The system f/m for the design filter is -

$$H(z) = \frac{0.15(1-z^2)}{(1+0.7z^2)}$$



Question no 4 (a)

A finite duration sequence of length  $L$  is given as

$$x(n) = \begin{cases} 1 & 0 \leq n \leq L-1 \\ 0 & \text{otherwise} \end{cases}$$

Determine  $N$  point DFT for this sequence for  $N \geq L$ .

Sol:

The Fourier Transform of this sequence is

$$X(\omega) = \sum_{n=0}^{L-1} x(n) e^{-j\omega n}$$

$$= \sum_{n=0}^{L-1} e^{-j\omega n} = \frac{1 - e^{-j\omega L}}{1 - e^{-j\omega}} \Rightarrow$$

$$= \frac{\sin(\omega L/2)}{\sin(\omega/2)} e^{-j\omega(L-1)/2}$$

The magnitude and phase of  $X(\omega)$  are illustrated in figure (1) for  $L=8$ . The  $N$ -point DFT of  $x(n)$  is simply  $X(\omega)$ .

evaluated at the set of  $N$  equally spaced frequencies  $\omega_k = \frac{2\pi k}{N}$   $k = 0, 1, 2, \dots, N-1$

$$X(k) = \frac{1 - e^{-j2\pi k L/N}}{1 - e^{-j2\pi k/N}} \quad k = 0, 1, 2, \dots, N-1$$

$$= \frac{\sin(\pi k L/N)}{\sin(\pi k/N)} e^{-j\pi k(L-1)/N}$$

②

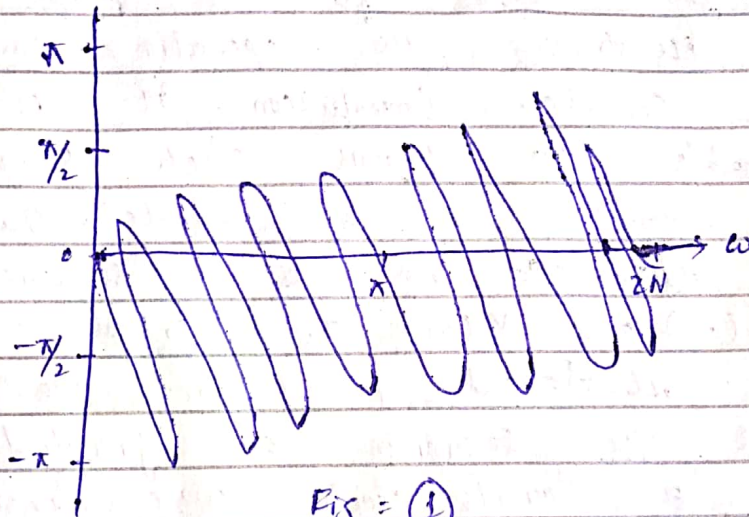
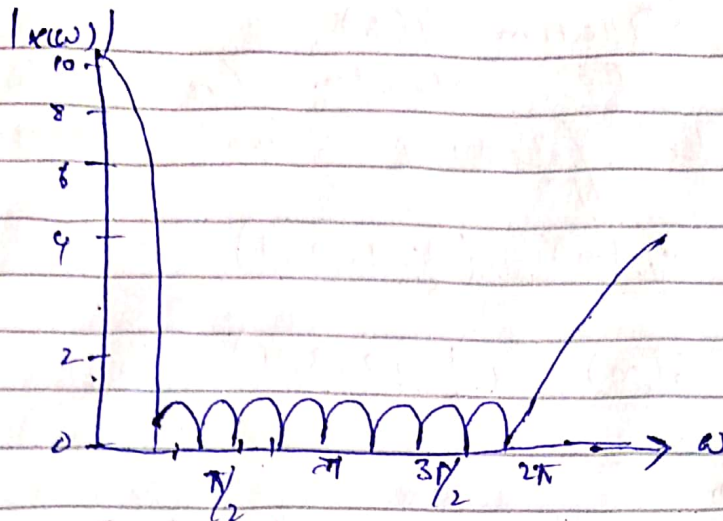


Fig = ①

If  $N$  is selected such that  $N=L$ , then DFT become,

$$X(k) = \begin{cases} L & k=0 \\ 0 & k=1, 2, 3, \dots, L-1 \end{cases}$$

Thus there is only one non zero value in the DFT. This is apparent from observation of  $x(\omega) = 0$  at frequencies  $\omega_k = \frac{2\pi k}{L}$   $k \neq 0$  then reader should verify that  $x(n)$  can be cat recovered from  $X(k)$  by performing an  $L$ -Point IDFT.

(3)

Question 4(b).  
Perform the circular convolution of the following two sequences.

$$x_1(n) = (2, 1, 2, 1)$$

$$x_2(n) = (1, 2, 3, 4)$$

Solution.

Each equation consist of four non-zero points. for this purpose of illustrating the operation involved in circular convolution. It is desirable to graph each sequence

as point on a circle. Thus as illustrated in the sequence

$x_1(n)$  and  $x_2(n)$  are graph

as illustrated in figure. we note that the sequence are graphed in a counter clockwise direction

on a circle. Now  $x_3(n)$  is obtained

by circularly convolving  $x_1(n)$  and  $x_2(n)$  as specified by beginning with  $n=0$  we have.

$$x_3(0) = \sum_{n=0}^3 x_1(n) x_2((-n))_N$$

$x_2((-n))_N$  is simply the sequence

$x_2(n)$  folded and graphed on circle illustrated in fig.

(9)

The product sequence is obtained by multiplying  $x_1(n)$  with  $x_2(l-n)$  point by point. This sequence

is illustrated in Fig. ---

Finally we sum the value in the product sequence to obtain.

$$x_3(0) = 14$$

for  $m=1$  we have

$$x_3(1) = \sum_{n=0}^3 x_1(n)x_2(1-n)$$

It is easily verified that  $x_2(l-n)$  is simply the sequence  $x_2(l-n)$  rotated counter clockwise by one unit in time as illustrated. This illustrated sequence multiplies  $x_1(n)$  to yield the product sequence.

$$x_3(1) = 16$$

$m=2$  we have

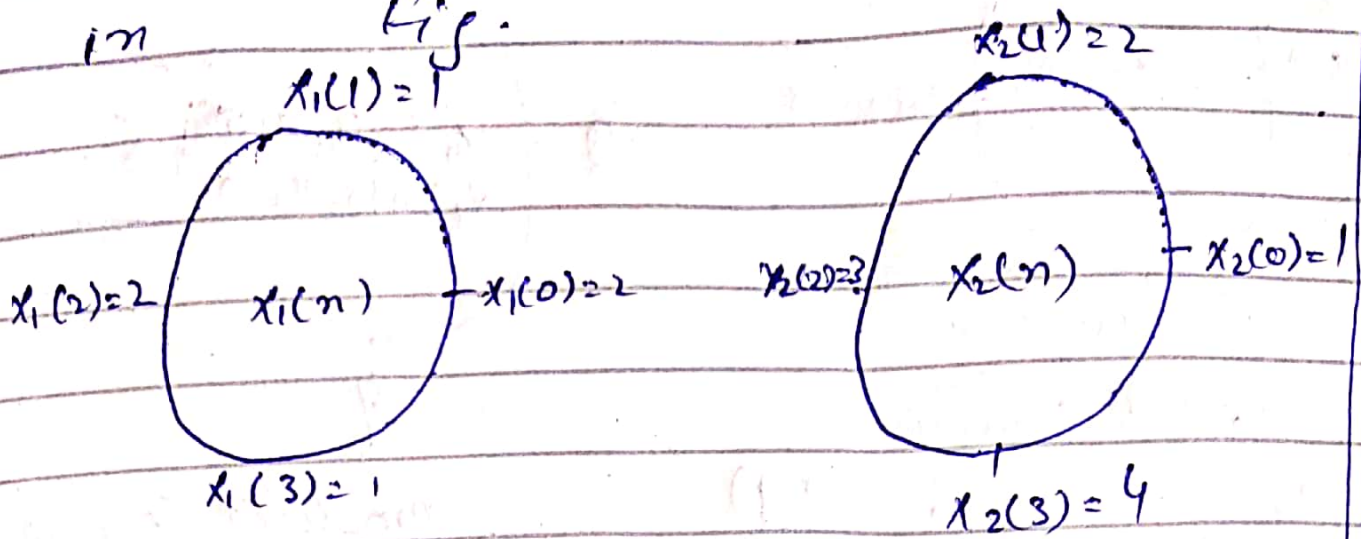
$$x_3(2) = \sum_{n=0}^3 x_1(n)x_2(2-n)$$

Now  $x_2(2-n)$  is the folded sequence in Fig. rotated two units of time in the counterclockwise direction.



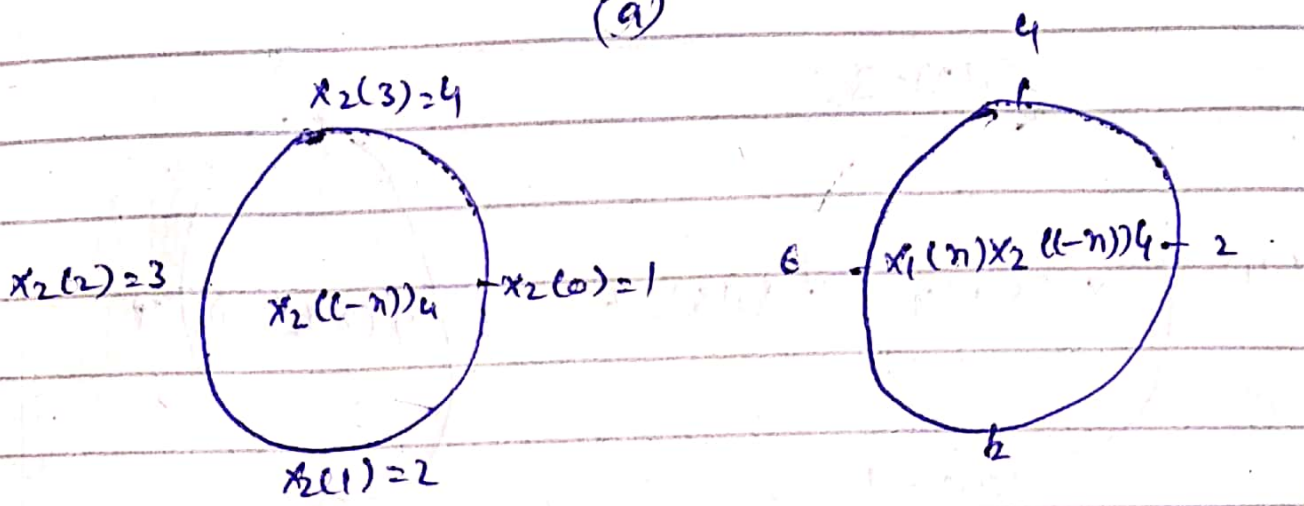
The resultant sequence is illustrated

in Fig.



Folded sequence.

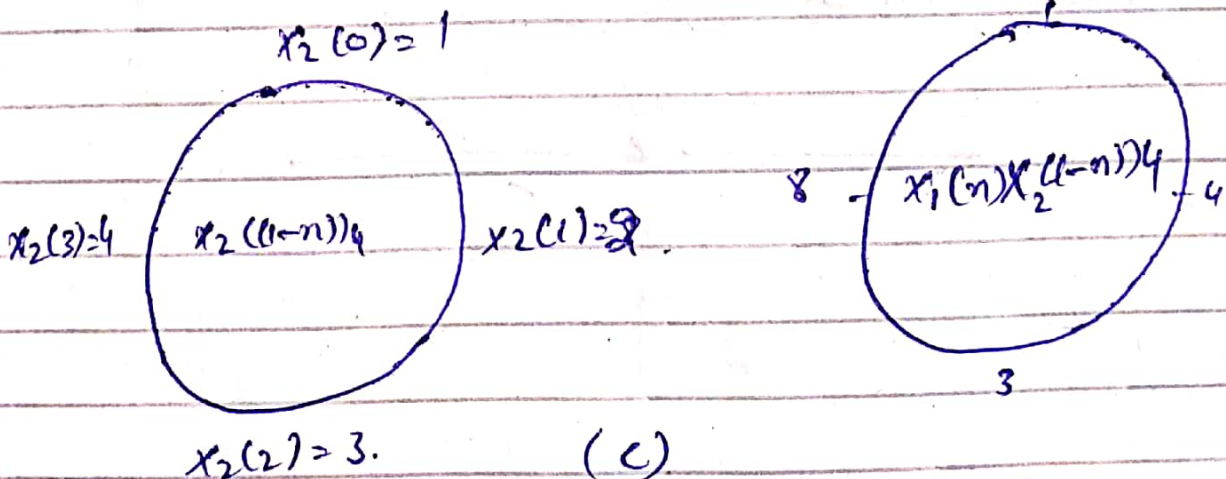
(a)



Folded Sequence

Product Sequence

(b)

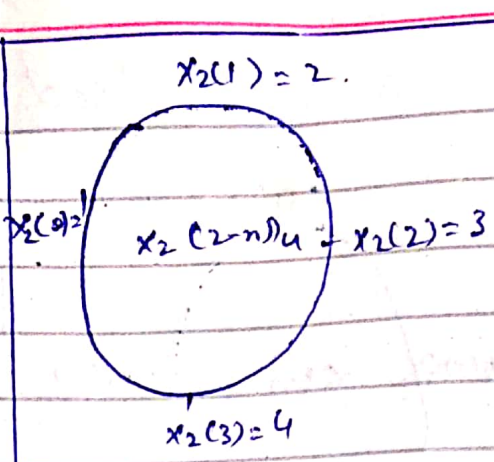


Folded sequence

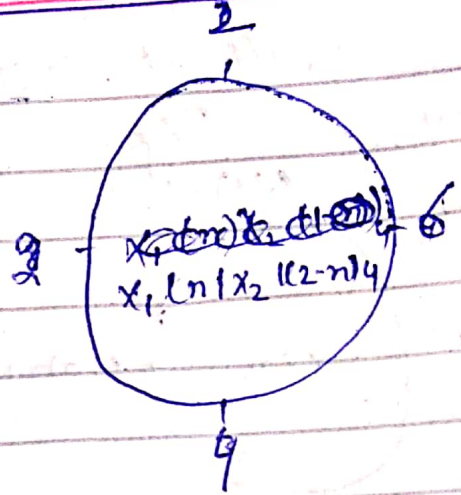
Product sequence

(c)

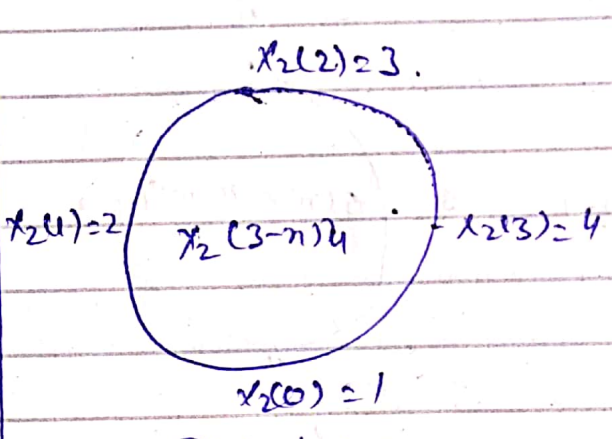
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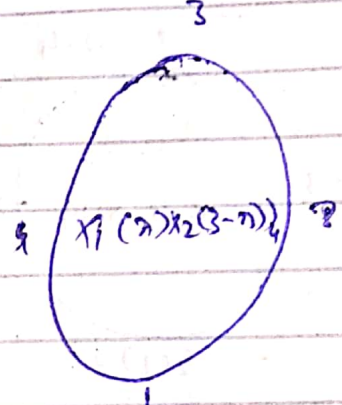
Folded sequence  
 by two unit in (d)  
 Time



product sequence



Folded sequence  
 rotated by 3  
 times Units.



product sequence

————— x —————